

# CC3235S 和 CC3235SF SimpleLink™ Wi-Fi® 双频带单芯片解决方案

## 1 器件概述

### 1.1 特性

- 多核心架构，片上系统 (SoC)
- 802.11 a/b/g/n: 2.4GHz 和 5GHz
- FIPS 140-2 1 级认证
- 多层安全 特性，帮助开发人员保护身份信息、数据和软件 IP
- 低功耗模式，适用于电池供电应用
- 与 2.4GHz 无线电共存
- 工业温度范围：-40°C 至 85°C
- 获得 Wi-Fi Alliance™ 的 Wi-Fi CERTIFIED 认证®
- 应用微控制器子系统：
  - Arm® Cortex®-M4 内核，运行频率 80MHz
  - 用户专用存储器
    - 256KB RAM
    - 可选的 1MB 可执行闪存
  - 多种外设和计时器
  - 27 个带灵活多路复用选项的 I/O 引脚
    - UART、I2S、I2C、SPI、SD、ADC，8 位并行接口
    - 计时器和 PWM
- Wi-Fi 网络处理器子系统：
  - Wi-Fi® 内核：
    - 802.11 a/b/g/n 2.4GHz 和 5GHz
    - 模式：
      - 接入点 (AP)
      - 基站 (STA)
      - Wi-Fi Direct® (仅在 2.4GHz 受支持)
    - 安全性：
      - WEP
      - WPA™/ WPA2™PSK
      - WPA2 企业
  - 互联网和应用协议：
    - HTTP 服务器、mDNS、DNS-SD 和 DHCP
    - IPv4 和 IPv6 TCP/IP 堆栈
    - 16 BSD 套接字 (完全安全的 TLS v1.2 和 SSL 3.0)
  - 内置的电源管理子系统：
    - 可配置的低功耗配置 (始终、间歇性、标签)
    - 高级低功耗模式
    - 集成式直流/直流稳压器
- 多层安全 特性：
  - 独立执行环境
  - 网络安全
  - 设备身份和密钥
  - 硬件加速器加密引擎 (AES、DES、SHA/MD5 和 CRC)
  - 应用级安全 (加密、身份验证、访问控制)
  - 初始安全编程
  - 软件篡改检测
  - 安全引导
  - 证书注册请求 (CSR)
  - 每个设备具有唯一密钥对
- 应用吞吐量：
  - UDP: 16Mbps, TCP: 13Mbps
  - 峰值: 72Mbps
- 电源管理子系统：
  - 集成式直流/直流转换器支持宽电源电压范围：
    - VBAT 宽电压模式: 2.1V 至 3.6V
    - VIO 始终与 VBAT 关联
  - 高级低功耗模式：
    - 关断: 1µA, 休眠: 4.5µA
    - 低功耗深度睡眠 (LPDS): 120µA
    - 空闲连接 (MCU 处于 LPDS 状态): 710µA
    - RX 流量 (MCU 处于活动模式): 59mA
    - TX 流量 (MCU 处于活动模式): 223mA
- Wi-Fi TX 功率：
  - 2.4GHz: 1 DSSS 时为 18.0dBm
  - 5GHz: 6 OFDM 时为 18.1dBm
- Wi-Fi RX 灵敏度：
  - 2.4GHz: 1 DSSS 时为 -96dBm
  - 5GHz: 6 OFDM 时为 -92dBm
- 时钟源：
  - 具有内部振荡器的 40.0MHz 晶体
  - 32.768kHz 晶体或外部 RTC
- RGK 封装
  - 64 引脚 9mm × 9mm 极薄四方扁平无引线 (VQFN) 封装, 0.5mm 间距
- 器件支持 SimpleLink™ MCU 平台开发人员生态系统



## 1.2 应用

- 适用于物联网的应用, 例如:
  - 楼宇和住宅自动化:
    - HVAC 系统和恒温器
    - 视频监控、可视门铃和低功耗摄像头
    - 楼宇安全系统和电子锁
  - 电器
  - 资产跟踪
  - 工厂自动化
  - 医疗和保健
  - 电网基础设施

## 1.3 说明

双频带无线 MCU CC3235x 器件有两种型号: CC3235S 和 C3235SF。

- **CC3235S** 包括 256KB RAM、IoT 网络安全性、设备身份/密钥以及 MCU 级安全特性, 如文件系统加密、用户 IP (MCU 图像) 加密、安全启动和调试安全性。
- **CC3235SF** 基于 CC3235S 而构建, 除了 256KB RAM 以外, 还集成了一个用户专用的 1MB 可执行闪存。

使用 Wi-Fi CERTIFIED™ 无线微控制器 (MCU) 简化您的 IoT 设计。SimpleLink™ Wi-Fi® CC3235x 器件系列是一种双频带片上系统 (SoC) 解决方案, 将两个处理器集成在一个芯片上, 包括:

- 应用处理器: Arm® Cortex®-M4 MCU, 具有用户专用的 256KB RAM 和可选的 1MB 可执行闪存
- 用以运行所有 Wi-Fi 和互联网逻辑层的网络处理器此基于 ROM 的子系统完全减轻了主机 MCU 的负载, 包括一个 802.11 a/b/g/n 双频带 2.4GHz 和 5GHz 无线电、基带和带有强大硬件加密引擎的 MAC

这些器件引进了可进一步简化物联网连接的新功能。主要的新特性包括:

- 802.11a (5GHz) 支持
- 与 BLE/2.4GHz 无线电共存
- 天线选择
- 安全性更强, 经过 FIPS 140-2 1 级认证和其他认证 <sup>(1)</sup>
- 可同时打开多达 16 个安全套接字
- 证书注册请求 (CSR)
- 在线证书状态协议 (OCSP)
- 获得 Wi-Fi® Alliance 认证的 IoT 省电特性 (如 DMS、代理 ARP 等)
- 降低模板包传输负载的无主机模式
- 改善了快速扫描

CC3235x 器件系列是 SimpleLink™ MCU 平台的一部分, 该平台是一个常见、易用的开发环境, 基于一个单核软件开发套件 (SDK), 具有丰富的工具集和参考设计。E2E™ 社区可支持 Wi-Fi®、低功耗蓝牙®、1GHz 以下器件和主机 MCU。关于更多信息, 请访问 [www.ti.com/simplelink](http://www.ti.com/simplelink) 或 [www.ti.com.cn/simplelinkwifi](http://www.ti.com.cn/simplelinkwifi)。

(1) 关于特定器件号的具体 FIPS 认证状态, 请参考 <https://csrc.nist.gov/publications/fips>。

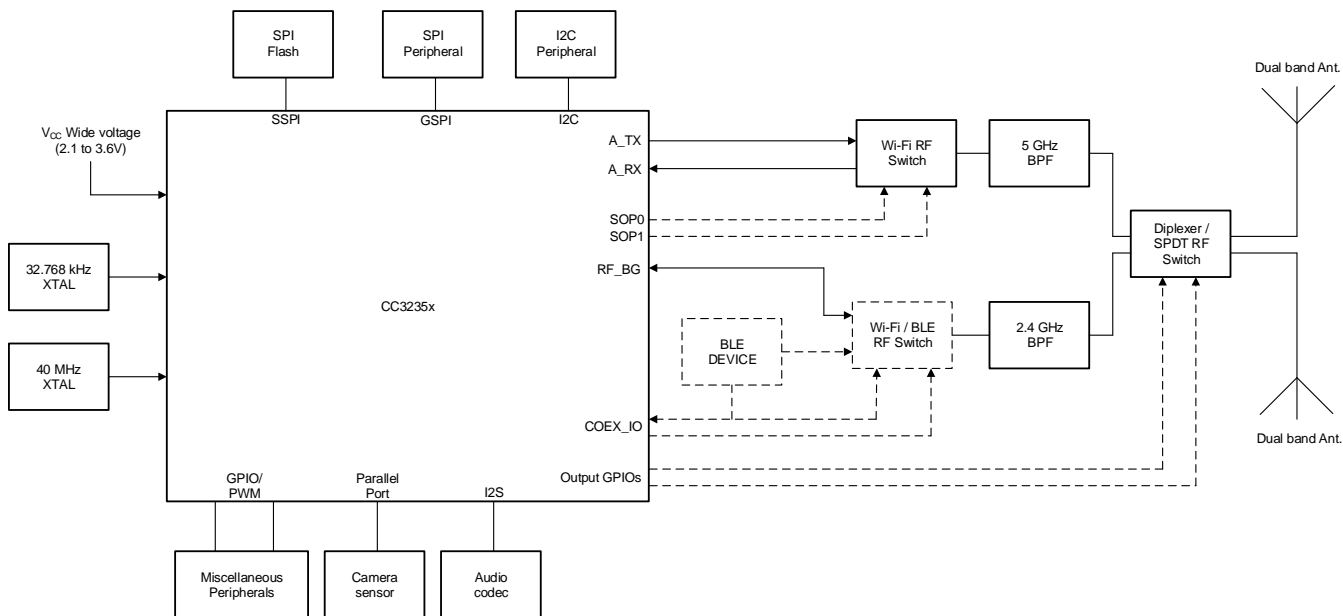
器件信息<sup>(1)</sup>

器件型号	封装	封装尺寸 (标称值)
CC3235SM2RGKR	VQFN (64)	9.00mm x 9.00mm
CC3235SF12RGKR	VQFN (64)	9.00mm x 9.00mm

(1) 有关所有可选封装, 请参阅 节 9。

### 1.4 功能方框图

图 1-1 显示了 CC3235x SimpleLink™ Wi-Fi® 解决方案的功能方框图。



注意：双信器用于信号天线解决方案。使用天线选择功能（双天线）时，需要在双信器后应用 1 个 SPDT 开关和 2 条 GPIO 线路。

图 1-1. 功能方框图

图 1-2 为 CC3235x 器件的硬件概览。

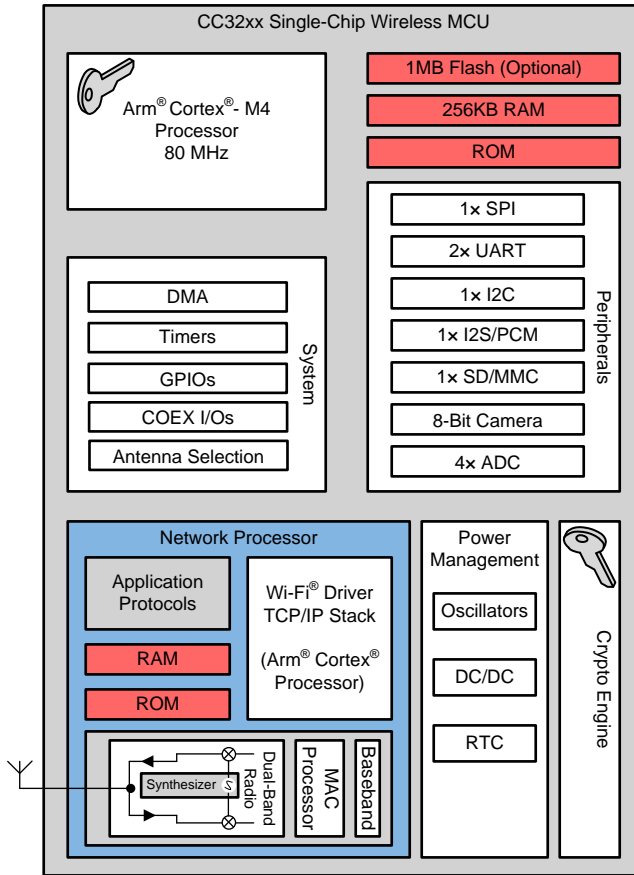


图 1-2. CC3235x 硬件概览

图 1-3 为 CC3235x 器件中的嵌入式软件概览。

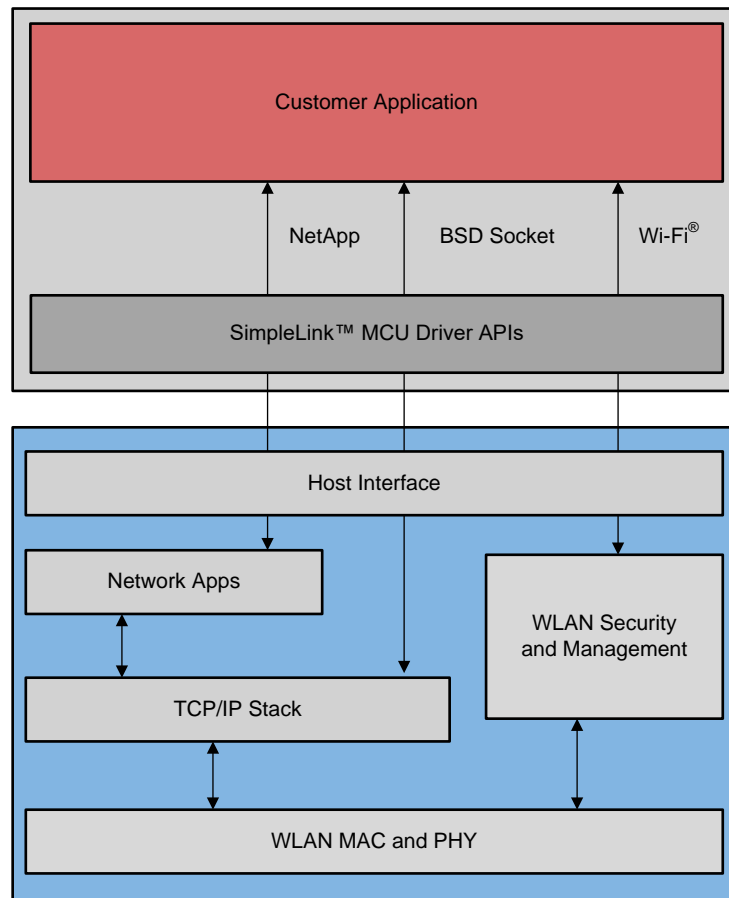


图 1-3. CC3235x 嵌入式软件概览

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## 2 修订历史记录

### Changes from Original (January 2019) to Revision A

Page

- Changed Operating Frequency Range in WLAN Transmitter Characteristics: 5 GHz Band table from 5845 to 5825. [40](#)
- 更新了 FIPS 和电源管理系统链接。 [92](#)

### 3 Device Comparison

Table 3-1 lists the features supported across different CC3x35 devices.

**Table 3-1. Comparison of Device Features**

FEATURE	DEVICE		
	CC3135	CC3235S	CC3235SF
Classification	Network processor	Wireless microcontroller	Wireless microcontroller
Standard	802.11a/b/g/n	802.11a/b/g/n	802.11a/b/g/n
TCP/IP stack	IPv4, IPv6	IPv4, IPv6	IPv4, IPv6
Sockets	16	16	16
Package	9-mm x 9-mm VQFN	9-mm x 9-mm VQFN	9-mm x 9-mm VQFN
<b>ON-CHIP APPLICATION MEMORY</b>			
Flash	—	—	1MB
RAM	—	256KB	256KB
<b>RF FEATURES</b>			
Frequency	2.4 GHz, 5 GHz	2.4 GHz, 5 GHz	2.4 GHz, 5 GHz
Coexistence with BLE Radio	Yes	Yes	Yes
<b>SECURITY FEATURES</b>			
Additional networking security	Unique device identity Trusted root-certificate catalog TI Root-of-trust public key Online certificate status protocol (OCSP) Certificate signing request (CSR) Unique per-device key pair	Unique device identity Trusted root-certificate catalog TI Root-of-trust public key Online certificate status protocol (OCSP) Certificate signing request (CSR) Unique per-device key pair	Unique device identity Trusted root-certificate catalog TI Root-of-trust public key Online certificate status protocol (OCSP) Certificate signing request (CSR) Unique per-device key pair
Hardware acceleration	Hardware crypto engines	Hardware crypto engines	Hardware crypto engines
Secure boot	—	Yes	Yes
Enhanced application level security	—	File system security Secure key storage Software tamper detection Cloning protection Initial secure programming	File system security Secure key storage Software tamper detection Cloning protection Initial secure programming
FIPS 140-2 Level 1 Certification <sup>(1)</sup>	Yes	Yes	Yes

(1) For exact status of FIPS certification for a specific part number, please refer to <https://csrc.nist.gov/publications/fips>.

### 3.1 Related Products

For information about other devices in this family of products or related products, see the links that follow.

**The SimpleLink™ MCU Portfolio** This portfolio offers a single development environment that delivers flexible hardware, software, and tool options for customers developing wired and wireless applications. With 100 percent code reuse across host MCUs, Wi-Fi®, Bluetooth® low energy, Sub-1 GHz devices and more, choose the MCU or connectivity standard that fits your design. A one-time investment with the SimpleLink™ software development kit (SDK) allows you to reuse often, opening the door to create unlimited applications.

**SimpleLink™ Wi-Fi® Family** This device platform offers several Internet-on-a-chip™ solutions, which address the need of battery-operated, security-enabled products. Texas Instruments offers a single-chip wireless microcontroller and a wireless network processor that can be paired with any MCU, allowing developers to design new Wi-Fi® products or upgrade existing products with Wi-Fi® capabilities.

**BoosterPack™ Plug-in Module** Extend the functionality of the TI LaunchPad™ Development Kit with the BoosterPack™ Plug-in Module. The application-specific BoosterPack Plug-in Module allows you to explore a broad range of applications, including capacitive touch, wireless sensing, LED Lighting control, and more. Stack multiple BoosterPack Plug-in Modules onto a single LaunchPad Development Kit to further enhance the functionality of your design.

**Reference Designs for CC3200, CC3220 and CC3235 Devices** The TI Designs Reference Design Library is a robust reference design library spanning analog, embedded processor and connectivity. Created by TI experts to help you jump start your system design, all TI Designs include schematic or block diagrams, BOMs and design files to speed your time to market.

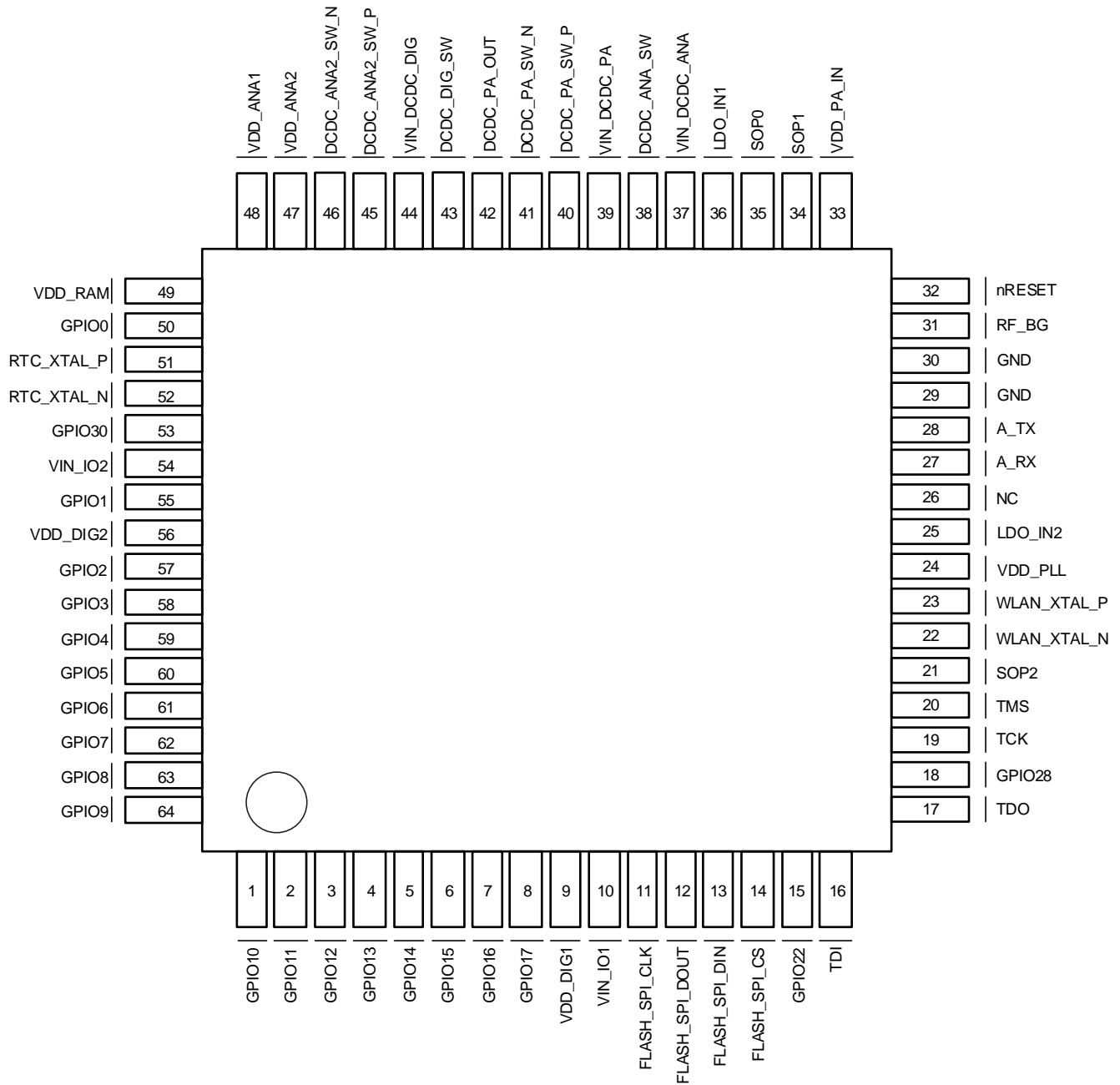
**The SimpleLink™ Wi-Fi® SDK** The SDK contains drivers for the CC3235 programmable MCU, sample applications, and documentation required to start development with CC3235x solutions.



## 4 Terminal Configuration and Functions

### 4.1 Pin Diagram

Figure 4-1 shows pin assignments for the 64-pin VQFN package.



NC = No internal connection

Figure 4-1. Top View Pin Assignment for 64-Pin VQFN

## 4.2 Pin Attributes

The device makes extensive use of pin multiplexing to accommodate the large number of peripheral functions in the smallest possible package. To achieve this configuration, pin multiplexing is controlled using a combination of hardware configuration (at device reset) and register control.

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### NOTE

TI highly recommends using the [Pin Mux Tool](#) to obtain the desired pinout. In addition refer to the user guide within the [SimpleLink™ CC32XX Software Development Kit \(SDK\)](#)

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The board and software designers are responsible for the proper pin multiplexing configuration. Hardware does not ensure that the proper pin multiplexing options are selected for the peripherals or interface mode used.

[Table 4-1](#) and [Table 4-2](#) list the pin descriptions and attributes. [Table 4-3](#) lists the signal descriptions. [Table 4-4](#) presents an overall view of pin multiplexing. All pin multiplexing options are configurable using the pin mux registers.

The following special considerations apply:

- All I/Os support drive strengths of 2, 4, and 6 mA. The drive strength is individually configurable for each pin.
- All I/Os support 10- $\mu$ A pullup and pulldown resistors.
- The  $V_{IO}$  and  $V_{BAT}$  supply must be tied together at all times.
- By default, all I/Os float in the Hibernate state. However, the default state can be changed by software.
- All digital I/Os are nonfail-safe.

---

### NOTE

If an external device drives a positive voltage to the signal pads and the CC3235x device is not powered, DC is drawn from the other device. If the drive strength of the external device is adequate, an unintentional wakeup and boot of the CC3235x device can occur. To prevent current draw, TI recommends any one of the following conditions:

- All devices interfaced to the CC3235x device must be powered from the same power rail as the chip.
- Use level shifters between the device and any external devices fed from other independent rails.
- The nRESET pin of the CC3235x device must be held low until the  $V_{BAT}$  supply to the device is driven and stable.
- All GPIO pins default to high impedance unless programmed by the MCU. The bootloader sets the TDI, TDO, TCK, TMS, and Flash\_SPI pins to mode 1. All the other pins are left in the Hi-Z state.

The ADC inputs are tolerant up to 1.8 V (see [Table 5-30](#) for more details about the usable range of the ADC). On the other hand, the digital pads can tolerate up to 3.6 V. Hence, take care to prevent accidental damage to the ADC inputs. TI recommends first disabling the output buffers of the digital I/Os corresponding to the desired ADC channel (that is, converted to Hi-Z state), and thereafter disabling the respective pass switches (S7 [Pin 57], S8 [Pin 58], S9 [Pin 59], and S10 [Pin 60]). For more information, see [Section 4.5](#).

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**Table 4-1. Pin Descriptions**

NO.	PINS		TYPE	DESCRIPTION	SELECT AS WAKEUP SOURCE	CONFIGURE ADDITIONAL ANALOG MUX	MUXED WITH JTAG
	NAME						
1	GPIO10		I/O	General-purpose input or output	No	No	No
2	GPIO11		I/O	General-purpose input or output	Yes	No	No
3	GPIO12		I/O	General-purpose input or output	No	No	No
4	GPIO13		I/O	General-purpose input or output	Yes	No	No
5	GPIO14		I/O	General-purpose input or output	No	No	No
6	GPIO15		I/O	General-purpose input or output	No	No	No
7	GPIO16		I/O	General-purpose input or output	No	No	No
8	GPIO17		I/O	General-purpose input or output	Yes	No	No
9	VDD_DIG1		Power	Internal digital core voltage	N/A	N/A	N/A
10	VIN_IO1		Power	I/O power supply (same as battery voltage)	N/A	N/A	N/A
11	FLASH_SPI_CLK		O	Serial flash interface: SPI clock	N/A	N/A	N/A
12	FLASH_SPI_DOUT		O	Serial flash interface: SPI data out	N/A	N/A	N/A
13	FLASH_SPI_DIN		I	Serial flash interface: SPI data in	N/A	N/A	N/A
14	FLASH_SPI_CS		O	Serial flash interface: SPI chip select	N/A	N/A	N/A
15	GPIO22		I/O	General-purpose input or output	No	No	No
16	TDI		I/O	JTAG interface: data input	No	No	Muxed with JTAG TDI
17	TDO		I/O	JTAG interface: data output	Yes	No	Muxed with JTAG TDO
18	GPIO28		I/O	General-purpose input or output	No	No	No
19	TCK		I/O	JTAG / SWD interface: clock	No	No	Muxed with JTAG/ SWD-TCK
20	TMS		I/O	JTAG / SWD interface: mode select or SWDIO	No	No	Muxed with JTAG/ SWD-TMSC
21 <sup>(1)</sup>	SOP2		O	Configuration sense-on-power	No	No	No
22	WLAN_XTAL_N		Analog	40-MHz XTAL	N/A	N/A	N/A
23	WLAN_XTAL_P		Analog	40-MHz XTAL or TCXO clock input	N/A	N/A	N/A
24	VDD_PLL		Power	Internal analog voltage	N/A	N/A	N/A
25	LDO_IN2		Power	Analog RF supply from analog DCDC output	N/A	N/A	N/A
26	NC		—	No Connect	N/A	N/A	N/A
27	A_RX		RF	RF A band: 5 GHz A_RX	N/A	N/A	N/A
28	A_TX		RF	RF A band: 5 GHz A_TX	N/A	N/A	N/A
29	GND		GND	Ground	N/A	N/A	N/A
30	GND		GND	Ground	N/A	N/A	N/A
31	RF_BG		RF	RF BG band: 2.4 GHz TX, RX	N/A	N/A	N/A
32	nRESET		I	Master chip reset input. Active low input.	N/A	N/A	N/A
33	VDD_PA_IN		Power	RF power amplifier (PA) input from PA DC-DC output	N/A	N/A	N/A
34	SOP1		O	Configuration sense-on-power and 5 GHz switch control	N/A	N/A	N/A

(1) This pin has dual functions: as a SOP[2] (device operation mode), and as an external TCXO enable. As a TXCO enable, the pin is an output on power up and driven logic high. During hibernate low-power mode, the pin is in a Hi-Z state but is pulled down for SOP mode to disable TCXO. Because of the SOP functionality, the pin must be used as an output only.

**Table 4-1. Pin Descriptions (continued)**

PINS		TYPE	DESCRIPTION	SELECT AS WAKEUP SOURCE	CONFIGURE ADDITIONAL ANALOG MUX	MUXED WITH JTAG
NO.	NAME					
35	SOP0	O	Configuration sense-on-power and 5 GHz switch control	N/A	N/A	N/A
36	LDO_IN1	Power	Analog RF supply from analog DCDC output	N/A	N/A	N/A
37	VIN_DCDC_ANA	Power	Analog DC-DC supply input (same as battery voltage)	N/A	N/A	N/A
38	DCDC_ANA_SW	Power	Analog DC/DC converter switching node	N/A	N/A	N/A
39	VIN_DCDC_PA	Power	PA DC/DC converter input supply (same as battery voltage)	N/A	N/A	N/A
40	DCDC_PA_SW_P	Power	PA DC/DC converter +ve switching node	N/A	N/A	N/A
41	DCDC_PA_SW_N	Power	PA DC/DC converter -ve switching node	N/A	N/A	N/A
42	DCDC_PA_OUT	Power	PA DC/DC converter output.	N/A	N/A	N/A
43	DCDC_DIG_SW	Power	Digital DC/DC converter switching node	N/A	N/A	N/A
44	VIN_DCDC_DIG	Power	Digital DC/DC converter supply input (same as battery voltage)	N/A	N/A	N/A
45	DCDC_ANA2_SW_P	I/O	Analog2 DCDC converter +ve switching node	No	User configuration not required <sup>(2)</sup>	No
46	DCDC_ANA2_SW_N	Power	Analog2 DC-DC converter -ve switching node	N/A	N/A	N/A
47	VDD_ANA2	Power	Analog2 DC-DC output	N/A	N/A	N/A
48	VDD_ANA1	Power	Analog1 power supply fed by ANA2 DC-DC output	N/A	N/A	N/A
49	VDD_RAM	Analog	SRAM LDO output	N/A	N/A	N/A
50	GPIO0	I/O	General-purpose input or output	No	User configuration not required <sup>(2)</sup>	No
51	RTC_XTAL_P	Analog	32.768-kHz XTAL_P or external CMOS level clock input	N/A	N/A	N/A
52	RTC_XTAL_N	Analog	32.768-kHz XTAL_N	N/A	User configuration not required <sup>(2)(3)</sup>	No
53	GPIO30	I/O	General-purpose input or output	No	User configuration not required <sup>(2)</sup>	No
54	VIN_IO2	Analog	Chip supply voltage (VBAT)	N/A	N/A	N/A
55	GPIO1	I/O	General-purpose input or output	No	No	No
56	VDD_DIG2	Analog	Internal digital core voltage	N/A	N/A	N/A
57	GPIO2	I/O	Analog input (1.5V max) or general-purpose input or output	Wake-up source	See <sup>(4)</sup>	No
58	GPIO3	I/O	Analog input (1.5V max) or general-purpose input or output	No	See <sup>(4)</sup>	No
59	GPIO4	I/O	Analog input (1.5V max) or general-purpose input or output	Wake-up source	See <sup>(4)</sup>	No
60	GPIO5	I/O	Analog input (1.5V max) or general-purpose input or output	No	See <sup>(4)</sup>	No
61	GPIO6	I/O	General-purpose input or output	No	No	No
62	GPIO7	I/O	General-purpose input or output	No	No	No
63	GPIO8	I/O	General-purpose input or output	No	No	No

(2) Device firmware automatically enables the digital path during ROM boot.

(3) To use the digital functions, RTC\_XTAL\_N must be pulled high to the supply voltage using a 100-kΩ resistor.

(4) Requires user configuration to enable the analog switch of the ADC channel (the switch is off by default.) The digital I/O is always connected and must be made Hi-Z before enabling the ADC switch.

**Table 4-1. Pin Descriptions (continued)**

PINS		TYPE	DESCRIPTION	SELECT AS WAKEUP SOURCE	CONFIGURE ADDITIONAL ANALOG MUX	MUXED WITH JTAG
NO.	NAME					
64	GPIO9	I/O	General-purpose input or output	No	No	No
GND_TAB		—	Thermal pad and electrical ground	N/A	N/A	N/A

**Table 4-2. Pin Attributes**

PIN NO.	SIGNAL NAME <sup>(1)</sup>	SIGNAL TYPE <sup>(2)</sup>	PIN MUX ENCODING	SIGNAL DIRECTION	PAD STATES		
					LPDS <sup>(3)</sup>	Hib <sup>(4)</sup>	nRESET = 0
1	GPIO10 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	I2C_SCL		1	I/O (open drain)	Hi-Z, Pull, Drive		
	GT_PWM06		3	O	Hi-Z, Pull, Drive		
	UART1_TX		7	O	1		
	SDCARD_CLK		6	O	0		
	GT_CCP01		12	I	Hi-Z, Pull, Drive		
2	GPIO11 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	I2C_SDA		1	I/O (open drain)	Hi-Z, Pull, Drive		
	GT_PWM07		3	O	Hi-Z, Pull, Drive		
	pXCLK(XVCLK)		4	O	0		
	SDCARD_CMD		6	I/O (open drain)	Hi-Z, Pull, Drive		
	UART1_RX		7	I	Hi-Z, Pull, Drive		
	GT_CCP02		12	I	Hi-Z, Pull, Drive		
MCAFSX	13	O	Hi-Z, Pull, Drive				
3	GPIO12 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	McACLK		3	O	Hi-Z, Pull, Drive		
	pVS(VSYNC)		4	I	Hi-Z, Pull, Drive		
	I2C_SCL		5	I/O (open drain)	Hi-Z, Pull, Drive		
	UART0_TX		7	O	1		
	GT_CCP03		12	I	Hi-Z, Pull, Drive		
4	GPIO13 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	I2C_SDA		5	I/O (open drain)			
	pHS(HSYNC)		4	I			
	UART0_RX		7	I			
	GT_CCP04		12	I			
5	GPIO14 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	I2C_SCL		5	I/O (open drain)			
	GSPI_CLK		7	I/O			
	pDATA8(CAM_D4)		4	I			
	GT_CCP05		12	I			

(1) Signals names with (PN) denote the default pin name.

(2) Signal Types: I = Input, O = Output, I/O = Input or Output.

(3) LPDS state: Unused I/Os are in a Hi-Z state. Software may program the I/Os to be input with pull or drive (regardless of active pin configuration), according to the need.

(4) Hibernate mode: The I/Os are in a Hi-Z state. Software may program the I/Os to be input with pull or drive (regardless of active pin configuration), according to the need.

Table 4-2. Pin Attributes (continued)

PIN NO.	SIGNAL NAME <sup>(1)</sup>	SIGNAL TYPE <sup>(2)</sup>	PIN MUX ENCODING	SIGNAL DIRECTION	PAD STATES		
					LPDS <sup>(3)</sup>	Hib <sup>(4)</sup>	nRESET = 0
6	GPIO15 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	I2C_SDA		5	I/O (open drain)			
	GSPI_MISO		7	I/O			
	pDATA9(CAM_D5)		4	I			
	GT_CCP06		13	I			
	SDCARD_DATA0		8	I/O			
7	GPIO16 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	GSPI_MOSI		7	I/O	Hi-Z, Pull, Drive		
	pDATA10(CAM_D6)		4	I	Hi-Z, Pull, Drive		
	UART1_TX		5	O	1		
	GT_CCP07		13	I	Hi-Z, Pull, Drive		
	SDCARD_CLK		8	O	0		
8	GPIO17 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	UART1_RX		5	I			
	GSPI_CS		7	I/O			
	pDATA11 (CAM_D7)		4	I			
	SDCARD_CMD		8	I/O			
9	VDD_DIG1 (PN)	—	N/A	N/A	N/A	N/A	N/A
10	VIN_IO1	—	N/A	N/A	N/A	N/A	N/A
11	FLASH_SPI_CLK	O	N/A	O	Hi-Z, Pull, Drive <sup>(5)</sup>	Hi-Z, Pull, Drive	Hi-Z
12	FLASH_SPI_DOUT	O	N/A	O	Hi-Z, Pull, Drive <sup>(5)</sup>	Hi-Z, Pull, Drive	Hi-Z
13	FLASH_SPI_DIN	I	N/A	I	Hi-Z, Pull, Drive <sup>(5)</sup>	Hi-Z	Hi-Z
14	FLASH_SPI_CS	O	N/A	O	1	Hi-Z, Pull, Drive	Hi-Z
15	GPIO22 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	McAFSX		7	O			
	GT_CCP04		5	I			
16	TDI (PN)	I/O	1	I	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	GPIO23		0	I/O			
	UART1_TX		2	O	1		
	I2C_SCL		9	I/O (open drain)	Hi-Z, Pull, Drive		
17	TDO (PN)	I/O	1	O	Hi-Z, Pull, Drive	Driven high in SWD; driven low in 4-wire JTAG	Hi-Z
	GPIO24		0	I/O			
	PWM0		5	O			
	UART1_RX		2	I			
	I2C_SDA		9	I/O (open drain)			
	GT_CCP06		4	I			
	McAFSX		6	O			
18	GPIO28 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
19	TCK (PN)	I/O	1	I	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	GT_PWM03		8	O			

(5) To minimize leakage in some serial flash vendors during LPDS, TI recommends that the user application always enables internal weak pulldown resistors on the FLASH\_SPI\_DIN, FLASH\_SPI\_DOUT, and FLASH\_SPI\_CLK pins.

**Table 4-2. Pin Attributes (continued)**

PIN NO.	SIGNAL NAME <sup>(1)</sup>	SIGNAL TYPE <sup>(2)</sup>	PIN MUX ENCODING	SIGNAL DIRECTION	PAD STATES		
					LPDS <sup>(3)</sup>	Hib <sup>(4)</sup>	nRESET = 0
20	TMS (PN)	I/O	1	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	GPIO29		0	I/O			
21 <sup>(6)</sup>	GPIO25	O	0	O	Hi-Z, Pull, Drive	Driven low	Hi-Z
	GT_PWM02		9	O			
	McAFSX		2	O			
	TCXO_EN		N/A	O			
	SOP2 (PN)		See <sup>(7)</sup>	I			
22	WLAN_XTAL_N	—	N/A	N/A	N/A	N/A	N/A
23	WLAN_XTAL_P	—	N/A	N/A	N/A	N/A	N/A
24	VDD_PLL	—	N/A	N/A	N/A	N/A	N/A
25	LDO_IN2	—	N/A	N/A	N/A	N/A	N/A
26	NC	—	N/A	N/A	N/A	N/A	N/A
27	A_RX	—	N/A	N/A	N/A	N/A	N/A
28	A_TX	—	N/A	N/A	N/A	N/A	N/A
29	GND	—	N/A	N/A	N/A	N/A	N/A
30	GND	—	N/A	N/A	N/A	N/A	N/A
31	RF_BG	—	N/A	N/A	N/A	N/A	N/A
32	nRESET	—	N/A	N/A	N/A	N/A	N/A
33	VDD_PA_IN	—	N/A	N/A	N/A	N/A	N/A
34 <sup>(8)</sup>	SOP1 (PN)	I/O	N/A	N/A	N/A	N/A	N/A
	A_SC1		N/A	N/A	N/A	N/A	N/A
35 <sup>(8)</sup>	SOP0 (PN)	I/O	N/A	N/A	N/A	N/A	N/A
	A_SC2		N/A	N/A	N/A	N/A	N/A
36	LDO_IN1	—	N/A	N/A	N/A	N/A	N/A
37	VIN_DCDC_ANA	—	N/A	N/A	N/A	N/A	N/A
38	DCDC_ANA_SW	—	N/A	N/A	N/A	N/A	N/A
39	VIN_DCDC_PA	—	N/A	N/A	N/A	N/A	N/A
40	DCDC_PA_SW_P	—	N/A	N/A	N/A	N/A	N/A
41	DCDC_PA_SW_N	—	N/A	N/A	N/A	N/A	N/A
42	DCDC_PA_OUT	—	N/A	N/A	N/A	N/A	N/A
43	DCDC_DIG_SW	—	N/A	N/A	N/A	N/A	N/A
44	VIN_DCDC_DIG	—	N/A	N/A	N/A	N/A	N/A
45 <sup>(9)</sup>	GPIO31	I/O	0	I/O	Hi-Z	Hi-Z	Hi-Z
	UART0_RX		9	I			
	McAFSX		12	O			
	UART1_RX		2	I			
	McAXR0		6	I/O			
	GSPI_CLK		7	I/O			
	DCDC_ANA2_SW_P (PN)	—	See <sup>(10)</sup>	N/A	N/A	N/A	N/A

- (6) This pin has dual functions: as a SOP[2] (device operation mode), and as an external TCXO enable. As a TCXO enable, the pin is an output on power up and driven logic high. During hibernate low-power mode, the pin is in a Hi-Z state but is pulled down for SOP mode to disable TCXO. Because of the SOP functionality, the pin must be used as an output only.
- (7) This pin is one of three that must have a passive pullup or pulldown resistor onboard to configure the device hardware power-up mode. For this reason, the pin must be output only when used for digital functions.
- (8) This pin has dual functions: as a SOP (device operation mode) input pin during boot up, and as the 5 GHz switch control (output) pin on power up.
- (9) Pin 45 is used by an internal DC/DC (ANA2\_DCDC). For CC3235S device, pin 45 can be used as GPIO\_31 if a supply is provided on pin 47.
- (10) For details on proper use, see [Section 4.5](#).

Table 4-2. Pin Attributes (continued)

PIN NO.	SIGNAL NAME <sup>(1)</sup>	SIGNAL TYPE <sup>(2)</sup>	PIN MUX ENCODING	SIGNAL DIRECTION	PAD STATES			
					LPDS <sup>(3)</sup>	Hib <sup>(4)</sup>	nRESET = 0	
46	DCDC_ANA2_SW_N	—	N/A	N/A	N/A	N/A	N/A	
47	VDD_ANA2	—	N/A	N/A	N/A	N/A	N/A	
48	VDD_ANA1	—	N/A	N/A	N/A	N/A	N/A	
49	VDD_RAM	—	N/A	N/A	N/A	N/A	N/A	
50	GPIO0 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z	
	UART0_CTS		12	I	Hi-Z, Pull, Drive			
	McAXR1		6	I/O	Hi-Z, Pull, Drive			
	GT_CCP00		7	I	Hi-Z, Pull, Drive			
	GSPI_CS		9	I/O	Hi-Z, Pull, Drive			
	UART1_RTS		10	O	1			
	UART0_RTS		3	O	1			
	McAXR0		4	I/O	Hi-Z, Pull, Drive			
51	RTC_XTAL_P	—	N/A	N/A	N/A	N/A	N/A	
52 <sup>(11)</sup>	RTC_XTAL_N (PN)	O	N/A	N/A	N/A	Hi-Z, Pull, Drive	Hi-Z	
	GPIO32		0	O	Hi-Z, Pull, Drive			
	McACLK		2	O				
	McAXR0		4	O	1			
	UART0_RTS		6	O				
	GSPI_MOSI		8	O				Hi-Z, Pull, Drive
53	GPIO30 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z	
	UART0_TX		9	O	1			
	McACLK		2	O	Hi-Z, Pull, Drive			
	McAFSX		3	O				
	GT_CCP05		4	I				
	GSPI_MISO		7	I/O				
54	VIN_IO2	—	N/A	N/A	N/A	N/A	N/A	
55	GPIO1 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z	
	UART0_TX		3	O	1			
	pCLK (PIXCLK)		4	I	Hi-Z, Pull, Drive			
	UART1_TX		6	O	1			
	GT_CCP01		7	I	Hi-Z, Pull, Drive			
56	VDD_DIG2	—	N/A	N/A	N/A	N/A	N/A	
57 <sup>(12)</sup>	ADC_CH0	Analog input (up to 1.5 V) or digital I/O	See <sup>(10)</sup>	I	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z	
	GPIO2 (PN)		0	I/O				
	UART0_RX		3	I				
	UART1_RX		6	I				
	GT_CCP02		7	I				
58 <sup>(12)</sup>	ADC_CH1	Analog input (up to 1.5 V) or digital I/O	See <sup>(10)</sup>	I	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z	
	GPIO3 (PN)		0	I/O				
	UART1_TX		6	O				1
	pDATA7 (CAM_D3)		4	I				Hi-Z, Pull, Drive

(11) Pin 52 is used by the RTC crystal oscillator. These devices use automatic configuration sensing. Therefore, some board-level configuration is required to use pin 52 as a digital pad. Pin 52 is used for RTC crystal in most applications. However, in some applications a 32.768-kHz square-wave clock might always be available onboard. When a 32.768-kHz square-wave clock is available, the crystal can be removed to free pin 52 for digital functions. The external clock must then be applied at pin 51. For the chip to automatically detect this configuration, a 100-kΩ pullup resistor must be connected between pin 52 and the supply line. To prevent false detection, TI recommends using pin 52 for output-only functions.

(12) This pin is shared by the ADC inputs and digital I/O pad cells.



**Table 4-2. Pin Attributes (continued)**

PIN NO.	SIGNAL NAME <sup>(1)</sup>	SIGNAL TYPE <sup>(2)</sup>	PIN MUX ENCODING	SIGNAL DIRECTION	PAD STATES		
					LPDS <sup>(3)</sup>	Hib <sup>(4)</sup>	nRESET = 0
59 <sup>(12)</sup>	ADC_CH2	Analog input (up to 1.5 V) or digital I/O	See <sup>(10)</sup>	I	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	GPIO4 (PN)		0	I/O			
	UART1_RX		6	I			
	pDATA6 (CAM_D2)		4	I			
60 <sup>(12)</sup>	ADC_CH3	Analog input (up to 1.5 V) or digital I/O	See <sup>(10)</sup>	I	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	GPIO5 (PN)		0	I/O			
	pDATA5 (CAM_D1)		4	I			
	McAXR1		6	I/O			
	GT_CCP05		7	I			
61	GPIO6 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	UART0_RTS		5	O	1		
	pDATA4 (CAM_D0)		4	I	Hi-Z, Pull, Drive		
	UART1_CTS		3	I			
	UART0_CTS		6	I			
	GT_CCP06		7	I			
62	GPIO7 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	McACLKX		13	O	1		
	UART1_RTS		3	O			
	UART0_RTS		10	O			
	UART0_TX		11	O			
63	GPIO8 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	SDCARD_IRQ		6	I			
	McAFSX		7	O			
	GT_CCP06		12	I			
64	GPIO9 (PN)	I/O	0	I/O	Hi-Z, Pull, Drive	Hi-Z, Pull, Drive	Hi-Z
	GT_PWM05		3	O			
	SDCARD_DATA0		6	I/O			
	McAXR0		7	I/O			
	GT_CCP00		12	I			
GND_TAB		—	N/A	N/A	N/A	N/A	N/A

### 4.3 Signal Descriptions

**Table 4-3. Signal Descriptions**

FUNCTION	SIGNAL NAME	PIN NO.	PIN TYPE	SIGNAL DIRECTION	DESCRIPTION
ADC	ADC_CH0	57	I/O	I	ADC channel 0 input (maximum of 1.5 V)
	ADC_CH1	58	I/O	I	ADC channel 1 input (maximum of 1.5 V)
	ADC_CH2	59	I/O	I	ADC channel 2 input (maximum of 1.5 V)
	ADC_CH3	60	I	I	ADC channel 3 input (maximum of 1.5 V)
Antenna selection	GPIO10	1	I/O	O	Antenna selection control
	GPIO11	2	I/O	O	
	GPIO12	3	I/O	O	
	GPIO13	4	I/O	O	
	GPIO14	5	I/O	O	
	GPIO15	6	I/O	O	
	GPIO16	7	I/O	O	
	GPIO17	8	I/O	O	
	GPIO22	15	I/O	O	
	GPIO28	18 <sup>(1)</sup>	I/O	O	
	GPIO25	21	O	O	
	GPIO31	45 <sup>(1)(2)</sup>	I/O	O	
	GPIO0	50	I/O	O	
	GPIO32	52 <sup>(1)</sup>	I/O	O	
	GPIO30	53 <sup>(1)</sup>	I/O	O	
	GPIO3	58	I/O	O	
	GPIO4	59	I/O	O	
	GPIO5	60	I/O	O	
GPIO6	61	I/O	O		
GPIO8	63	I/O	O		
GPIO9	64	I/O	O		

(1) LPDS retention unavailable.

(2) Pin 45 is used by an internal DC/DC (ANA2\_DCDC). For CC3235S device, pin 45 can be used as GPIO\_31 if a supply is provided on pin 47.

**Table 4-3. Signal Descriptions (continued)**

FUNCTION	SIGNAL NAME	PIN NO.	PIN TYPE	SIGNAL DIRECTION	DESCRIPTION
BLE/2.4 GHz radio coexistence	GPIO10	1	I/O	I/O	Coexistence inputs and outputs
	GPIO11	2	I/O	O	
	GPIO12	3	I/O	I/O	
	GPIO13	4	I/O	I/O	
	GPIO14	5	I/O	I/O	
	GPIO15	6	I/O	I/O	
	GPIO16	7	I/O	I/O	
	GPIO17	8	I/O	O	
	GPIO22	15	I/O	I/O	
	GPIO28	18 <sup>(1)</sup>	I/O	I/O	
	GPIO25	21	O	O	
	GPIO31	45 <sup>(1)(2)</sup>	I/O	I/O	
	GPIO0	50	I/O	I/O	
	GPIO32	52 <sup>(1)</sup>	I/O	I/O	
	GPIO30	53 <sup>(1)</sup>	I/O	I/O	
	GPIO3	58	I/O	O	
	GPIO4	59	I/O	O	
	GPIO5	60	I/O	I/O	
GPIO6	61	I/O	I/O		
GPIO8	63	I/O	I/O		
GPIO9	64	I/O	I/O		
Clock	WLAN_XTAL_N	22	—	—	40-MHz crystal; pull down if external TCXO is used
	WLAN_XTAL_P	23	—	—	40-MHz crystal or TCXO clock input
	RTC_XTAL_P	51	—	—	Connect 32.768-kHz crystal or force external CMOS level clock
	RTC_XTAL_N	52	—	—	Connect 32.768-kHz crystal or connect 100-kΩ resistor to supply voltage

**Table 4-3. Signal Descriptions (continued)**

FUNCTION	SIGNAL NAME	PIN NO.	PIN TYPE	SIGNAL DIRECTION	DESCRIPTION
Hostless mode	HM_IO	1	I/O	I/O	Hostless mode inputs and outputs
		2	I/O	O	
		3	I/O	I/O	
		4	I/O	I/O	
		5	I/O	I/O	
		6	I/O	I/O	
		7	I/O	I/O	
		8	I/O	O	
		15	I/O	I/O	
		18 <sup>(1)</sup>	I/O	I/O	
		21	O	O	
		45 <sup>(1)(2)</sup>	I/O	I/O	
		50	I/O	I/O	
		52 <sup>(1)</sup>	I/O	I/O	
		53 <sup>(1)</sup>	I/O	I/O	
		58	O	O	
		59	O	O	
	60	I/O	I/O		
	61	I/O	I/O		
	63	I/O	I/O		
	64	I/O	I/O		
JTAG / SWD	TDI	16	I/O	I	JTAG TDI. Reset default pinout.
	TDO	17	I/O	O	JTAG TDO. Reset default pinout.
	TCK	19	I/O	I	JTAG/SWD TCK. Reset default pinout.
	TMS	20	I/O	I/O	JTAG/SWD TMS. Reset default pinout.
I <sup>2</sup> C	I2C_SCL	1	I/O	I/O (open drain)	I <sup>2</sup> C clock data
		3			
		5			
		16			
	I2C_SDA	2	I/O	I/O (open drain)	I <sup>2</sup> C data
		4			
		6			
		17			

**Table 4-3. Signal Descriptions (continued)**

FUNCTION	SIGNAL NAME	PIN NO.	PIN TYPE	SIGNAL DIRECTION	DESCRIPTION
Timers	GT_PWM06	1	I/O	O	Pulse-width modulated O/P
	GT_CCP01	1	I/O	I	Timer capture port
	GT_PWM07	2	I/O	O	Pulse-width modulated O/P
	GT_CCP02	2	I/O	I	Timer capture ports
	GT_CCP03	3	I/O	I	
	GT_CCP04	4	I/O	I	
		15	I/O	I	
	GT_CCP05	5	I/O	I	
	GT_CCP06	6	I/O	I	
		17	I/O	I	
		61	I/O	I	
		63	I/O	I	
	GT_CCP07	7	I/O	I	
	PWM0	17	I/O	O	
	GT_PWM03	19	I/O	O	
	GT_PWM02	21	O	O	
	GT_CCP00	50	I/O	I	
		64	I/O	I	
	GT_CCP05	53	I/O	I	Timer capture ports
	GT_CCP01	55	I/O	I	
GT_CCP02	57	I/O	I		
GT_CCP05	60	I	I	Timer capture port Input	
GT_PWM05	64	I/O	O	Pulse-width modulated output	

**Table 4-3. Signal Descriptions (continued)**

FUNCTION	SIGNAL NAME	PIN NO.	PIN TYPE	SIGNAL DIRECTION	DESCRIPTION
GPIO	GPIO10	1	I/O	I/O	General-purpose inputs or outputs
	GPIO11	2	I/O	I/O	
	GPIO12	3	I/O	I/O	
	GPIO13	4	I/O	I/O	
	GPIO14	5	I/O	I/O	
	GPIO15	6	I/O	I/O	
	GPIO16	7	I/O	I/O	
	GPIO17	8	I/O	I/O	
	GPIO22	15	I/O	I/O	
	GPIO23	16	I/O	I/O	
	GPIO24	17	I/O	I/O	
	GPIO28	18	I/O	I/O	
	GPIO29	20	I/O	I/O	
	GPIO25	21	O	O	
	GPIO31	45 <sup>(2)</sup>	I/O	I/O	
	GPIO0	50	I/O	I/O	
	GPIO32	52	I/O	O	
	GPIO30	53	I/O	I/O	
	GPIO1	55	I/O	I/O	
	GPIO2	57	I/O	I/O	
	GPIO3	58	I/O	I/O	
GPIO4	59	I/O	I/O		
GPIO5	60	I/O	I/O		
GPIO6	61	I/O	I/O		
GPIO7	62	I/O	I/O		
GPIO8	63	I/O	I/O		
GPIO9	64	I/O	I/O		
McASP I <sup>2</sup> S or PCM	MCAFSX	2	I/O	O	I <sup>2</sup> S audio port frame sync
		15			
		17			
		21			
		45 <sup>(2)</sup>			
		53			
		63			
	McACLK	3	I/O	O	I <sup>2</sup> S audio port clock outputs
		52	O	O	
		53	I/O	O	
	McAXR1	50	I/O	I/O	I <sup>2</sup> S audio port data 1 (RX/TX)
		60	I	I/O	I <sup>2</sup> S audio port data 1 (RX and TX)
	McAXR0	45 <sup>(2)</sup>	I/O	I/O	I <sup>2</sup> S audio port data 0 (RX and TX)
		50	I/O	I/O	
52		O	O	I <sup>2</sup> S audio port data (only output mode is supported on pin 52)	
64		I/O	I/O	I <sup>2</sup> S audio port data (RX and TX)	
McACLKX	62	I/O	O	I <sup>2</sup> S audio port clock	

**Table 4-3. Signal Descriptions (continued)**

FUNCTION	SIGNAL NAME	PIN NO.	PIN TYPE	SIGNAL DIRECTION	DESCRIPTION
Multimedia card (MMC or SD)	SDCARD_CLK	1	I/O	O	SD card clock data
		7			
	SDCARD_CMD	2	I/O	I/O (open drain)	SD card command line
		8	I/O	I/O	
	SDCARD_DATA0	6	I/O	I/O	SD card data
64					
	SDCARD_IRQ	63	I/O	I	Interrupt from SD card <sup>(3)</sup>
Parallel interface (8-bit $\pi$ )	pXCLK (XVCLK)	2	I/O	O	Free clock to parallel camera
	pVS (VSYNC)	3	I/O	I	Parallel camera vertical sync
	pHS (HSYNC)	4	I/O	I	Parallel camera horizontal sync
	pDATA8 (CAM_D4)	5	I/O	I	Parallel camera data bit 4
	pDATA9 (CAM_D5)	6	I/O	I	Parallel camera data bit 5
	pDATA10 (CAM_D6)	7	I/O	I	Parallel camera data bit 6
	pDATA11 (CAM_D7)	8	I/O	I	Parallel camera data bit 7
	pCLK (PIXCLK)	55	I/O	I	Pixel clock from parallel camera sensor
	pDATA7 (CAM_D3)	58	I/O	I	Parallel camera data bit 3
	pDATA6 (CAM_D2)	59	I/O	I	Parallel camera data bit 2
	pDATA5 (CAM_D1)	60	I	I	Parallel camera data bit 1
	pDATA4 (CAM_D0)	61	I/O	I	Parallel camera data bit 0
Power	VDD_DIG1	9	—	—	Internal digital core voltage
	VIN_IO1	10	—	—	Device supply voltage ( $V_{BAT}$ )
	VDD_PLL	24	—	—	Internal analog voltage
	LDO_IN2	25	—	—	Internal analog RF supply from analog DC/DC output
	VDD_PA_IN	33	—	—	Internal PA supply voltage from PA DC/DC output
	LDO_IN1	36	—	—	Internal analog RF supply from analog DC/DC output
	VIN_DCDC_ANA	37	—	—	Analog DC/DC input (connected to device input supply [ $V_{BAT}$ ])
	DCDC_ANA_SW	38	—	—	Internal analog DC/DC switching node
	VIN_DCDC_PA	39	—	—	PA DC/DC input (connected to device input supply [ $V_{BAT}$ ])
	DCDC_PA_SW_P	40	—	—	Internal PA DC/DC switching node
	DCDC_PA_SW_N	41	—	—	Internal PA DC/DC switching node
	DCDC_PA_OUT	42	—	—	Internal PA buck converter output
	DCDC_DIG_SW	43	—	—	Internal digital DC/DC switching node
	VIN_DCDC_DIG	44	—	—	Digital DC/DC input (connected to device input supply [ $V_{BAT}$ ])
	DCDC_ANA2_SW_P	45 <sup>(2)</sup>	—	—	Analog to DC/DC converter +ve switching node
	DCDC_ANA2_SW_N	46	—	—	Internal analog to DC/DC converter –ve switching node
	VDD_ANA2	47	—	—	Internal analog to DC/DC output
	VDD_ANA1	48	—	—	Internal analog supply fed by ANA2 DC/DC output
VDD_RAM	49	—	—	Internal SRAM LDO output	
VIN_IO2	54	—	—	Device supply voltage ( $V_{BAT}$ )	
VDD_DIG2	56	—	—	Internal digital core voltage	
Reset	nRESET	32	I	I	Global master device reset (active low)
RF	A_RX	27	I	I	WLAN analog A-band receive
	A_TX	28	O	O	WLAN analog A-band transmit
	RF_BG	31	I/O	I/O	WLAN analog RF 802.11 b/g bands

(3) Future support.

Table 4-3. Signal Descriptions (continued)

FUNCTION	SIGNAL NAME	PIN NO.	PIN TYPE	SIGNAL DIRECTION	DESCRIPTION
SPI	GSPI_CLK	5	I/O	I/O	General SPI clock
		45 <sup>(2)</sup>	I/O	I/O	
	GSPI_MISO	6	I/O	I/O	General SPI MISO
		53	I/O	I/O	
	GSPI_CS	8	I/O	I/O	General SPI device select
		50	I/O	I/O	
GSPI_MOSI	7	I/O	I/O	General SPI MOSI	
	52	O	O		
FLASH SPI	FLASH_SPI_CLK	11	O	O	Clock to SPI serial flash (fixed default)
	FLASH_SPI_DOUT	12	O	O	Data to SPI serial flash (fixed default)
	FLASH_SPI_DIN	13	I	I	Data from SPI serial flash (fixed default)
	FLASH_SPI_CS	14	O	O	Device select to SPI serial flash (fixed default)
UART	UART1_TX	1	I/O	O	UART TX data
		7	I/O	O	
		16	I/O	O	
		55	I/O	O	
		58	I/O	O	
	UART1_RX	2	I/O	I	UART RX data
		8	I/O	I	
		17	I/O	I	
		45 <sup>(2)</sup>	I/O	I	
		57	I/O	I	
		59	I/O	I	
	UART1_RTS	50	I/O	O	UART1 request-to-send (active low)
		62	I/O	O	
	UART1_CTS	61	I/O	I	UART1 clear-to-send (active low)
	UART0_TX	3	I/O	O	UART0 TX data
		53	I/O	O	
		55	I/O	O	
		62	I/O	O	
	UART0_RX	4	I/O	I	UART0 RX data
		45 <sup>(2)</sup>	I/O	I	UART0 RX data
		57	I/O	I	UART0 RX data
	UART0_CTS	50	I/O	I	UART0 clear-to-send input (active low)
		61			
	UART0_RTS	50	I/O	O	UART0 request-to-send (active low)
52					
61					
62					
Sense-On-Power	SOP2	21 <sup>(4)</sup>	O	I	Sense-on-power 2
	SOP1	34	I	I	Configuration sense-on-power 1
	SOP0	35	I	I	Configuration sense-on-power 0

(4) This pin has dual functions: as a SOP[2] (device operation mode), and as an external TCXO enable. As a TCXO enable, the pin is an output on power up and driven logic high. During hibernate low-power mode, the pin is in a Hi-Z state but is pulled down for SOP mode to disable TCXO. Because of the SOP functionality, the pin must be used as an output only.



## 4.4 Pin Multiplexing

**Table 4-4. Pin Multiplexing**

Register Address	Register Name	Pin	ANALOG OR SPECIAL FUNCTION				Digital Function (XXX Field Encoding) <sup>(1)</sup>													
			JTAG	Hostless Mode	BLE COEX		0	1	2	3	4	5	6	7	8	9	10	11	12	13
					CC_COEX_SW_OUT	CC_COEX_BLE_IN														
0x4402E0C8	GPIO_PAD_CONFIG_10	1	—	Y	Y	Y	GPIO10	I2C_SCL	—	GT_PWM06	—	—	SDCARD_CLK	UART1_TX	—	—	—	—	GT_CCP01	—
0x4402E0CC	GPIO_PAD_CONFIG_11	2	—	Y <sup>(2)</sup>	Y	—	GPIO11	I2C_SDA	—	GT_PWM07	pXCLK (XVCLK)	—	SDCARD_CMD	UART1_RX	—	—	—	—	GT_CCP02	MCAFSX
0x4402E0D0	GPIO_PAD_CONFIG_12	3	—	Y	Y	Y	GPIO12	—	—	McACLK	pVS (VSYNC)	I2C_SCL	—	UART0_TX	—	—	—	—	GT_CCP03	—
0x4402E0D4	GPIO_PAD_CONFIG_13	4	—	Y	Y	Y	GPIO13	—	—	—	pHS (HSYNC)	I2C_SDA	—	UART0_RX	—	—	—	—	GT_CCP04	—
0x4402E0D8	GPIO_PAD_CONFIG_14	5	—	Y	Y	Y	GPIO14	—	—	—	pDATA8 (CAM_D4)	I2C_SCL	—	GSPI_CLK	—	—	—	—	GT_CCP05	—
0x4402E0DC	GPIO_PAD_CONFIG_15	6	—	Y	Y	Y	GPIO15	—	—	—	pDATA9 (CAM_D5)	I2C_SDA	—	GSPI_MISO	SDCARD_DATA0	—	—	—	—	GT_CCP06
0x4402E0E0	GPIO_PAD_CONFIG_16	7	—	Y	Y	Y	GPIO16	—	—	—	pDATA10 (CAM_D6)	UART1_TX	—	GSPI_MOSI	SDCARD_CLK	—	—	—	—	GT_CCP07
0x4402E0E4	GPIO_PAD_CONFIG_17	8	—	Y <sup>(2)</sup>	Y	—	GPIO17	—	—	—	pDATA11 (CAM_D7)	UART1_RX	—	GSPI_CS	SDCARD_CMD	—	—	—	—	—
0x4402E0F8	GPIO_PAD_CONFIG_22	15	—	Y	Y	Y	GPIO22	—	—	—	—	GT_CCP04	—	McAFSX	—	—	—	—	—	—
0x4402E0FC	GPIO_PAD_CONFIG_23	16	Muxed with JTAG	—	—	—	GPIO23	TDI	UART1_TX	—	—	—	—	—	—	I2C_SCL	—	—	—	—
0x4402E100	GPIO_PAD_CONFIG_24	17	Muxed with JTAG TDO	—	—	—	GPIO24	TDO	UART1_RX	—	GT_CCP06	PWM0	McAFSX	—	—	I2C_SDA	—	—	—	—
0x4402E140	GPIO_PAD_CONFIG_40	18	—	Y <sup>(3)</sup>	Y <sup>(3)</sup>	Y <sup>(3)</sup>	GPIO28	—	—	—	—	—	—	—	—	—	—	—	—	—
0x4402E110	GPIO_PAD_CONFIG_28	19	Muxed with JTAG or SWD and TCK	—	—	—	—	TCK	—	—	—	—	—	—	GT_PWM03	—	—	—	—	—
0x4402E114	GPIO_PAD_CONFIG_29	20	Muxed with JTAG or SWD and TMS	—	—	—	GPIO29	TMS	—	—	—	—	—	—	—	—	—	—	—	—

(1) Pin mux encodings with (RD) denote the default encoding after reset release.

(2) Output Only

(3) LPDS retention unavailable.

**Table 4-4. Pin Multiplexing (continued)**

Register Address	Register Name	Pin	ANALOG OR SPECIAL FUNCTION				Digital Function (XXX Field Encoding) <sup>(1)</sup>														
			JTAG	Hostless Mode	BLE COEX		0	1	2	3	4	5	6	7	8	9	10	11	12	13	
					CC_COEX_SW_OUT	CC_COEX_BLE_IN															
0x4402E104	GPIO_PAD_CONFIG_25	21 <sup>(4)</sup>	—	Y <sup>(2)</sup>	Y	—	GPIO25	—	McAFSX	—	—	—	—	—	—	GT_PWM02	—	—	—	—	
0x4402E11C	GPIO_PAD_CONFIG_31	45 <sup>(3)(5)</sup>	—	Y	Y	Y	GPIO31	—	UART1_RX	—	—	—	—	McAXR0	GSPL_CLK	—	UART0_RX	—	—	McAFSX	—
0x4402E0A0	GPIO_PAD_CONFIG_0	50	—	Y	Y	Y	GPIO0	—	—	UART0_RTS	McAXR0	—	McAXR1	GT_CCP00	—	GSPL_CS	UART1_RTS	—	UART0_CTS	—	
0x4402E120	GPIO_PAD_CONFIG_32	52	—	Y <sup>(3)</sup>	Y <sup>(3)</sup>	Y <sup>(3)</sup>	GPIO32	—	McACLK	—	McAXR0	—	UART0_RTS	—	GSPL_MOSI	—	—	—	—	—	
0x4402E118	GPIO_PAD_CONFIG_30	53	—	Y <sup>(3)</sup>	Y <sup>(3)</sup>	Y <sup>(3)</sup>	GPIO30	—	McACLK	McAFSX	GT_CCP05	—	—	GSPL_MISO	—	UART0_TX	—	—	—	—	
0x4402E0A4	GPIO_PAD_CONFIG_1	55	—	—	—	—	GPIO1	—	—	UART0_TX	pCLK (PIXCLK)	—	UART1_TX	GT_CCP01	—	—	—	—	—	—	
0x4402E0A8	GPIO_PAD_CONFIG_2	57	—	—	—	—	GPIO2	—	—	UART0_RX	—	—	UART1_RX	GT_CCP02	—	—	—	—	—	—	
0x4402E0AC	GPIO_PAD_CONFIG_3	58	—	Y <sup>(2)</sup>	Y	—	GPIO3	—	—	—	pDATA7 (CAM_D3)	—	UART1_TX	—	—	—	—	—	—	—	
0x4402E0B0	GPIO_PAD_CONFIG_4	59	—	Y <sup>(2)</sup>	Y	—	GPIO4	—	—	—	pDATA6 (CAM_D2)	—	UART1_RX	—	—	—	—	—	—	—	
0x4402E0B4	GPIO_PAD_CONFIG_5	60	—	Y	Y	Y	GPIO5	—	—	—	pDATA5 (CAM_D1)	—	McAXR1	GT_CCP05	—	—	—	—	—	—	
0x4402E0B8	GPIO_PAD_CONFIG_6	61	—	Y	Y	Y	GPIO6	—	—	UART1_CTS	pDATA4 (CAM_D0)	UART0_RTS	UART0_CTS	GT_CCP06	—	—	—	—	—	—	
0x4402E0BC	GPIO_PAD_CONFIG_7	62	—	—	—	—	GPIO7	—	—	UART1_RTS	—	—	—	—	—	—	UART0_RTS	UART0_TX	—	McACLKX	
0x4402E0C0	GPIO_PAD_CONFIG_8	63	—	Y	Y	Y	GPIO8	—	—	—	—	—	SDCARD_IRQ	McAFSX	—	—	—	—	—	GT_CCP06	
0x4402E0C4	GPIO_PAD_CONFIG_9	64	—	Y	Y	Y	GPIO9	—	—	GT_PWM05	—	—	SDCARD_DATA0	McAXR0	—	—	—	—	—	GT_CCP00	

- (4) This pin has dual functions: as a SOP[2] (device operation mode), and as an external TCXO enable. As a TXCO enable, the pin is an output on power up and driven logic high. During hibernate low-power mode, the pin is in a Hi-Z state but is pulled down for SOP mode to disable TCXO. Because of the SOP functionality, the pin must be used as an output only.
- (5) Pin 45 is used by an internal DC/DC (ANA2\_DCDC). For CC3235S device, pin 45 can be used as GPIO\_31 if a supply is provided on pin 47.

## 4.5 Drive Strength and Reset States for Analog and Digital Multiplexed Pins

Table 4-5 describes the use, drive strength, and default state of analog and digital multiplexed pins at first-time power up and reset (nRESET pulled low).

**Table 4-5. Drive Strength and Reset States for Analog and Digital Multiplexed Pins**

Pin	Board-Level Configuration and Use	Default State at First Power Up or Forced Reset	State After Configuration of Analog Switches (ACTIVE, LPDS, and HIB Power Modes)	Maximum Effective Drive Strength (mA)
45	VDD_ANA2 (pin 47) must be shorted to the input supply rail. Otherwise, the pin is driven by the ANA2 DC/DC.	Analog is isolated. The digital I/O cell is also isolated.	Determined by the I/O state, as are other digital I/Os.	4
50	Generic I/O	Analog is isolated. The digital I/O cell is also isolated.	Determined by the I/O state, as are other digital I/Os.	4
52	The pin must have an external pullup of 100 kΩ to the supply rail and must be used in output signals only.	Analog is isolated. The digital I/O cell is also isolated.	Determined by the I/O state, as are other digital I/Os.	4
53	Generic I/O	Analog is isolated. The digital I/O cell is also isolated.	Determined by the I/O state, as are other digital I/Os.	4
57	Analog signal (1.8-V absolute, 1.46-V full scale)	ADC is isolated. The digital I/O cell is also isolated.	Determined by the I/O state, as are other digital I/Os.	4
58	Analog signal (1.8-V absolute, 1.46-V full scale)	ADC is isolated. The digital I/O cell is also isolated.	Determined by the I/O state, as are other digital I/Os.	4
59	Analog signal (1.8-V absolute, 1.46-V full scale)	ADC is isolated. The digital I/O cell is also isolated.	Determined by the I/O state, as are other digital I/Os.	4
60	Analog signal (1.8-V absolute, 1.46-V full scale)	ADC is isolated. The digital I/O cell is also isolated.	Determined by the I/O state, as are other digital I/Os.	4

## 4.6 Pad State After Application of Power to Device, Before Reset Release

When a stable power is applied to the CC3235x device for the first time or when supply voltage is restored to the proper value following a period with supply voltage less than 1.5 V, the level of each digital pad is undefined in the period starting from the release of nRESET and until DIG\_DCDC powers up. This period is less than approximately 10 ms. During this period, pads can be internally pulled weakly in either direction. If a certain set of pins is required to have a definite value during this pre-reset period, an appropriate pullup or pulldown resistor must be used at the board level. The recommended value of this external pull is 2.7 kΩ.

## 4.7 Connections for Unused Pins

All unused pin should be configured as stated in [Table 4-6](#).

**Table 4-6. Connections for Unused Pins**

FUNCTION	SIGNAL DESCRIPTION	PIN NUMBER	ACCEPTABLE PRACTICE	PREFERRED PRACTICE
GPIO	General-purpose input or output		Wake up I/O source should not be floating during hibernate. All the I/O pins will float while in Hibernate and Reset states. Ensure pullup and pulldown resistors are available on board to maintain the state of the I/O. Leave unused GPIOs as NC	
No Connect	NC	26	Unused pin, leave as NC.	Unused pin, leave as NC
SOP	Configuration sense-on-power		Ensure pulldown resistors are available on unused SOP pins	69.8K Pull down resistor on SOP0 and SOP1 used as switch control pins, 100K pull down on SOP2
Reset	RESET input for the device		Never leave the reset pin floating	
Clock	RTC_XTAL_N		When using an external oscillator, add a 100-k $\Omega$ pullup resistor to VIO	
	WLAN_XTAL_N		When using an external oscillator, connect to ground if unused	
JTAG	JTAG interface		Leave as NC if unused	

## 5 Specifications

All measurements are referenced at the device pins, unless otherwise indicated. All specifications are over process and voltage, unless otherwise indicated.

### 5.1 Absolute Maximum Ratings

All measurements are referenced at the device pins unless otherwise indicated. All specifications are over process and overvoltage unless otherwise indicated.

Over operating free-air temperature range (unless otherwise noted)<sup>(1)(2)</sup>

			MIN	MAX	UNIT
Supply voltage	$V_{BAT}$ and $V_{IO}$	Pins: 37, 39, 44	-0.5	3.8	V
	$V_{IO} - V_{BAT}$ (differential)	Pins: 10, 54	$V_{BAT}$ and $V_{IO}$ should be tied together		V
Digital inputs			-0.5	$V_{IO} + 0.5$	V
RF pins			-0.5	2.1	V
Analog pins, Crystal		Pins: 22, 23, 51, 52	-0.5	2.1	V
Operating temperature, $T_A$			-40	85	°C
Storage temperature, $T_{stg}$			-55	125	°C

- Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- All voltage values are with respect to  $V_{SS}$ , unless otherwise noted.

### 5.2 ESD Ratings

			VALUE	UNIT
$V_{ESD}$	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±2000	V
		Charged device model (CDM), per ANSI/ESDA/JEDEC JS-002 <sup>(2)</sup>	±500	

- JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 5.3 Power-On Hours (POH)

This information is provided solely for your convenience and does not extend or modify the warranty provided under TI's standard terms and conditions for TI semiconductor products.

OPERATING CONDITION	POWER-ON HOURS [POH] (hours)
$T_A$ up to 85°C <sup>(1)</sup>	87,600

- The TX duty cycle (power amplifier ON time) is assumed to be 10% of the device POH. Of the remaining 90% of the time, the device can be in any other state.

### 5.4 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)<sup>(1)(2)</sup>

				MIN	TYP	MAX	UNIT
Supply voltage	$V_{BAT}$ , $V_{IO}$ (shorted to $V_{BAT}$ )	Pins: 10, 37, 39, 44, 54	Direct battery connection <sup>(3)</sup>	2.1 <sup>(4)</sup>	3.3	3.6	V
Ambient thermal slew				-20		20	°C/minute

- Operating temperature is limited by crystal frequency variation.
- When operating at an ambient temperature of over 75°C, the transmit duty cycle must remain below 50% to avoid the auto-protect feature of the power amplifier. If the auto-protect feature triggers, the device takes a maximum of 60 seconds to restart the transmission.
- To ensure WLAN performance, ripple on the supply must be less than ±300 mV.
- The minimum voltage specified includes the ripple on the supply voltage and all other transient dips. The brownout condition is also 2.1 V, and care must be taken when operating at the minimum specified voltage.

## 5.5 Current Consumption Summary (CC3235S)

**Table 5-1. Current Consumption Summary (CC3235S) 2.4 GHz RF Band**
 $T_A = 25^\circ\text{C}$ ,  $V_{\text{BAT}} = 3.6\text{ V}$ 

PARAMETER		TEST CONDITIONS <sup>(1) (2)</sup>		MIN	TYP <sup>(3)</sup>	MAX	UNIT
MCU ACTIVE	NWP ACTIVE	TX	1 DSSS	TX power level = 0	272		mA
				TX power level = 4	190		
			6 OFDM	TX power level = 0	248		
				TX power level = 4	182		
			54 OFDM	TX power level = 0	223		
				TX power level = 4	160		
		RX	1 DSSS	59			
			54 OFDM	59			
NWP idle connected <sup>(4)</sup>				15.3			
MCU SLEEP	NWP ACTIVE	TX	1 DSSS	TX power level = 0	269		mA
				TX power level = 4	187		
			6 OFDM	TX power level = 0	245		
				TX power level = 4	179		
			54 OFDM	TX power level = 0	220		
				TX power level = 4	157		
		RX	1 DSSS	56			
			54 OFDM	56			
NWP idle connected <sup>(4)</sup>				12.2			
MCU LPDS	NWP ACTIVE	TX	1 DSSS	TX power level = 0	266		mA
				TX power level = 4	184		
			6 OFDM	TX power level = 0	242		
				TX power level = 4	176		
			54 OFDM	TX power level = 0	217		
				TX power level = 4	154		
		RX	1 DSSS	53			
			54 OFDM	53			
NWP LPDS <sup>(5)</sup>		120 $\mu\text{A}$ at 64KB 135 $\mu\text{A}$ at 256KB		135		$\mu\text{A}$	
NWP idle connected <sup>(4)</sup>				710			
MCU SHUTDOWN	MCU shutdown			1		$\mu\text{A}$	
MCU HIBERNATE	MCU hibernate			4.5		$\mu\text{A}$	
Peak calibration current <sup>(6)</sup>		$V_{\text{BAT}} = 3.6\text{ V}$		420		mA	
		$V_{\text{BAT}} = 3.3\text{ V}$		450			
		$V_{\text{BAT}} = 2.1\text{ V}$		670			

- (1) TX power level = 0 implies maximum power (see [Figure 5-1](#), [Figure 5-2](#), and [Figure 5-3](#)). TX power level = 4 implies output power backed off approximately 4 dB.
- (2) The CC3235x system is a constant power-source system. The active current numbers scale based on the  $V_{\text{BAT}}$  voltage supplied.
- (3) Typical numbers assume a VSWR of 1.5:1.
- (4) DTIM = 1
- (5) LPDS current does not include the external serial flash. The LPDS number of reported is with retention of 256KB of MCU SRAM. The CC3235x device can be configured to retain 0KB, 64KB, 128KB, 192KB, or 256KB of SRAM in LPDS. Each 64-KB block of MCU retained SRAM increases LPDS current by 4  $\mu\text{A}$ .
- (6) The complete calibration can take up to 17 mJ of energy from the battery over a time of 24 ms. In default mode, calibration is performed sparingly, and typically occurs when re-enabling the NWP and when the temperature has changed by more than 20°C. There are two additional calibration modes that may be used to reduced or completely eliminate the calibration event. For further details, see [CC31xx, CC32xx SimpleLink™ Wi-Fi® and IoT Network Processor Programmer's Guide](#).

**Table 5-2. Current Consumption Summary (CC3235S) 5 GHz RF Band**
 $T_A = 25^\circ\text{C}$ ,  $V_{\text{BAT}} = 3.6\text{ V}$ 

PARAMETER		TEST CONDITIONS <sup>(1) (2)</sup>		MIN	TYP <sup>(3)</sup>	MAX	UNIT
MCU ACTIVE	NWP ACTIVE	TX	6 OFDM		318		mA
			54 OFDM		293		
	RX	54 OFDM		67			
	NWP idle connected <sup>(4)</sup>				15.3		
MCU SLEEP	NWP ACTIVE	TX	6 OFDM		315		mA
			54 OFDM		290		
	RX	54 OFDM		64			
	NWP idle connected <sup>(4)</sup>				12.2		
MCU LPDS	NWP ACTIVE	TX	6 OFDM		312		mA
			54 OFDM		287		
		RX	54 OFDM		61		
	NWP LPDS <sup>(5)</sup>		120 $\mu\text{A}$ at 64KB 135 $\mu\text{A}$ at 256KB		135		$\mu\text{A}$
	NWP idle connected <sup>(6)</sup>				710		
MCU SHUTDOWN	MCU shutdown				1		$\mu\text{A}$
MCU HIBERNATE	MCU hibernate				4.5		$\mu\text{A}$
Peak calibration current <sup>(7)</sup>	$V_{\text{BAT}} = 3.6\text{ V}$				290		mA
	$V_{\text{BAT}} = 3.3\text{ V}$				310		
	$V_{\text{BAT}} = 2.7\text{ V}$				310		
	$V_{\text{BAT}} = 2.1\text{ V}$				400		

(1) Measurements taken at maximum TX power

(2) The CC3235x system is a constant power-source system. The active current numbers scale based on the  $V_{\text{BAT}}$  voltage supplied.

(3) Typical numbers assume a VSWR of 1.5:1.

(4) DTIM = 1

(5) LPDS current does not include the external serial flash. The LPDS number of reported is with retention of 256KB of MCU SRAM. The CC3235x device can be configured to retain 0KB, 64KB, 128KB, 192KB, or 256KB of SRAM in LPDS. Each 64-KB block of MCU retained SRAM increases LPDS current by 4  $\mu\text{A}$ .

(6) DTIM = 1

(7) The complete calibration can take up to 17 mJ of energy from the battery over a time of 24 ms. In default mode, calibration is performed sparingly, and typically occurs when re-enabling the NWP and when the temperature has changed by more than 20°C. There are two additional calibration modes that may be used to reduced or completely eliminate the calibration event. For further details, see [CC31xx, CC32xx SimpleLink™ Wi-Fi® and IoT Network Processor Programmer's Guide](#).

## 5.6 Current Consumption Summary (CC3235SF)

**Table 5-3. Current Consumption Summary (CC3235SF) 2.4 GHz RF Band**
 $T_A = 25^\circ\text{C}$ ,  $V_{\text{BAT}} = 3.6\text{ V}$ 

PARAMETER		TEST CONDITIONS <sup>(1) (2)</sup>		MIN	TYP <sup>(3)</sup>	MAX	UNIT
MCU ACTIVE	NWP ACTIVE	TX	1 DSSS	TX power level = 0		286	mA
				TX power level = 4		202	
			6 OFDM	TX power level = 0		255	
				TX power level = 4		192	
			54 OFDM	TX power level = 0		232	
				TX power level = 4		174	
		RX	1 DSSS		74		
54 OFDM			74				
NWP idle connected <sup>(4)</sup>					25.2		
MCU SLEEP	NWP ACTIVE	TX	1 DSSS	TX power level = 0		282	mA
				TX power level = 4		198	
			6 OFDM	TX power level = 0		251	
				TX power level = 4		188	
			54 OFDM	TX power level = 0		228	
				TX power level = 4		170	
		RX	1 DSSS		70		
54 OFDM			70				
NWP idle connected <sup>(4)</sup>					21.2		
MCU LPDS	NWP active	TX	1 DSSS	TX power level = 0		266	mA
				TX power level = 4		184	
			6 OFDM	TX power level = 0		242	
				TX power level = 4		176	
			54 OFDM	TX power level = 0		217	
				TX power level = 4		154	
		RX	1 DSSS		53		
54 OFDM			53				
NWP LPDS <sup>(5)</sup>		120 $\mu\text{A}$ at 64KB 135 $\mu\text{A}$ at 256KB		135	$\mu\text{A}$		
NWP idle connected <sup>(4)</sup>					710		
MCU SHUTDOWN	MCU shutdown				1	$\mu\text{A}$	
MCU HIBERNATE	MCU hibernate				4.5	$\mu\text{A}$	
Peak calibration current <sup>(6)</sup>	$V_{\text{BAT}} = 3.6\text{ V}$				420	mA	
	$V_{\text{BAT}} = 3.3\text{ V}$				450		
	$V_{\text{BAT}} = 2.1\text{ V}$				670		

- (1) TX power level = 0 implies maximum power (see [Figure 5-1](#), [Figure 5-2](#), and [Figure 5-3](#)). TX power level = 4 implies output power backed off approximately 4 dB.
- (2) The CC3235x system is a constant power-source system. The active current numbers scale based on the  $V_{\text{BAT}}$  voltage supplied.
- (3) Typical numbers assume a VSWR of 1.5:1.
- (4) DTIM = 1
- (5) LPDS current does not include the external serial flash. The LPDS number of reported is with retention of 256KB of MCU SRAM. The CC3235x device can be configured to retain 0KB, 64KB, 128KB, 192KB, or 256KB of SRAM in LPDS. Each 64-KB block of MCU retained SRAM increases LPDS current by 4  $\mu\text{A}$ .
- (6) The complete calibration can take up to 17 mJ of energy from the battery over a period of 24 ms. Calibration is performed sparingly, typically when coming out of HIBERNATE and only if temperature has changed by more than 20°C. The calibration event can be controlled by a configuration file in the serial flash.



**Table 5-4. Current Consumption Summary (CC3235SF) 5 GHz RF Band**

T<sub>A</sub> = 25°C, V<sub>BAT</sub> = 3.6 V

PARAMETER		TEST CONDITIONS <sup>(1) (2)</sup>		MIN	TYP <sup>(3)</sup>	MAX	UNIT	
MCU ACTIVE	NWP ACTIVE	TX	6 OFDM		329		mA	
			54 OFDM		306			
	RX	54 OFDM		80				
	NWP idle connected <sup>(4)</sup>			25.2				
MCU SLEEP	NWP ACTIVE	TX	6 OFDM		325		mA	
			54 OFDM		302			
	RX	54 OFDM		76				
	NWP idle connected <sup>(4)</sup>			21.2				
MCU LPDS	NWP active	TX	6 OFDM		312		mA	
			54 OFDM		289			
	RX	54 OFDM		63				
	NWP LPDS <sup>(5)</sup>			120 µA at 64KB 135 µA at 256KB		135		µA
	NWP idle connected <sup>(4)</sup>					710		
MCU SHUTDOWN	MCU shutdown				1		µA	
MCU HIBERNATE	MCU hibernate				4.5		µA	
Peak calibration current <sup>(6)</sup>	V <sub>BAT</sub> = 3.6 V				290		mA	
	V <sub>BAT</sub> = 3.3 V				310			
	V <sub>BAT</sub> = 2.7 V				310			
	V <sub>BAT</sub> = 2.1 V				400			

- (1) Measurements taken at maximum TX power
- (2) The CC3235x system is a constant power-source system. The active current numbers scale based on the V<sub>BAT</sub> voltage supplied.
- (3) Typical numbers assume a VSWR of 1.5:1.
- (4) DTIM = 1
- (5) LPDS current does not include the external serial flash. The LPDS number of reported is with retention of 256KB of MCU SRAM. The CC3235x device can be configured to retain 0KB, 64KB, 128KB, 192KB, or 256KB of SRAM in LPDS. Each 64-KB block of MCU retained SRAM increases LPDS current by 4 µA.
- (6) The complete calibration can take up to 17 mJ of energy from the battery over a period of 24 ms. Calibration is performed sparingly, typically when coming out of HIBERNATE and only if temperature has changed by more than 20°C. The calibration event can be controlled by a configuration file in the serial flash.

## 5.7 TX Power Control for 2.4 GHz Band

The CC3235x has several options for modifying the output power of the device when required. For the 2.4 GHz band it is possible to lower the overall output power at a global level using the global TX power level setting. In addition, the 2.4 GHz band allows the user to enter additional back-offs<sup>(1)</sup>, per channel, region<sup>(2)</sup> and modulation rates<sup>(3)</sup>, via Image creator (see the [Uniflash with Image Creator User Guide](#) for more details).

Figure 5-1, Figure 5-2, and Figure 5-3 show TX power and IBAT versus TX power level settings for the CC3235S device at modulations of 1 DSSS, 6 OFDM, and 54 OFDM, respectively. For the CC3235SF device, the IBAT current has an increase of approximately 10 mA to 15 mA depending on the transmitted rate. The TX power level will remain the same.

In Figure 5-1, the area enclosed in the circle represents a significant reduction in current during transition from TX power level 3 to level 4. In the case of lower range requirements (14-dBm output power), TI recommends using TX power level 4 to reduce the current.

- (1) The back-off range is between -6 dB to +6 dB in 0.25dB increments.
- (2) FCC/ISED, ETSI (Europe), and Japan are supported.
- (3) Back-off rates are grouped into 11b rates, high modulation rates (MCS7, 54 OFDM and 48 OFDM), and lower modulation rates (all other rates).

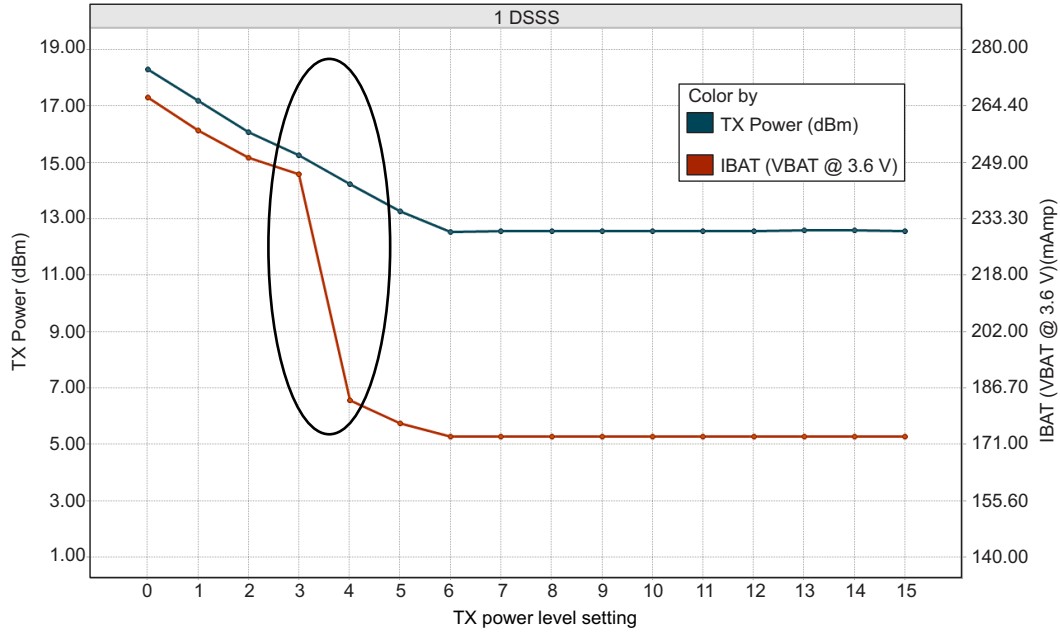


Figure 5-1. TX Power and IBAT vs TX Power Level Settings (1 DSSS)

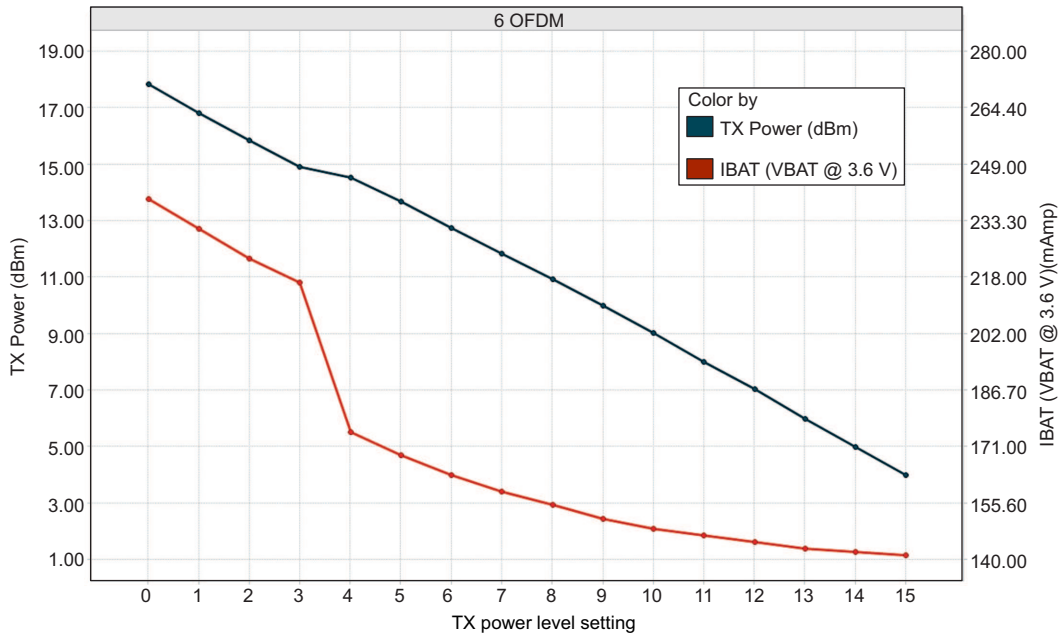


Figure 5-2. TX Power and IBAT vs TX Power Level Settings (6 OFDM)

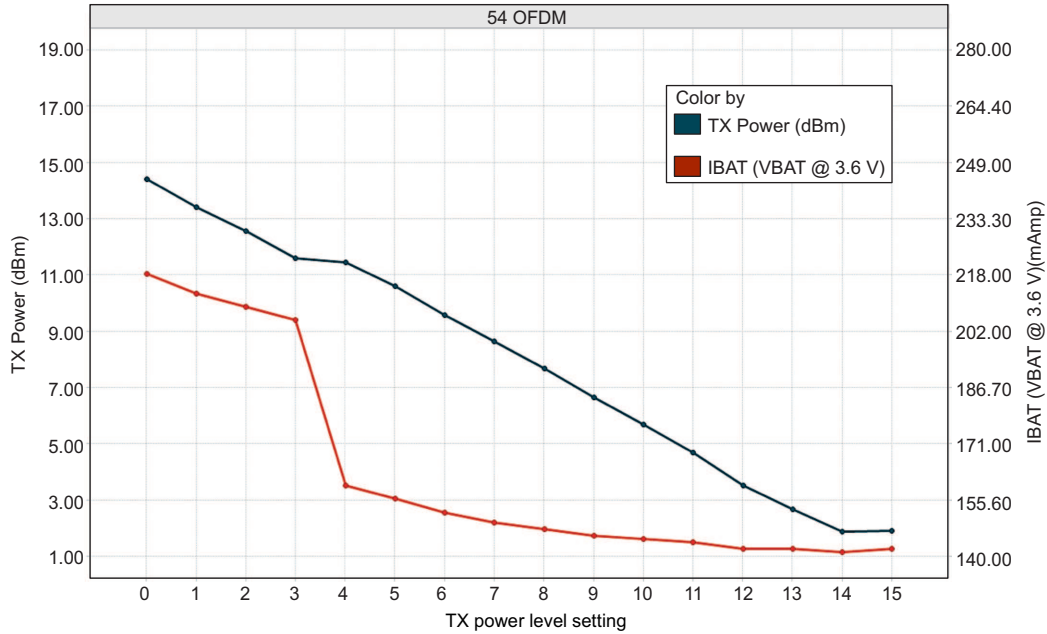


Figure 5-3. TX Power and IBAT vs TX Power Level Settings (54 OFDM)

### 5.8 TX Power Control for 5 GHz

5 GHz power control is done via Image Creator where the maximum transmit power is provided <sup>(1)</sup>. Within Image Creator power control is possible per channel, region <sup>(2)</sup>, and modulation rates <sup>(3)</sup>. In addition, it is possible to enter an additional back-off <sup>(4)</sup> factor per channel and modulation rate for further margin to regulatory requirements.

Finally, it is also possible to set the TX and RX trace losses to the antenna per band <sup>(5)</sup>. The peak antenna gain <sup>(6)</sup> can also be provided, thus allowing further control. For a full description of options and capabilities see [Uniflash with Image Creator User Guide](#).

- (1) The maximum transmit power range is 18dBm to 0.125dBm in 0.125dBm decrements.
- (2) FCC/ISED, ETSI (Europe), and Japan are supported.
- (3) Rates are grouped into high modulation rates (MCS7, 54 OFDM and 48 OFDM) and lower modulation rates (all other rates).
- (4) The back-off range is 0 dBm to 18 dBm in 0.125 dBm increments, with the maximum back-off not exceed that of the maximum transmit power.
- (5) The range of losses if from 0 dBm to 7.75 dBm in 0.125 dBm increments.
- (6) The antenna gain has a range of -2 dBi to 5.75 dBi in 0.125 dBi increments.

### 5.9 Brownout and Blackout Conditions

The device enters a brownout condition when the input voltage drops below  $V_{\text{brownout}}$  (see Figure 5-4 and Figure 5-5). This condition must be considered during design of the power supply routing, especially when operating from a battery. High-current operations, such as a TX packet or any external activity (not necessarily related directly to networking) can cause a drop in the supply voltage, potentially triggering a brownout condition. The resistance includes the internal resistance of the battery, the contact resistance of the battery holder (four contacts for 2x AA batteries), and the wiring and PCB routing resistance.

**NOTE**

When the device is in HIBERNATE state, brownout is not detected. Only blackout is in effect during HIBERNATE state.

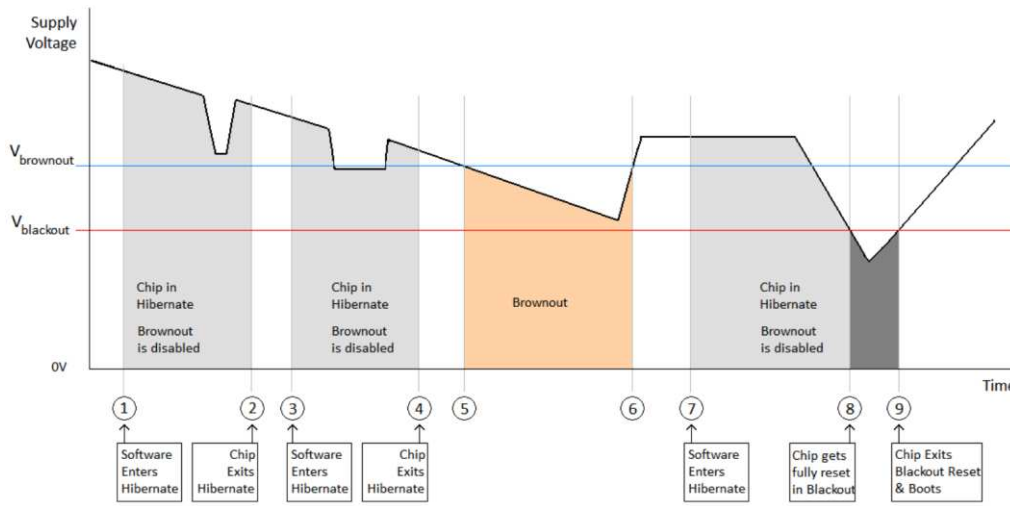


Figure 5-4. Brownout and Blackout Levels (1 of 2)

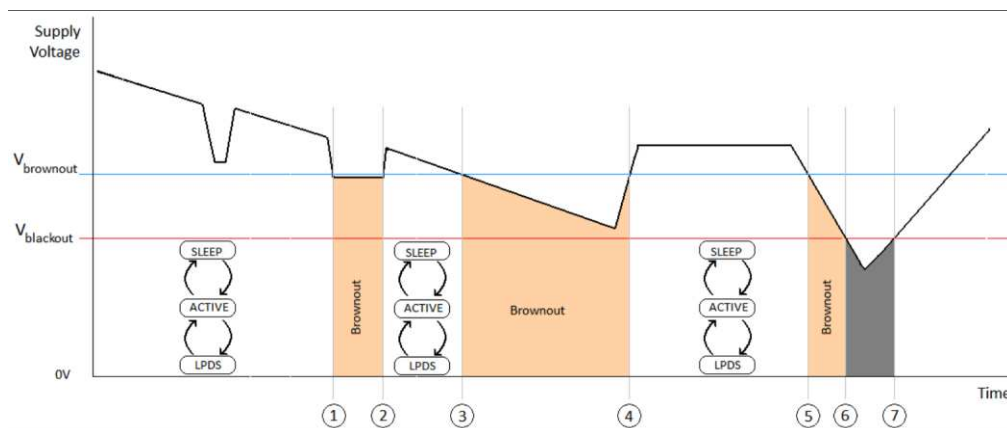


Figure 5-5. Brownout and Blackout Levels (2 of 2)

In the brownout condition, all sections of the device (including the 32-kHz RTC) shut down except for the Hibernate module, which remains on. The current in this state can reach approximately 400  $\mu$ A. The blackout condition is equivalent to a hardware reset event in which all states within the device are lost.

Table 5-5 lists the brownout and blackout voltage levels.

**Table 5-5. Brownout and Blackout Voltage Levels**

CONDITION	VOLTAGE LEVEL	UNIT
$V_{\text{brownout}}$	2.1	V
$V_{\text{blackout}}$	1.67	V

## 5.10 Electrical Characteristics for GPIO Pins

**Table 5-6. Electrical Characteristics: GPIO Pins Except 50, 52, and 53**
 $T_A = 25^\circ\text{C}$ ,  $V_{\text{BAT}} = 2.1\text{ V to }3.3\text{ V}$ .<sup>(1)</sup>

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$C_{\text{IN}}$	Pin capacitance			4		pF
$V_{\text{IH}}$	High-level input voltage		$0.65 \times V_{\text{DD}}$		$V_{\text{DD}} + 0.5\text{ V}$	V
$V_{\text{IL}}$	Low-level input voltage		-0.5		$0.35 \times V_{\text{DD}}$	V
$I_{\text{IH}}$	High-level input current			5		nA
$I_{\text{IL}}$	Low-level input current			5		nA
$V_{\text{OH}}$	High-level output voltage	IL = 2 mA; configured I/O drive strength = 2 mA; $2.4\text{ V} \leq V_{\text{DD}} < 3.6\text{ V}$			$V_{\text{DD}} \times 0.8$	V
		IL = 4 mA; configured I/O drive strength = 4 mA; $2.4\text{ V} \leq V_{\text{DD}} < 3.6\text{ V}$			$V_{\text{DD}} \times 0.7$	
		IL = 6 mA; configured I/O drive strength = 6 mA; $2.4\text{ V} \leq V_{\text{DD}} < 3.6\text{ V}$			$V_{\text{DD}} \times 0.7$	
		IL = 2 mA; configured I/O drive strength = 2 mA; $2.1\text{ V} \leq V_{\text{DD}} < 2.4\text{ V}$			$V_{\text{DD}} \times 0.75$	
$V_{\text{OL}}$	Low-level output voltage	IL = 2 mA; configured I/O drive strength = 2 mA; $2.4\text{ V} \leq V_{\text{DD}} < 3.6\text{ V}$	$V_{\text{DD}} \times 0.2$			V
		IL = 4 mA; configured I/O drive strength = 4 mA; $2.4\text{ V} \leq V_{\text{DD}} < 3.6\text{ V}$	$V_{\text{DD}} \times 0.2$			
		IL = 6 mA; configured I/O drive strength = 6 mA; $2.4\text{ V} \leq V_{\text{DD}} < 3.6\text{ V}$	$V_{\text{DD}} \times 0.2$			
		IL = 2 mA; configured I/O drive strength = 2 mA; $2.1\text{ V} \leq V_{\text{DD}} < 2.4\text{ V}$	$V_{\text{DD}} \times 0.25$			
$I_{\text{OH}}$	High-level source current	2-mA drive	2			mA
		4-mA drive	4			
		6-mA drive	6			
$I_{\text{OL}}$	Low-level sink current	2-mA drive	2			mA
		4-mA drive	4			
		6-mA drive	6			

(1) TI recommends using the lowest possible drive strength that is adequate for the applications. This recommendation minimizes the risk of interference to the WLAN radio and reduces any potential degradation of RF sensitivity and performance. The default drive strength setting is 6 mA.

**Table 5-7. Electrical Characteristics: GPIO Pins 50, 52, and 53** $T_A = 25^\circ\text{C}$ ,  $V_{\text{BAT}} = 2.1\text{ V to }3.6\text{ V}$ .<sup>(1)</sup>

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$C_{\text{IN}}$	Pin capacitance			7		pF
$V_{\text{IH}}$	High-level input voltage		$0.65 \times V_{\text{DD}}$		$V_{\text{DD}} + 0.5\text{ V}$	V
$V_{\text{IL}}$	Low-level input voltage		-0.5		$0.35 \times V_{\text{DD}}$	V
$I_{\text{IH}}$	High-level input current			50		nA
$I_{\text{IL}}$	Low-level input current			50		nA
$V_{\text{OH}}$	High-level output voltage	IL = 2 mA; configured I/O drive strength = 2 mA; $2.4\text{ V} \leq V_{\text{DD}} < 3.6\text{ V}$			$V_{\text{DD}} \times 0.8$	V
		IL = 4 mA; configured I/O drive strength = 4 mA; $2.4\text{ V} \leq V_{\text{DD}} < 3.6\text{ V}$			$V_{\text{DD}} \times 0.7$	
		IL = 6 mA; configured I/O drive strength = 6 mA; $2.4\text{ V} \leq V_{\text{DD}} < 3.6\text{ V}$			$V_{\text{DD}} \times 0.7$	
		IL = 2 mA; configured I/O drive strength = 2 mA; $2.1\text{ V} \leq V_{\text{DD}} < 2.4\text{ V}$			$V_{\text{DD}} \times 0.75$	
$V_{\text{OL}}$	Low-level output voltage	IL = 2 mA; configured I/O drive strength = 2 mA; $2.4\text{ V} \leq V_{\text{DD}} < 3.6\text{ V}$		$V_{\text{DD}} \times 0.2$		V
		IL = 4 mA; configured I/O drive strength = 4 mA; $2.4\text{ V} \leq V_{\text{DD}} < 3.6\text{ V}$		$V_{\text{DD}} \times 0.2$		
		IL = 6 mA; configured I/O drive strength = 6 mA; $2.4\text{ V} \leq V_{\text{DD}} < 3.6\text{ V}$		$V_{\text{DD}} \times 0.2$		
		IL = 2 mA; configured I/O drive strength = 2 mA; $2.1\text{ V} \leq V_{\text{DD}} < 2.4\text{ V}$		$V_{\text{DD}} \times 0.25$		
$I_{\text{OH}}$	High-level source current, $V_{\text{OH}} = 2.4$	2-mA drive		1.5		mA
		4-mA drive		2.5		
		6-mA drive		3.5		
$I_{\text{OL}}$	Low-level sink current	2-mA drive		1.5		mA
		4-mA drive		2.5		
		6-mA drive		3.5		
$V_{\text{IL}}$	nRESET			0.6		V

(1) TI recommends using the lowest possible drive strength that is adequate for the applications. This recommendation minimizes the risk of interference to the WLAN radio and reduces any potential degradation of RF sensitivity and performance. The default drive strength setting is 6 mA.

### 5.11 Electrical Characteristics for Pin Internal Pullup and Pulldown

 $T_A = 25^\circ\text{C}$ ,  $V_{\text{BAT}} = 3.0\text{ V}$ .

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$I_{\text{OH}}$	Pullup current, $V_{\text{OH}} = 2.4$ ( $V_{\text{DD}} = 3.0\text{ V}$ )		5		10	$\mu\text{A}$
$I_{\text{OL}}$	Pulldown current, $V_{\text{OL}} = 0.4$ ( $V_{\text{DD}} = 3.0\text{ V}$ )		5			$\mu\text{A}$

## 5.12 WLAN Receiver Characteristics

**Table 5-8. WLAN Receiver Characteristics: 2.4 GHz Band**

$T_A = 25^\circ\text{C}$ ,  $V_{\text{BAT}} = 2.1\text{ V to }3.6\text{ V}$ . Parameters are measured at the SoC pin on channel 6 (2437 MHz).

PARAMETER	TEST CONDITIONS (Mbps)	MIN	TYP	MAX	UNIT
Sensitivity (8% PER for 11b rates, 10% PER for 11g/11n rates) <sup>(1)</sup>	1 DSSS		-96.0		dBm
	2 DSSS		-94.0		
	11 CCK		-88.0		
	6 OFDM		-90.5		
	9 OFDM		-90.0		
	18 OFDM		-86.5		
	36 OFDM		-80.5		
	54 OFDM		-74.5		
	MCS7 (GF) <sup>(2)</sup>		-71.5		
Maximum input level (10% PER)	802.11b		-4.0		dBm
	802.11g		-10.0		

(1) Sensitivity is 1-dB worse on channel 13 (2472 MHz).

(2) Sensitivity for mixed mode is 1-dB worse.

**Table 5-9. WLAN Receiver Characteristics: 5 GHz Band**

$T_A = 25^\circ\text{C}$ ,  $V_{\text{BAT}} = 2.1\text{ V to }3.6\text{ V}$ . Parameters measured at SoC pin are the average of channels 40, 56, 120, and 157.

PARAMETER	TEST CONDITIONS (Mbps)	MIN	TYP	MAX	UNIT
Sensitivity (10% PER for 11g/11n rates)	6 OFDM		-92.0		dBm
	9 OFDM		-91.0		
	18 OFDM		-88.0		
	36 OFDM		-81.5		
	54 OFDM		-75.0		
		MCS7 (GF) <sup>(1)</sup>		-71.0	
Maximum input level	802.11a		-20		dBm

(1) Sensitivity for mixed mode is 1-dB worse.

## 5.13 WLAN Transmitter Characteristics

**Table 5-10. WLAN Transmitter Characteristics: 2.4 GHz Band**
 $T_A = 25^\circ\text{C}$ ,  $V_{\text{BAT}} = 2.1\text{ V to }3.6\text{ V}$ . Parameters measured at SoC pin on channel 6 (2437 MHz).<sup>(1)(2)</sup>

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Operating frequency range <sup>(3)(4)</sup>		2412		2472	MHz
Maximum RMS output power measured at 1 dB from IEEE spectral mask or EVM	1 DSSS		18.0		dBm
	2 DSSS		18.0		
	11 CCK		18.3		
	6 OFDM		17.3		
	9 OFDM		17.3		
	18 OFDM		17.0		
	36 OFDM		16.0		
	54 OFDM		14.5		
	MCS7		13.0		
Transmit center frequency accuracy		-25		25	ppm

(1) The OFDM and MCS7 edge channels (2412 and 2462 MHz) have reduced TX power to meet FCC emission limits.

(2) Power of 802.11b rates are reduced to meet ETSI requirements in Europe.

(3) Channels 1 (2142 MHz) through 11 (2462 MHz) are supported for FCC.

(4) Channels 1 (2142 MHz) through 13 (2472MHz) are supported for Europe and Japan. Note that channel 14 is not supported for Japan.

**Table 5-11. WLAN Transmitter Characteristics: 5 GHz Band**
 $T_A = 25^\circ\text{C}$ ,  $V_{\text{BAT}} = 2.1\text{ V to }3.6\text{ V}$ .<sup>(1)</sup> Parameters measured at SoC pin are the average of channels 40, 56, 120, and 157.<sup>(2)</sup>

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Operating frequency range <sup>(3)(4)(5)</sup>		5180		5825	MHz
Maximum RMS output power measured at 1 dB from IEEE spectral mask or EVM	6 OFDM		18.1		dBm
	9 OFDM		18.1		
	18 OFDM		18.1		
	36 OFDM		16.6		
	54 OFDM		15.0		
	MCS7		14.0		
Transmit center frequency accuracy		-20		20	ppm

(1) Transmit power will be reduced by 1.5dB for  $V_{\text{BAT}} < 2.8\text{V}$

(2) FCC, Europe, and Japan channel power limits per modulation rates can be found in the [Uniflash with Image Creator User Guide](#).

(3) FCC band covers U-NII-1, U-NII-2A, U-NII-2C, and U-NII-3 20-MHz BW modulations.

(4) Europe bands 1, 2 and 3, 20-MHz BW modulations are supported.

(5) For Japan, W52, W53 and W56, 20-MHz BW modulations are supported.



## 5.14 WLAN Transmitter Out-of-Band Emissions

Both the 2.4 GHz and the 5 GHz RF paths require an external band-pass filter to meet the various emission standards, including FCC. [Table 5-12](#) and [Table 5-13](#) presents the minimum attenuation requirements for the 2.4 GHz and 5 GHz band-pass filter, respectively. TI recommends using the same filter, switch, diplexer, and so on, used in the reference design to ease the process of certification.

**Table 5-12. WLAN 2.4 GHz Filter Requirements**

PARAMETER	FREQUENCY (MHz)	MIN	TYP	MAX	UNIT
Return loss	2412 to 2484	10			dB
Insertion loss <sup>(1)</sup>	2412 to 2484		1	1.5	dB
Attenuation	804 to 828	30	42		dB
	1608 to 1656	20	23		
	3216 to 3312	30	49		
	4020 to 4140	40	52		
	4824 to 4968	20	30		
	5628 to 5796	20	27		
	6432 to 6624	20	42		
	7200 to 7500	35	44		
	7500 to 10000	20	30		
Reference impedance	2412 to 2484		50		Ω
Filter type	Bandpass				

(1) Insertion loss directly impacts output power and sensitivity. At customer discretion, insertion loss can be relaxed to meet attenuation requirements.

**Table 5-13. WLAN 5 GHz Filter Requirements**

PARAMETER	FREQUENCY (MHz)	MIN	TYP	MAX	UNIT
Return loss	5150 to 5925	10			dB
Insertion loss <sup>(1)</sup>	5150 to 5925		1	2	dB
Attenuation	600 to 2700	41	42		dB
	2950 to 3850	27	31		
	4400 to 4600	20	27		
	6600 to 6900	20	28		
	7000 to 7775	20	27		
	10300 to 11850	25	37		
Reference impedance	5150 to 5925		50		Ω
Filter type	Bandpass				

(1) Insertion loss directly impacts output power and sensitivity. At customer discretion, insertion loss can be relaxed to meet attenuation requirements.

## 5.15 BLE/2.4 GHz Radio Coexistence and WLAN Coexistence Requirements

For proper BLE/2.4 GHz radio coexistence, the following requirements needs to met:

**Table 5-14. COEX Isolation Requirement**

PARAMETER	Band	MIN	TYP	MAX	UNIT
Port-to-port isolation	Single antenna	20 <sup>(1)</sup>			dB
	Dual antenna Configuration	20 <sup>(2)</sup>			

(1) WLAN/BLE switch used must provide a minimum of 20 dB isolation between ports.

(2) For dual antenna configuration antenna placement must be such that isolation between the BLE and WLAN ports is at least 20 dB.

## 5.16 Thermal Resistance Characteristics for RGK Package

THERMAL METRICS <sup>(1)</sup>		°C/W <sup>(2) (3)</sup>	AIR FLOW (m/s) <sup>(4)</sup>
R <sub>θJC</sub>	Junction-to-case	6.3	0.0051
R <sub>θJB</sub>	Junction-to-board	2.4	0.0051
R <sub>θJA</sub>	Junction-to-free air	23	0.0051
R <sub>θJMA</sub>	Junction-to-moving air	14.6	0.765
		12.4	1.275
		10.8	2.55
Psi <sub>JT</sub>	Junction-to-package top	0.2	0.0051
		0.2	0.765
		0.3	1.275
Psi <sub>JB</sub>	Junction-to-board	0.1	2.55
		2.3	0.0051
		2.3	0.765
		2.2	1.275
		2.4	2.55

(1) For more information about traditional and new thermal metrics, see [Semiconductor and IC Package Thermal Metrics](#).

(2) °C/W = degrees Celsius per watt.

(3) These values are based on a JEDEC-defined 2S2P system (with the exception of the Theta JC [R<sub>θJC</sub>] value, which is based on a JEDEC-defined 1S0P system) and will change based on environment as well as application. For more information, see these EIA/JEDEC standards:

- JESD51-2, *Integrated Circuits Thermal Test Method Environmental Conditions - Natural Convection (Still Air)*
- JESD51-3, *Low Effective Thermal Conductivity Test Board for Leaded Surface Mount Packages*
- JESD51-7, *High Effective Thermal Conductivity Test Board for Leaded Surface Mount Packages*
- JESD51-9, *Test Boards for Area Array Surface Mount Package Thermal Measurements*

Power dissipation of 2 W and an ambient temperature of 70°C is assumed.

(4) m/s = meters per second.

## 5.17 Timing and Switching Characteristics

### 5.17.1 Power Supply Sequencing

For proper operation of the CC3235x device, perform the recommended power-up sequencing as follows:

1. Tie the following pins together on the board:
  - V<sub>BAT</sub> (pins 37, 39, and 44)
  - V<sub>IO</sub> (pins 54 and 10)
2. Hold the RESET pin low while the supplies are ramping up. TI recommends using a simple RC circuit (100 K ||, 0.01 μF, RC = 1 ms).
3. For an external RTC, ensure that the clock is stable before RESET is deasserted (high).

For timing diagrams, see [Section 5.17.3](#).

### 5.17.2 Device Reset

When a device restart is required, the user may issue a negative pulse to the nRESET pin. The user must follow one of the following alternatives to ensure the reset is properly applied:

- A negative reset pulse (on pin 32) of at least 200-ms duration
- If the 200-ms pulse duration cannot be ensured, a pulldown resistor of 2 MΩ must be connected to pin 52 (RTC\_XTAL\_N). If implemented, a shorter pulse of at least 100 μs can be used.

To ensure a proper reset sequence, the user must call the sl\_stop function prior to toggling the reset. When a reset is required, it is preferable to use the software reset instead of an external trigger.

### 5.17.3 Reset Timing

#### 5.17.3.1 nRESET (32-kHz Crystal)

Figure 5-6 shows the reset timing diagram for the 32-kHz crystal first-time power-up and reset removal.

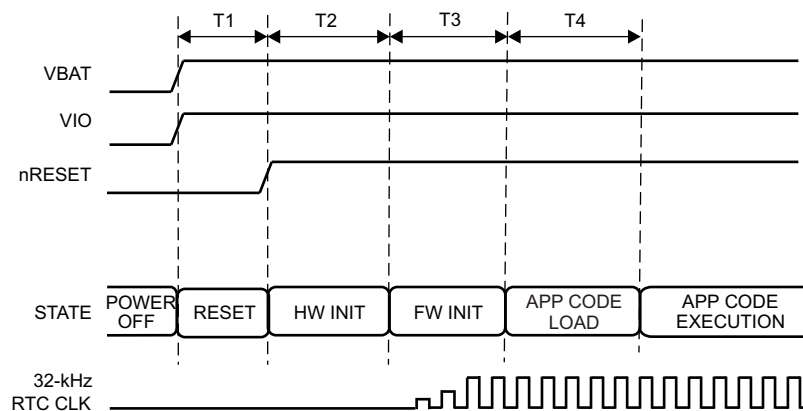


Figure 5-6. First-Time Power-Up and Reset Removal Timing Diagram (32-kHz Crystal)

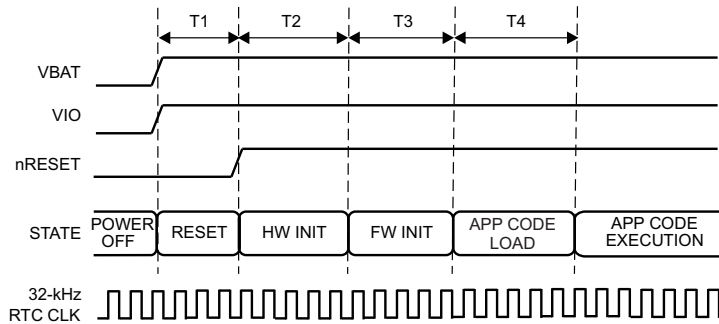
Table 5-15 describes the timing requirements for the 32-kHz clock crystal first-time power-up and reset removal.

Table 5-15. First-Time Power-Up and Reset Removal Timing Requirements (32-kHz Crystal)

ITEM	NAME	DESCRIPTION	MIN	NOM	MAX	UNIT
T1	nReset timing	nReset timing after VBAT and VIO supply are stable		1		ms
T2	Hardware wake-up time			25		ms
T3	Time taken by ROM firmware to initialize hardware	Includes 32.768-kHz XOSC settling time		1.1		s
T4	App code load time for CC3235S	CC3235S	Image size (KB) × 1.7 ms			
	App code integrity check time for CC3235SF	CC3235SF	Image size (KB) × 0.06 ms			

**5.17.3.2 nRESET (External 32-kHz Clock)**

Figure 5-7 shows the reset timing diagram for the external 32-kHz clock first-time power-up and reset removal.



**Figure 5-7. First-Time Power-Up and Reset Removal Timing Diagram (External 32-kHz Clock)**

Table 5-16 describes the timing requirements for the external 32-kHz clock first-time power-up and reset removal.

**Table 5-16. First-Time Power-Up and Reset Removal Timing Requirements (External 32-kHz Clock)**

ITEM	NAME	DESCRIPTION	MIN	NOM	MAX	UNIT
T1	nReset time	nReset timing after VBAT and VIO supply are stable		1		ms
T2	Hardware wake-up time			25		ms
T3	Time taken by ROM firmware to initialize hardware	CC3235S		10.3		ms
		CC3235SF		17.3		
T4	App code load time for CC3235R and CC3235S	CC3235S		Image size (KB) × 1.7 ms		
	App code integrity check time for CC3235SF	CC3235SF		Image size (KB) × 0.06 ms		

### 5.17.4 Wakeup From HIBERNATE Mode

**NOTE**

The 32.768-kHz crystal is enabled by default when the chip goes into HIBERNATE mode.

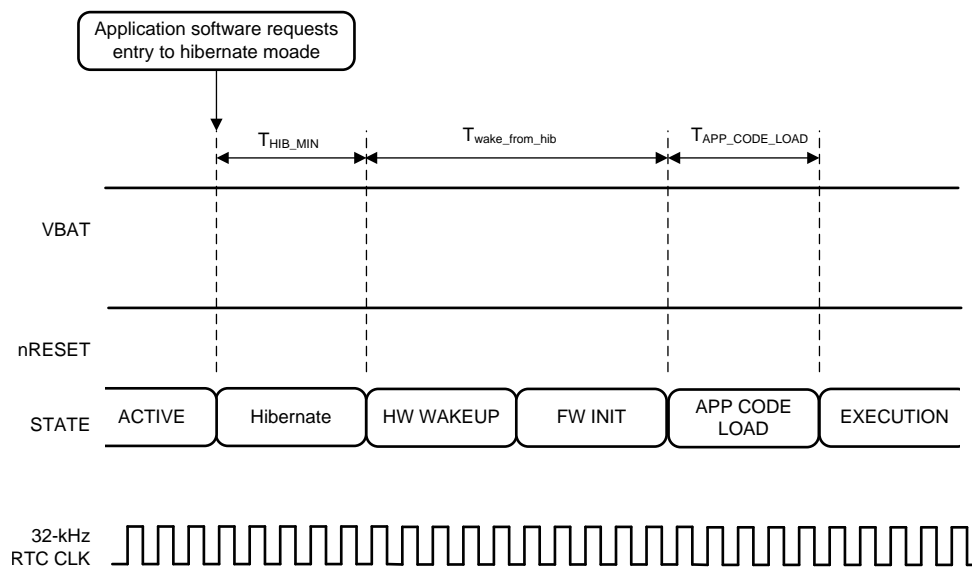
Table 5-17 lists the software hibernate timing requirements.

**Table 5-17. Software Hibernate Timing Requirements**

ITEM	NAME	DESCRIPTION	MIN	TYP	MAX	UNIT
T <sub>HIB_MIN</sub>	Minimum hibernate time		10			ms
T <sub>wake_from_hib</sub> <sup>(1)</sup>	Hardware wakeup time plus firmware initialization time			50 <sup>(2)</sup>		ms
T <sub>APP_CODE_LOAD</sub>	App code load time for CC3235S	CC3235S	Image size (KB) × 1.7 ms			ms
	App code load time for CC3235SF	CC3235SF	Image size (KB) × 0.06 ms			

- (1) T<sub>wake\_from\_hib</sub> can be 200 ms on rare occasions when calibration is performed. Calibration is performed sparingly, typically when exiting Hibernate and only if temperature has changed by more than 20°C or more than 24 hours have elapsed since a prior calibration.
- (2) Wake-up time can extend to 75 ms if a patch is downloaded from the serial Flash.

Figure 5-8 shows the timing diagram for wakeup from HIBERNATE mode.



**Figure 5-8. Wakeup From HIBERNATE Timing Diagram**

### 5.17.5 Clock Specifications

The CC3235x device requires two separate clocks for operation:

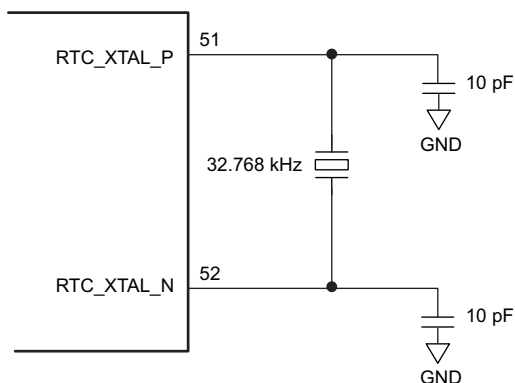
- A slow clock running at 32.768 kHz is used for the RTC.
- A fast clock running at 40 MHz is used by the device for the internal processor and the WLAN subsystem.

The device features internal oscillators that enable the use of less-expensive crystals rather than dedicated TCXOs for these clocks. The RTC can also be fed externally to provide reuse of an existing clock on the system and to reduce overall cost.

#### 5.17.5.1 Slow Clock Using Internal Oscillator

The RTC crystal connected on the device supplies the free-running slow clock. The accuracy of the slow clock frequency must be 32.768 kHz ±150 ppm. In this mode of operation, the crystal is tied between RTC\_XTAL\_P (pin 51) and RTC\_XTAL\_N (pin 52) with a suitable load capacitance to meet the ppm requirement.

Figure 5-9 shows the crystal connections for the slow clock.



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**Figure 5-9. RTC Crystal Connections**

Table 5-18 lists the RTC crystal requirements.

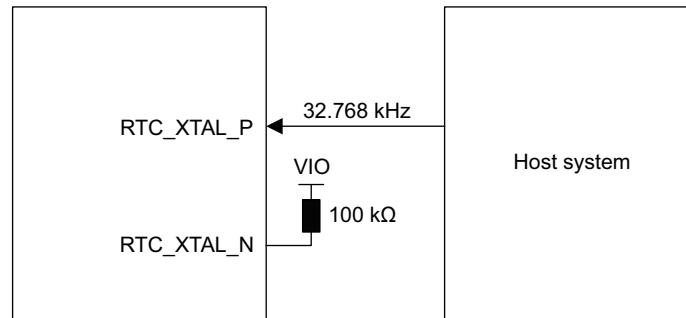
**Table 5-18. RTC Crystal Requirements**

CHARACTERISTICS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Frequency			32.768		kHz
Frequency accuracy	Initial plus temperature plus aging			±150	ppm
Crystal ESR	32.768 kHz			70	kΩ

### 5.17.5.2 Slow Clock Using an External Clock

When an RTC oscillator is present in the system, the CC3235x device can accept this clock directly as an input. The clock is fed on the RTC\_XTAL\_P line, and the RTC\_XTAL\_N line is held to  $V_{IO}$ . The clock must be a CMOS-level clock compatible with  $V_{IO}$  fed to the device.

Figure 5-10 shows the external RTC input connection.



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Figure 5-10. External RTC Input

Table 5-19 lists the external RTC digital clock requirements.

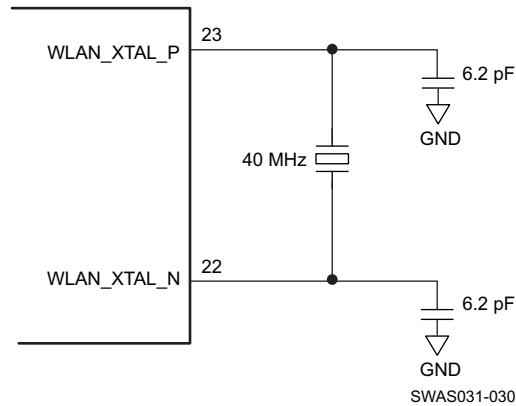
Table 5-19. External RTC Digital Clock Requirements

CHARACTERISTICS		TEST CONDITIONS	MIN	TYP	MAX	UNIT
Frequency				32768		Hz
Frequency accuracy (Initial plus temperature plus aging)				±150		ppm
$t_r, t_f$	Input transition time $t_r, t_f$ (10% to 90%)				100	ns
Frequency input duty cycle			20%	50%	80%	
$V_{in}$	Slow clock input voltage limits	Square wave, DC coupled	$0.65 \times V_{IO}$		$V_{IO}$	V
$V_{il}$			0		$0.35 \times V_{IO}$	$V_{peak}$
Input impedance			1			MΩ
					5	pF

### 5.17.5.3 Fast Clock ( $F_{ref}$ ) Using an External Crystal

The CC3235x device also incorporates an internal crystal oscillator to support a crystal-based fast clock. The crystal is fed directly between WLAN\_XTAL\_P (pin 23) and WLAN\_XTAL\_N (pin 22) with suitable loading capacitors.

Figure 5-11 shows the crystal connections for the fast clock.



SWAS031-030

NOTE: The crystal capacitance must be tuned to ensure that the PPM requirement is met. See [CC31xx & CC32xx Frequency Tuning](#) for information on frequency tuning.

**Figure 5-11. Fast Clock Crystal Connections**

Table 5-20 lists the WLAN fast-clock crystal requirements.

**Table 5-20. WLAN Fast-Clock Crystal Requirements**

CHARACTERISTICS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Frequency			40		MHz
Frequency accuracy	Initial plus temperature plus aging			±20	ppm
Crystal ESR	40 MHz			60	Ω

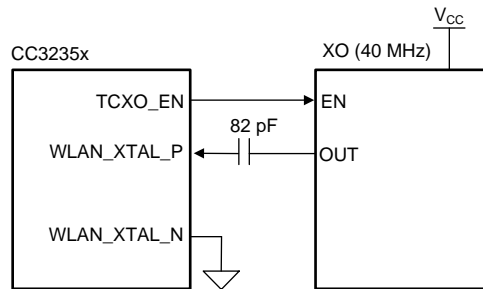


### 5.17.5.4 Fast Clock ( $F_{ref}$ ) Using an External Oscillator

The CC3235x device can accept an external TCXO/XO for the 40-MHz clock. In this mode of operation, the clock is connected to WLAN\_XTAL\_P (pin 23). WLAN\_XTAL\_N (pin 22) is connected to GND. The external TCXO/XO can be enabled by TCXO\_EN (pin 21) from the device to optimize the power consumption of the system.

If the TCXO does not have an enable input, an external LDO with an enable function can be used. Using the LDO improves noise on the TCXO power supply.

Figure 5-12 shows the connection.



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Figure 5-12. External TCXO Input

Table 5-21 lists the external  $F_{ref}$  clock requirements.

Table 5-21. External  $F_{ref}$  Clock Requirements (–40°C to +85°C)

CHARACTERISTICS		TEST CONDITIONS	MIN	TYP	MAX	UNIT
Frequency				40.00		MHz
Frequency accuracy (initial plus temperature plus aging)					±20	ppm
Frequency input duty cycle			45%	50%	55%	
$V_{pp}$	Clock voltage limits	Sine or clipped sine wave, AC coupled	0.7		1.2	$V_{pp}$
Phase noise at 40 MHz		at 1 kHz			–125	dBc/Hz
		at 10 kHz			–138.5	
		at 100 kHz			–143	
Input impedance	Resistance		12			kΩ
	Capacitance				7	pF

### 5.17.6 Peripherals Timing

This section describes the peripherals that are supported by the CC3235x device:

- SPI
- I<sup>2</sup>S
- GPIOs
- I<sup>2</sup>C
- IEEE 1149.1 JTAG
- ADC
- Camera Parallel Port
- UART
- SD Host
- Timers

### 5.17.6.1 SPI

#### 5.17.6.1.1 SPI Master

The CC3235x microcontroller includes one SPI module that can be configured as a master or slave device. The SPI includes a serial clock with programmable frequency, polarity, and phase; a programmable timing control between chip select and external clock generation; and a programmable delay before the first SPI word is transmitted. Slave mode does not include a dead cycle between two successive words.

Figure 5-13 shows the timing diagram for the SPI master.

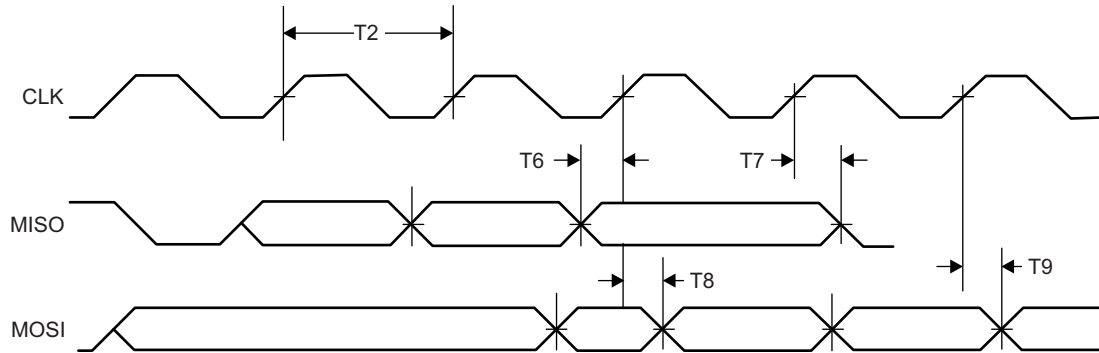


Figure 5-13. SPI Master Timing Diagram

Table 5-22 lists the timing parameters for the SPI master.

Table 5-22. SPI Master Timing Parameters

PARAMETER NUMBER			MIN	MAX	UNIT
	F <sup>(1)</sup>	Clock frequency		30	MHz
T2	T <sub>clk</sub> <sup>(1)</sup>	Clock period	33.3		ns
	D <sup>(1)</sup>	Duty cycle	45%	55%	
T6	t <sub>IS</sub> <sup>(1)</sup>	RX data setup time	1		ns
T7	t <sub>IH</sub> <sup>(1)</sup>	RX data hold time	2		ns
T8	t <sub>OD</sub> <sup>(1)</sup>	TX data output delay		8.5	ns
T9	t <sub>OH</sub> <sup>(1)</sup>	TX data hold time		8	ns

(1) Timing parameter assumes a maximum load of 20 pF.

5.17.6.1.2 SPI Slave

Figure 5-14 shows the timing diagram for the SPI slave.

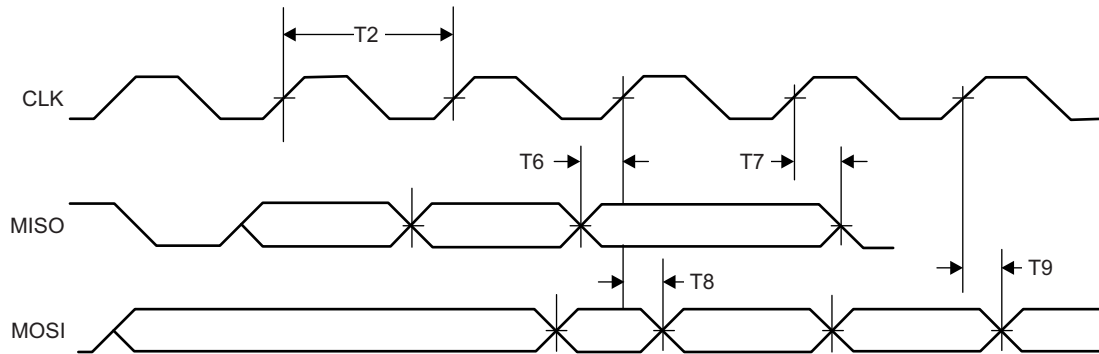


Figure 5-14. SPI Slave Timing Diagram

Table 5-23 lists the timing parameters for the SPI slave.

Table 5-23. SPI Slave Timing Parameters

PARAMETER NUMBER			MIN	MAX	UNIT
	F <sup>(1)</sup>	Clock frequency at V <sub>BAT</sub> = 3.3 V		20	MHz
		Clock frequency at V <sub>BAT</sub> ≤ 2.1 V		12	
T2	T <sub>clk</sub> <sup>(1)</sup>	Clock period	50		ns
	D <sup>(1)</sup>	Duty cycle	45%	55%	
T6	t <sub>IS</sub> <sup>(1)</sup>	RX data setup time	4		ns
T7	t <sub>IH</sub> <sup>(1)</sup>	RX data hold time	4		ns
T8	t <sub>OD</sub> <sup>(1)</sup>	TX data output delay		20	ns
T9	t <sub>OH</sub> <sup>(1)</sup>	TX data hold time		24	ns

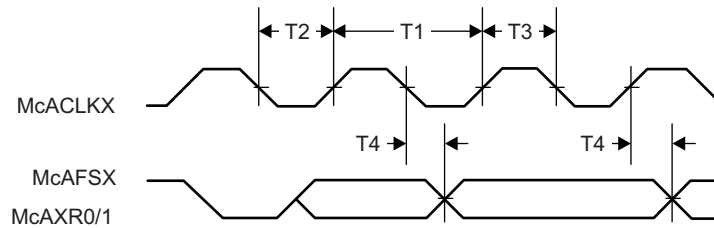
(1) Timing parameter assumes a maximum load of 20 pF at 3.3 V.

**5.17.6.2 I<sup>2</sup>S**

The McASP interface functions as a general-purpose audio serial port optimized for multichannel audio applications and supports transfer of two stereo channels over two data pins. The McASP consists of transmit and receive sections that operate synchronously and have programmable clock and frame-sync polarity. A fractional divider is available for bit-clock generation.

**5.17.6.2.1 I<sup>2</sup>S Transmit Mode**

Figure 5-15 shows the timing diagram for the I<sup>2</sup>S transmit mode.



**Figure 5-15. I<sup>2</sup>S Transmit Mode Timing Diagram**

Table 5-24 lists the timing parameters for the I<sup>2</sup>S transmit mode.

**Table 5-24. I<sup>2</sup>S Transmit Mode Timing Parameters**

PARAMETER NUMBER			MIN	MAX	UNIT
T1	$f_{clk}^{(1)}$	Clock frequency		9.216	MHz
T2	$t_{LP}^{(1)}$	Clock low period		1/2 fclk	ns
T3	$t_{HT}^{(1)}$	Clock high period		1/2 fclk	ns
T4	$t_{OH}^{(1)}$	TX data hold time		22	ns

(1) Timing parameter assumes a maximum load of 20 pF.

### 5.17.6.2.2 I<sup>2</sup>S Receive Mode

Figure 5-16 shows the timing diagram for the I<sup>2</sup>S receive mode.

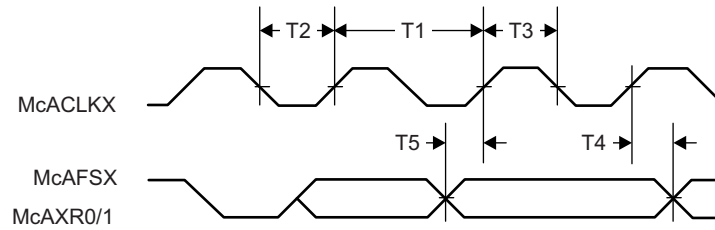


Figure 5-16. I<sup>2</sup>S Receive Mode Timing Diagram

Table 5-25 lists the timing parameters for the I<sup>2</sup>S receive mode.

Table 5-25. I<sup>2</sup>S Receive Mode Timing Parameters

PARAMETER NUMBER			MIN	MAX	UNIT
T1	$f_{\text{clk}}^{(1)}$	Clock frequency		9.216	MHz
T2	$t_{\text{LP}}^{(1)}$	Clock low period		$1/2 f_{\text{clk}}$	ns
T3	$t_{\text{HT}}^{(1)}$	Clock high period		$1/2 f_{\text{clk}}$	ns
T4	$t_{\text{OH}}^{(1)}$	RX data hold time		0	ns
T5	$t_{\text{OS}}^{(1)}$	RX data setup time		15	ns

(1) Timing parameter assumes a maximum load of 20 pF.

5.17.6.3 GPIOs

All digital pins of the device can be used as general-purpose input/output (GPIO) pins. The GPIO module consists of four GPIO blocks, each of which provides eight GPIOs. The GPIO module supports 24 programmable GPIO pins, depending on the peripheral used. Each GPIO has configurable pullup and pulldown strength (weak 10  $\mu$ A), configurable drive strength (2, 4, and 6 mA), and open-drain enable.

**NOTE**

Unless otherwise stated, GPIO specifications also applies to pins configured as COEX IOs and network scripiter interface

Figure 5-17 shows the GPIO timing diagram.

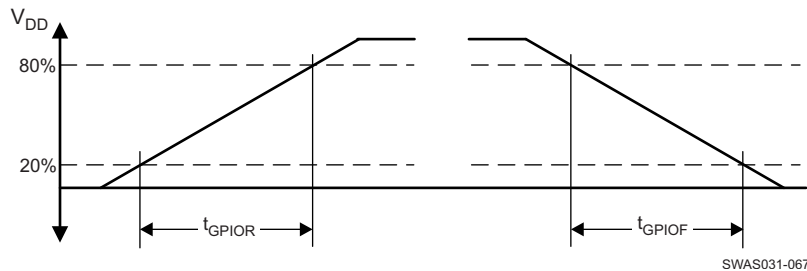


Figure 5-17. GPIO Timing Diagram

5.17.6.3.1 GPIO Output Transition Time Parameters ( $V_{supply} = 3.3 V$ )

Table 5-26 lists the GPIO output transition times for  $V_{supply} = 3.3 V$ .

Table 5-26. GPIO Output Transition Times ( $V_{supply} = 3.3 V$ )<sup>(1)(2)</sup>

DRIVE STRENGTH (mA)	DRIVE STRENGTH CONTROL BITS	$t_r$			$t_f$			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	
2 <sup>(3)</sup>	2MA_EN=1	8.0	9.3	10.7	8.2	9.5	11.0	ns
	4MA_EN=0							
4 <sup>(3)</sup>	2MA_EN=0	6.6	7.1	7.6	4.7	5.2	5.8	ns
	4MA_EN=1							
6	2MA_EN=1	3.2	3.5	3.7	2.3	2.6	2.9	ns
	4MA_EN=1							

- (1)  $V_{supply} = 3.3 V$ ,  $T = 25^\circ C$ , total pin load = 30 pF
- (2) The transition data applies to the pins except the multiplexed analog-digital pins 29, 30, 45, 50, 52, and 53.
- (3) The 2-mA and 4-mA drive strength does not apply to the COEX I/O pins. Pins configured as COEX lines are invariably driven at 6 mA.

5.17.6.3.2 GPIO Input Transition Time Parameters

Table 5-27 lists the input transition time parameters.

Table 5-27. GPIO Input Transition Time Parameters

		MIN	MAX	UNIT
$t_r$	Input transition time ( $t_r, t_f$ ), 10% to 90%	1	3	ns
$t_f$		1	3	ns

### 5.17.6.4 I<sup>2</sup>C

The CC3235x microcontroller includes one I<sup>2</sup>C module operating with standard (100 kbps) or fast (400 kbps) transmission speeds.

Figure 5-18 shows the I<sup>2</sup>C timing diagram.

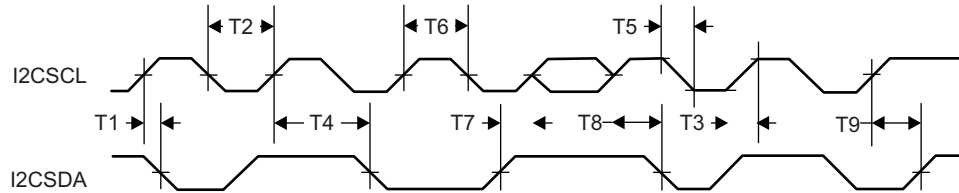


Figure 5-18. I<sup>2</sup>C Timing Diagram

Table 5-28 lists the I<sup>2</sup>C timing parameters.

Table 5-28. I<sup>2</sup>C Timing Parameters<sup>(1)</sup>

PARAMETER NUMBER			MIN	MAX	UNIT
T2	t <sub>LP</sub>	Clock low period	See <sup>(2)</sup>		System clock
T3	t <sub>SRT</sub>	SCL/SDA rise time		See <sup>(3)</sup>	ns
T4	t <sub>DH</sub>	Data hold time	N/A		
T5	t <sub>SFT</sub>	SCL/SDA fall time	3		ns
T6	t <sub>HT</sub>	Clock high time	See <sup>(2)</sup>		System clock
T7	t <sub>DS</sub>	Data setup time	t <sub>LP</sub> /2		System clock
T8	t <sub>SCSR</sub>	Start condition setup time	36		System clock
T9	t <sub>SCS</sub>	Stop condition setup time	24		System clock

(1) All timing is with 6-mA drive and 20-pF load.

(2) This value depends on the value programmed in the clock period register of I<sup>2</sup>C. Maximum output frequency is the result of the minimal value programmed in this register.

(3) Because I<sup>2</sup>C is an open-drain interface, the controller can drive logic 0 only. Logic is the result of an external pullup resistor. Rise time depends on the value of the external signal capacitance and external pullup register.

### 5.17.6.5 IEEE 1149.1 JTAG

The Joint Test Action Group (JTAG) port is an IEEE standard that defines a test access port (TAP) and boundary scan architecture for digital integrated circuits and provides a standardized serial interface to control the associated test logic. For detailed information on the operation of the JTAG port and TAP controller, see the IEEE Standard 1149.1, *Test Access Port and Boundary-Scan Architecture*.

Figure 5-19 shows the JTAG timing diagram.

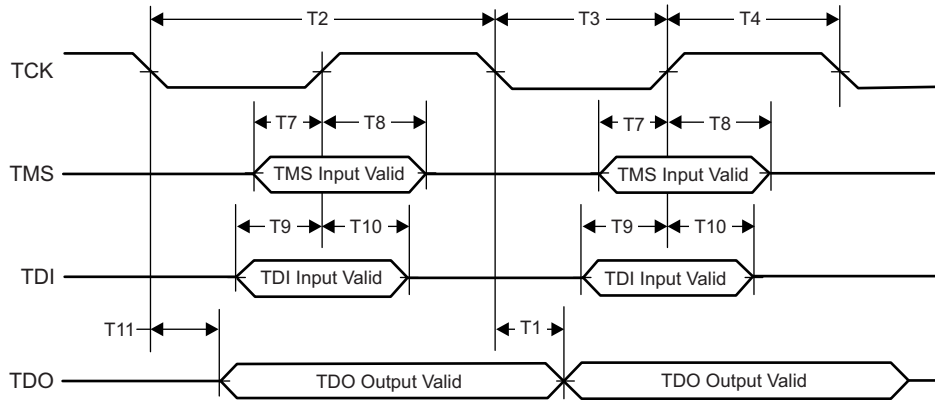


Figure 5-19. JTAG Timing Diagram

Table 5-29 lists the JTAG timing parameters.

Table 5-29. JTAG Timing Parameters

PARAMETER NUMBER			MIN	MAX	UNIT
T1	$f_{TCK}$	Clock frequency		15	MHz
T2	$t_{TCK}$	Clock period		$1 / f_{TCK}$	ns
T3	$t_{CL}$	Clock low period		$t_{TCK} / 2$	ns
T4	$t_{CH}$	Clock high period		$t_{TCK} / 2$	ns
T7	$t_{TMS\_SU}$	TMS setup time	1		ns
T8	$t_{TMS\_HO}$	TMS hold time	16		ns
T9	$t_{TDI\_SU}$	TDI setup time	1		ns
T10	$t_{TDI\_HO}$	TDI hold time	16		ns
T11	$t_{TDO\_HO}$	TDO hold time		15	ns



5.17.6.6 ADC

Figure 5-20 shows the ADC clock timing diagram.

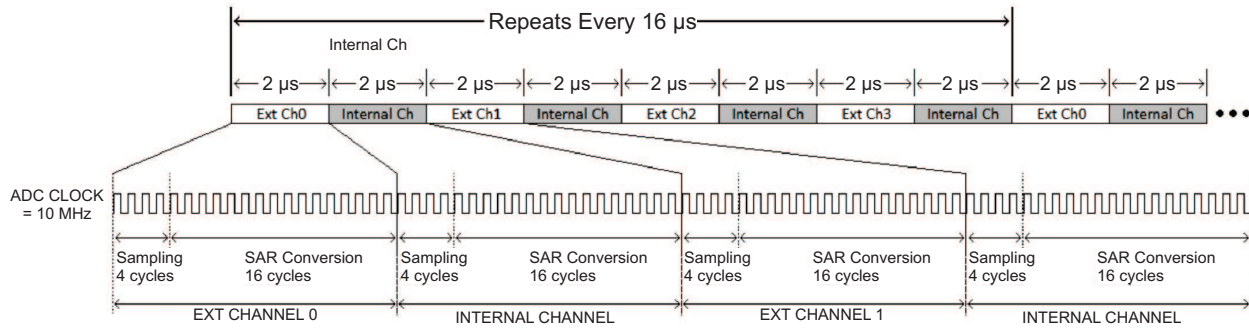


Figure 5-20. ADC Clock Timing Diagram

Table 5-30 lists the ADC electrical specifications. See [CC32xx ADC Appnote](#) for further information on using the ADC and for application-specific examples.

Table 5-30. ADC Electrical Specifications

PARAMETER	DESCRIPTION	TEST CONDITIONS and ASSUMPTIONS	MIN	TYP	MAX	UNIT
Nbits	Number of bits			12		Bits
INL	Integral nonlinearity	Worst-case deviation from histogram method over full scale (not including first and last three LSB levels)	-2.5		2.5	LSB
DNL	Differential nonlinearity	Worst-case deviation of any step from ideal	-1		4	LSB
Input range			0		1.4	V
Driving source impedance					100	Ω
FCLK	Clock rate	Successive approximation input clock rate		10		MHz
Input capacitance				12		pF
Input impedance		ADC Pin 57		2.15		kΩ
		ADC Pin 58		0.7		
		ADC Pin 59		2.12		
		ADC Pin 60		1.17		
Number of channels			4			
F <sub>sample</sub>	Sampling rate of each pin			62.5		KSPS
F <sub>input_max</sub>	Maximum input signal frequency				31	kHz
SINAD	Signal-to-noise and distortion	Input frequency DC to 300 Hz and 1.4 V <sub>pp</sub> sine wave input	55	60		dB
I <sub>active</sub>	Active supply current	Average for analog-to-digital during conversion without reference current		1.5		mA
I <sub>PD</sub>	Power-down supply current for core supply	Total for analog-to-digital when not active (this must be the SoC level test)		1		μA
Absolute offset error		FCLK = 10 MHz		±2		mV
Gain error				±2%		
V <sub>ref</sub>	ADC reference voltage			1.467		V

### 5.17.6.7 Camera Parallel Port

The fast camera parallel port interfaces with a variety of external image sensors, stores the image data in a FIFO, and generates DMA requests. The camera parallel port supports 8 bits.

Figure 5-21 shows the timing diagram for the camera parallel port.

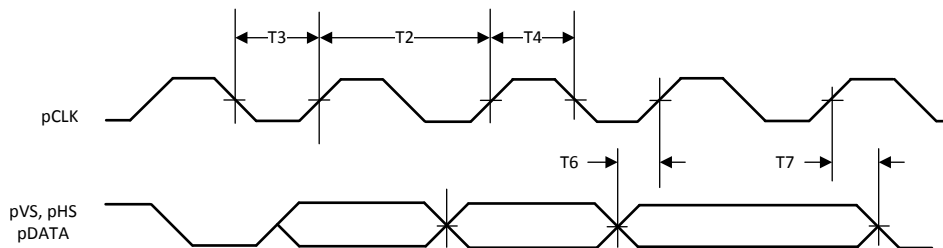


Figure 5-21. Camera Parallel Port Timing Diagram

Table 5-31 lists the timing parameters for the camera parallel port.

Table 5-31. Camera Parallel Port Timing Parameters

PARAMETER NUMBER			MIN	MAX	UNIT
	pCLK	Clock frequency		2	MHz
T2	$T_{clk}$	Clock period		$1/pCLK$	ns
T3	$t_{LP}$	Clock low period		$T_{clk}/2$	ns
T4	$t_{HT}$	Clock high period		$T_{clk}/2$	ns
T6	$t_{IS}$	RX data setup time		2	ns
T7	$t_{IH}$	RX data hold time		2	ns
	D	Duty cycle	45%	55%	

### 5.17.6.8 UART

The CC3235x device includes two UARTs with the following features:

- Programmable baud-rate generator allows speeds up to 3 Mbps
- Separate 16-bit x 8-bit TX and RX FIFOs to reduce CPU interrupt service loading
- Programmable FIFO length, including a 1-byte-deep operation providing conventional double-buffered interface
- FIFO trigger levels of 1/8, 1/4, 1/2, 3/4, and 7/8
- Standard asynchronous communication bits for start, stop, and parity
- Generation and detection of line-breaks
- Fully programmable serial interface characteristics:
  - 5, 6, 7, or 8 data bits
  - Generation and detection of even, odd, stick, or no-parity bits
  - Generation of 1 or 2 stop-bits
- RTS and CTS hardware flow support
- Standard FIFO-level and End-of-Transmission interrupts

- Efficient transfers using  $\mu$ DMA:
  - Separate channels for transmit and receive
  - Receive single request asserted when data is in the FIFO; burst request asserted at programmed FIFO level
  - Transmit single request asserted when there is space in the FIFO; burst request asserted at programmed FIFO level
- System clock is used to generate the baud clock.

#### 5.17.6.9 SD Host

The CC3235x device provides an interface between a local host (LH), such as an MCU and an SD memory card, and handles SD transactions with minimal LH intervention.

The SD host does the following:

- Provides SD card access in 1-bit mode
- Deals with SD protocol at the transmission level
- Handles data packing
- Adds cyclic redundancy checks (CRC)
- Start and end bit
- Checks for syntactical correctness

The application interface sends every SD command and either polls for the status of the adapter or waits for an interrupt request. The result is then sent back to the application interface in case of exceptions or to warn of end-of-operation. The controller can be configured to generate DMA requests and work with minimum CPU intervention. Given the nature of integration of this peripheral on the CC3235x platform, TI recommends that developers use peripheral library APIs to control and operate the block. This section emphasizes understanding the SD host APIs provided in the peripheral library of the CC3235x Software Development Kit (SDK).

The SD Host features are as follows:

- Full compliance with SD command and response sets, as defined in the SD memory card
  - Specifications, v2.0
  - Includes high-capacity (size >2 GB) HC and SD cards
- Flexible architecture allows support for new command structure
- 1-bit transfer mode specifications for SD cards
- Built-in 1024-byte buffer for read or write
  - 512-byte buffer for both transmit and receive
  - Each buffer is 32-bits wide by 128-words deep
- 32-bit-wide access bus to maximize bus throughput
- Single interrupt line for multiple interrupt source events
- Two slave DMA channels (1 for TX, 1 for RX)
- Programmable clock generation
- Integrates an internal transceiver that allows a direct connection to the SD card without external transceiver
- Supports configurable busy and response timeout
- Support for a wide range of card clock frequency with odd and even clock ratio
- Maximum frequency supported is 24 MHz

### 5.17.6.10 Timers

Programmable timers can be used to count or time external events that drive the timer input pins. The CC3235x general-purpose timer module (GPTM) contains 16- or 32-bit GPTM blocks. Each 16- or 32-bit GPTM block provides two 16-bit timers or counters (referred to as Timer A and Timer B) that can be configured to operate independently as timers or event counters, or they can be concatenated to operate as one 32-bit timer. Timers can also be used to trigger  $\mu$ DMA transfers.

The GPTM contains four 16- or 32-bit GPTM blocks with the following functional options:

- Operating modes:
  - 16- or 32-bit programmable one-shot timer
  - 16- or 32-bit programmable periodic timer
  - 16-bit general-purpose timer with an 8-bit prescaler
  - 16-bit input-edge count or time-capture modes with an 8-bit prescaler
  - 16-bit PWM mode with an 8-bit prescaler and software-programmable output inversion of the PWM signal
- Counts up or counts down
- Sixteen 16- or 32-bit capture compare pins (CCP)
- User-enabled stalling when the microcontroller asserts CPU Halt flag during debug
- Ability to determine the elapsed time between the assertion of the timer interrupt and entry into the interrupt service routine
- Efficient transfers using micro direct memory access controller ( $\mu$ DMA):
  - Dedicated channel for each timer
  - Burst request generated on timer interrupt
- Runs from system clock (80 MHz)

## 6 Detailed Description

### 6.1 Overview

The CC3235x wireless MCU family has a rich set of peripherals for diverse application requirements. This section briefly highlights the internal details of the CC3235x devices and offers suggestions for application configurations.

### 6.2 Arm® Cortex®-M4 Processor Core Subsystem

The high-performance Arm® Cortex®-M4 processor provides a cost-conscious platform that meets the needs of minimal memory implementation, reduced pin count, and low power consumption, while delivering outstanding computational performance and exceptional system response to interrupts.

- The Arm Cortex-M4 core has low-latency interrupt processing with the following features:
  - A 32-bit Arm® Thumb® instruction set optimized for embedded applications
  - Handler and thread modes
  - Low-latency interrupt handling by automatic processor state saving and restoration during entry and exit
  - Support for Armv6 unaligned accesses
- Nested vectored interrupt controller (NVIC) closely integrated with the processor core to achieve low-latency interrupt processing. The NVIC includes the following features:
  - Bits of priority configurable from 3 to 8
  - Dynamic reprioritization of interrupts
  - Priority grouping that enables selection of preempting interrupt levels and nonpreempting interrupt levels
  - Support for tail-chaining and late arrival of interrupts, which enables back-to-back interrupt processing without the overhead of state saving and restoration between interrupts
  - Processor state automatically saved on interrupt entry and restored on interrupt exit with no instruction overhead
  - Wake-up interrupt controller (WIC) providing ultra-low-power sleep mode support
- Bus interfaces:
  - Advanced high-performance bus (AHB-Lite) interfaces: system bus interfaces
  - Bit-band support for memory and select peripheral that includes atomic bit-band write and read operations
- Cost-conscious debug solution featuring:
  - Debug access to all memory and registers in the system, including access to memory-mapped devices, access to internal core registers when the core is halted, and access to debug control registers even while SYSRESETn is asserted
  - Serial wire debug port (SW-DP) or serial wire JTAG debug port (SWJ-DP) debug access
  - Flash patch and breakpoint (FPB) unit to implement breakpoints and code patches

## 6.3 Wi-Fi® Network Processor Subsystem

The Wi-Fi network processor subsystem includes a dedicated Arm MCU to completely offload the host MCU along with an 802.11a/b/g/n radio, baseband, and MAC with a powerful crypto engine for a fast, secure WLAN and Internet connections with 256-bit encryption. The CC3235x devices support station, AP, and Wi-Fi Direct modes. The device also supports WPA2 personal and enterprise security and WPS 2.0. The Wi-Fi network processor includes an embedded IPv6, IPv4 TCP/IP stack, TLS stack and network applications such as HTTPS server.

### 6.3.1 WLAN

The WLAN features are as follows:

- 802.11a/b/g/n integrated radio, modem, and MAC supporting WLAN communication as a BSS station, AP, Wi-Fi Direct client, and group owner with CCK and OFDM rates in the 2.4 GHz band (channels 1 through 13), and the 5 GHz 20-MHz BW U-NII bands (U-NII-1, U-NII-2A, U-NII-2C, and U-NII-3).

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#### NOTE

802.11n is supported only in Wi-Fi station and Wi-Fi direct

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- The automatically calibrated radio with a single-ended 50-Ω interface enables easy connection to the antenna without requiring expertise in radio circuit design.
- Advanced connection manager with multiple user-configurable profiles stored in serial flash allows automatic fast connection to an access point without user or host intervention.
- Supports all common Wi-Fi security modes for personal and enterprise networks with on-chip security accelerators, including: WEP, WPA/WPA2 PSK, WPA2 Enterprise (802.1x).
- Smart provisioning options deeply integrated within the device providing a comprehensive end-to-end solution. With elaborate events notification to the host, enabling the application to control the provisioning decision flow. The wide variety of Wi-Fi provisioning methods include:
  - Access Point with HTTP server
  - WPS - Wi-Fi Protected Setup, supporting both push button and pin code options.
  - SmartConfig™ Technology: TI proprietary, easy to use, one-step, one-time process used to connect a CC3235x-enabled device to the home wireless network.
- 802.11 transceiver mode allows transmitting and receiving of proprietary data through a socket. The 802.11 transceiver mode provides the option to select the working channel, rate, and transmitted power. The receiver mode works with the filtering options.
- Antenna selection for best connection
- BLE/2.4 GHz radio coexistence mechanism to avoid interference

### 6.3.2 Network Stack

The Network Stack features are as follows:

- Integrated IPv4, IPv6 TCP/IP stack with BSD socket APIs for simple Internet connectivity with any MCU, microprocessor, or ASIC

#### NOTE

Not all APIs are 100% BSD compliant. Not all BSD APIs are supported.

- Support of 16 simultaneous TCP, UDP, RAW, SSL\TLS sockets
- Built-in network protocols:
  - Static IP, LLA, DHCPv4, DHCPv6 with DAD and stateless autoconfiguration
  - ARP, ICMPv4, IGMP, ICMPv6, MLD, ND
  - DNS client for easy connection to the local network and the Internet
- Built-in network applications and utilities:
  - HTTP/HTTPS
    - Web page content stored on serial flash
    - RESTful APIs for setting and configuring application content
    - Dynamic user callbacks
  - Service discovery: Multicast DNS service discovery lets a client advertise its service without a centralized server. After connecting to the access point, the CC3235x device provides critical information, such as device name, IP, vendor, and port number.
  - DHCP server
  - Ping

Table 6-1 describes the NWP features.

**Table 6-1. NWP Features**

Feature	Description
Wi-Fi standards	802.11a/b/g/n station 802.11a/b/g AP supporting up to four stations Wi-Fi Direct client and group owner
Wi-Fi channels	2.4 GHz ISM and 5 GHz U-NII Channels
Channel Bandwidth	20 MHz
Wi-Fi security	WEP, WPA/WPA2 PSK, WPA2 enterprise (802.1x)
Wi-Fi provisioning	SmartConfig technology, Wi-Fi protected setup (WPS2), AP mode with internal HTTP web server
IP protocols	IPv4/IPv6
IP addressing	Static IP, LLA, DHCPv4, DHCPv6 with DAD
Cross layer	ARP, ICMPv4, IGMP, ICMPv6, MLD, NDP
Transport	UDP, TCP SSLv3.0/TLSv1.0/TLSv1.1/TLSv1.2 RAW
Network applications and utilities	Ping HTTP/HTTPS web server mDNS DNS-SD DHCP server
Host interface	UART/SPI

**Table 6-1. NWP Features (continued)**

Feature	Description
Security	Device identity Trusted root-certificate catalog TI root-of-trust public key The CC3235S and CC3235SF variants also support: <ul style="list-style-type: none"> <li>• Secure key storage</li> <li>• Online certificate status protocol (OCSP)</li> <li>• Certificate signing request (CSR)</li> <li>• Unique per device Key-Pair</li> <li>• File system security</li> <li>• Software tamper detection</li> <li>• Cloning protection</li> <li>• Secure boot</li> <li>• Validate the integrity and authenticity of the run-time binary during boot</li> <li>• Initial secure programming</li> <li>• Debug security</li> <li>• JTAG and debug</li> </ul>
Power management	Enhanced power policy management uses 802.11 power save and deep-sleep power modes
Other	Transceiver Programmable RX filters with event-trigger mechanism Rx Metrics for tracking the surrounding RF environment

## 6.4 Security

The SimpleLink™ Wi-Fi® CC3235x Internet-on-a chip device enhances the security capabilities available for development of IoT devices, while completely offloading these activities from the MCU to the networking subsystem. The security capabilities include the following key features:

### Wi-Fi and Internet Security:

- Personal and enterprise Wi-Fi security
  - Personal standards
    - AES (WPA2-PSK)
    - TKIP (WPA-PSK)
    - WEP
  - Enterprise standards
    - EAP Fast
    - EAP PEAPv0/1
    - EAP PEAPv0 TLS
    - EAP PEAPv1 TLS EAP LS
    - EAP TLS
    - EAP TTLS TLS
    - EAP TTLS MSCHAPv2



- Secure sockets
  - Protocol versions: OCSP, SSL v3, TLS 1.0, TLS 1.1, TLS 1.2
  - Powerful crypto engine for fast, secure Wi-Fi and internet connections with 256-bit AES encryption for TLS and SSL connections
  - Ciphers suites
    - SL\_SEC\_MASK\_SSL\_RSA\_WITH\_RC4\_128\_SHA
    - SL\_SEC\_MASK\_SSL\_RSA\_WITH\_RC4\_128\_MD5
    - SL\_SEC\_MASK\_TLS\_RSA\_WITH\_AES\_256\_CBC\_SHA
    - SL\_SEC\_MASK\_TLS\_DHE\_RSA\_WITH\_AES\_256\_CBC\_SHA
    - SL\_SEC\_MASK\_TLS\_ECDHE\_RSA\_WITH\_AES\_256\_CBC\_SHA
    - SL\_SEC\_MASK\_TLS\_ECDHE\_RSA\_WITH\_RC4\_128\_SHA
    - SL\_SEC\_MASK\_TLS\_RSA\_WITH\_AES\_128\_CBC\_SHA256
    - SL\_SEC\_MASK\_TLS\_RSA\_WITH\_AES\_256\_CBC\_SHA256
    - SL\_SEC\_MASK\_TLS\_ECDHE\_RSA\_WITH\_AES\_128\_CBC\_SHA256
    - SL\_SEC\_MASK\_TLS\_ECDHE\_ECDSA\_WITH\_AES\_128\_CBC\_SHA256
    - SL\_SEC\_MASK\_TLS\_ECDHE\_ECDSA\_WITH\_AES\_256\_CBC\_SHA
    - SL\_SEC\_MASK\_TLS\_RSA\_WITH\_AES\_128\_GCM\_SHA256
    - SL\_SEC\_MASK\_TLS\_RSA\_WITH\_AES\_256\_GCM\_SHA384
    - SL\_SEC\_MASK\_TLS\_DHE\_RSA\_WITH\_AES\_128\_GCM\_SHA256
    - SL\_SEC\_MASK\_TLS\_DHE\_RSA\_WITH\_AES\_256\_GCM\_SHA384
    - SL\_SEC\_MASK\_TLS\_ECDHE\_RSA\_WITH\_AES\_128\_GCM\_SHA256
    - SL\_SEC\_MASK\_TLS\_ECDHE\_RSA\_WITH\_AES\_256\_GCM\_SHA384
    - SL\_SEC\_MASK\_TLS\_ECDHE\_ECDSA\_WITH\_AES\_128\_GCM\_SHA256
    - SL\_SEC\_MASK\_TLS\_ECDHE\_ECDSA\_WITH\_AES\_256\_GCM\_SHA384
    - SL\_SEC\_MASK\_TLS\_ECDHE\_ECDSA\_WITH\_CHACHA20\_POLY1305\_SHA256
    - SL\_SEC\_MASK\_TLS\_ECDHE\_RSA\_WITH\_CHACHA20\_POLY1305\_SHA256
    - SL\_SEC\_MASK\_TLS\_DHE\_RSA\_WITH\_CHACHA20\_POLY1305\_SHA256
  - Server authentication
  - Client authentication
  - Domain name verification
  - Runtime socket upgrade to secure socket – STARTTLS
- Secure HTTP server (HTTPS)
- Trusted root-certificate catalog – Verifies that the CA used by the application is trusted and known secure content delivery
- TI root-of-trust public key – Hardware-based mechanism that allows authenticating TI as the genuine origin of a given content using asymmetric keys
- Secure content delivery – Allows encrypted file transfer to the system using asymmetric keys created by the device

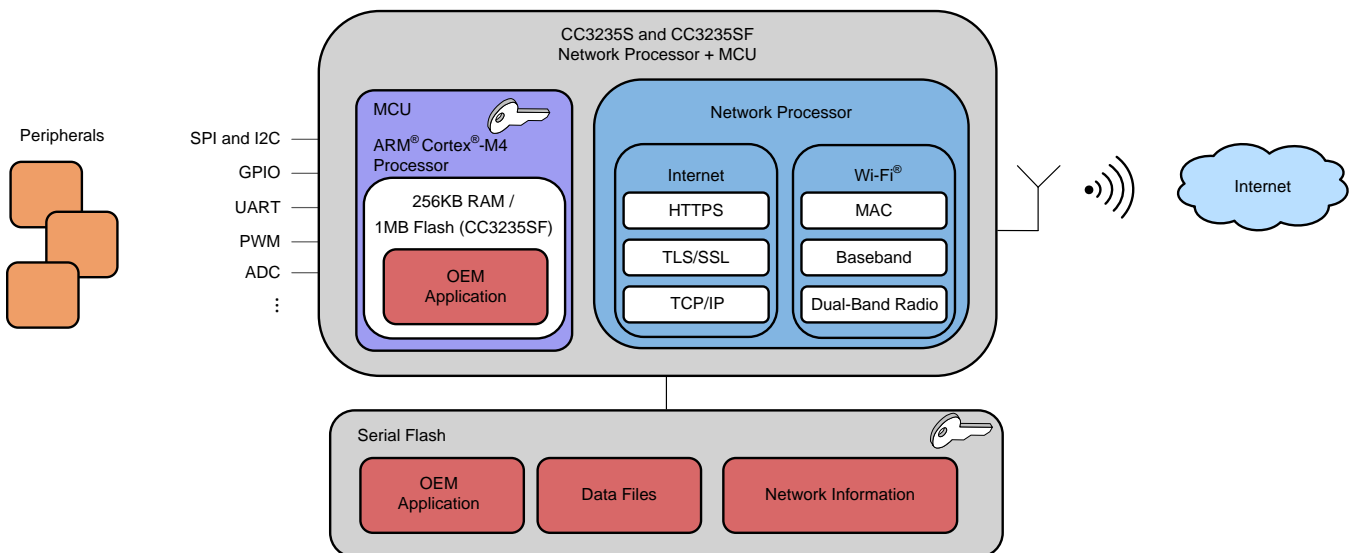
### Code and Data Security:

- Network passwords and certificates are encrypted and signed.
- Cloning protection – Application and data files are encrypted by a unique key per device.
- Access control – Access to application and data files only by using a token provided in file creation time. If an unauthorized access is detected, a tamper protection lock down mechanism takes effect.
- Secured boot – Authentication of the application image on every boot
- Code and data encryption – User application and data files can be encrypted in the serial flash
- Code and data authentication – User Application and data files are authenticated with a public key certificate
- Offloaded crypto library for asymmetric keys, including the ability to create key-pair, sign and verify data buffer
- Recovery mechanism

### Device Security:

- Separate execution environments – Application processor and network processor run on separate Arm cores
- Initial secure programming – Allows for keeping the content confidential on the production line
- Debug security
  - JTAG lock
  - Debug ports lock
- True random number generator

Figure 6-1 shows the high-level structure of the CC3235S and CC3235SF devices. The application image, user data, and network information files (passwords, certificates) are encrypted using a device-specific key.



**Figure 6-1. CC3235S and CC3235SF High-Level Structure**

## 6.5 FIPS 140-2 Level 1 Certification

The Federal Information Processing Standard (FIPS) Publication 140-2 is a U.S. government computer security standard. It is commonly referred to as FIPS 140-2 and is used to accredit the design and implementation of cryptographic modules. A cryptographic module within a security system is necessary to maintain the confidentiality and integrity of the information protected by the device.

The security engines of the CC3235x device are FIPS validated for FIPS 140-2 level 1 certification <sup>(1)</sup>. This certification involves testing the device for all areas related to the secure design and implementation of the cryptographic modules and covers topics such as: cryptographic specifications, ports and interfaces, a finite state model for the cryptographic module, the operational environment of the module, and how cryptographic keys are managed.

## 6.6 Power-Management Subsystem

The CC3235x power-management subsystem contains DC/DC converters to accommodate the different voltage or current requirements of the system.

- Digital DC/DC (Pin 44)
  - Input:  $V_{BAT}$  wide voltage (2.1 to 3.6 V)
- ANA1 DC/DC (Pin 37)
  - Input:  $V_{BAT}$  wide voltage (2.1 to 3.6 V)
- PA DC/DC (Pin 39)
  - Input:  $V_{BAT}$  wide voltage (2.1 to 3.6 V)
- ANA2 DC/DC (Pin 47, CC3235SF only)
  - Input:  $V_{BAT}$  wide voltage (2.1 to 3.6 V)

The CC3235x device is a single-chip WLAN radio solution used on an embedded system with a wide-voltage supply range. The internal power management, including DC/DC converters and LDOs, generates all of the voltages required for the device to operate from a wide variety of input sources.

## 6.7 Low-Power Operating Mode

From a power-management perspective, the CC3235x device comprises the following two independent subsystems:

- Arm<sup>®</sup> Cortex<sup>®</sup>-M4 application processor subsystem
- Networking subsystem

Each subsystem operates in one of several power states.

The Arm<sup>®</sup> Cortex<sup>®</sup>-M4 application processor runs the user application loaded from an external serial flash, or internal flash (in CC3235SF). The networking subsystem runs preprogrammed TCP/IP and Wi-Fi data link layer functions.

(1) For exact status of FIPS certification for a specific part number, please refer to <https://csrc.nist.gov/publications/fips>.

The user program controls the power state of the application processor subsystem. The application processor can be in one of the five modes described in [Table 6-2](#).

**Table 6-2. User Program Modes**

APPLICATION PROCESSOR (MCU) MODE <sup>(1)</sup>	DESCRIPTION
MCU active mode	MCU executing code at a state rate of 80 MHz
MCU sleep mode	The MCU clocks are gated off in sleep mode and the entire state of the device is retained. Sleep mode offers instant wakeup. The MCU can be configured to wake up by an internal fast timer or by activity from any GPIO line or peripheral.
MCU LPDS mode	State information is lost and only certain MCU-specific register configurations are retained. The MCU can wake up from external events or by using an internal timer. (The wake-up time is less than 3 ms.) Certain parts of memory can be retained while the MCU is in LPDS mode. The amount of memory retained is configurable. Users can choose to preserve code and the MCU-specific setting. The MCU can be configured to wake up using the RTC timer or by an external event on specific GPIOs as the wake-up source.
MCU hibernate mode	The lowest power mode in which all digital logic is power-gated. Only a small section of the logic directly powered by the input supply is retained. The RTC continues running and the MCU supports wakeup from an external event or from an RTC timer expiry. Wake-up time is longer than LPDS mode at about 15 ms plus the time to load the application from serial flash, which varies according to code size. In this mode, the MCU can be configured to wake up using the RTC timer or external event on a GPIO.
MCU shutdown mode	The lowest power mode system-wide. All device logics are off, including the RTC. The wake-up time in this mode is longer than hibernate at about 1.1 s. To enter or exit the shutdown mode, the state of the nRESET line is changed (low to shut down, high to turn on).

(1) Modes are listed in order of power consumption, with highest power modes listed first.

The NWP can be active or in LPDS mode and takes care of its own mode transitions. When there is no network activity, the NWP sleeps most of the time and wakes up only for beacon reception (see [Table 6-3](#)).

**Table 6-3. Networking Subsystem Modes**

NETWORK PROCESSOR MODE	DESCRIPTION
Network active mode (processing layer 3, 2, and 1)	Transmitting or receiving IP protocol packets
Network active mode (processing layer 2 and 1)	Transmitting or receiving MAC management frames; IP processing is not required
Network active listen mode	Special power-optimized active mode for receiving beacon frames (no other frames are supported)
Network connected Idle	A composite mode that implements 802.11 infrastructure power-save operation. The CC3235x NWP automatically enters LPDS mode between beacons and then wakes into active listen mode to receive a beacon and determine if there is pending traffic at the AP. If not, the NWP returns to LPDS mode and the cycle repeats.  Advanced features of long sleep interval and IoT low power for extending LPDS time for up to 22 seconds while maintaining Wi-Fi connection is supported in this mode.
Network LPDS mode	Low-power state between beacons in which the state is retained by the NWP, allowing for a rapid wake up
Network disabled	The network is disabled

The operation of the application and network processor ensures that the device remains in the lowest power mode most of the time to preserve battery life.

The following examples show the use of the power modes in applications:

- A product that is continuously connected to the network in the 802.11 infrastructure power-save mode but sends and receives little data spends most of the time in connected idle, which is a composite of receiving a beacon frame and waiting for the next beacon.
- A product that is not continuously connected to the network but instead wakes up periodically (for example, every 10 minutes) to send data, spends most of the time in hibernate mode, jumping briefly to active mode to transmit data.

## 6.8 Memory

### 6.8.1 External Memory Requirements

The CC3235x device maintains a proprietary file system on the serial flash. The CC3235x file system stores the MCU binary, service pack file, system files, configuration files, certificate files, web page files, and user files. By using a format command through the API, users can provide the total size allocated for the file system. The starting address of the file system cannot be set and is always at the beginning of the serial flash. The applications microcontroller must access the serial flash memory area allocated to the file system directly through the CC3235x file system. The applications microcontroller must not access the serial flash memory area directly.

The file system manages the allocation of serial flash blocks for stored files according to download order, which means that the location of a specific file is not fixed in all systems. Files are stored on serial flash using human-readable filenames rather than file IDs. The file system API works using plain text, and file encryption and decryption is invisible to the user. Encrypted files can be accessed only through the file system.

All file types can have a maximum of 100 supported files in the file system. All files are stored in 4-KB blocks and thus use a minimum of 4KB of flash space. Fail-safe files require twice the original size and use a minimum of 8KB. Encrypted files are counted as fail-safe in terms of space. The maximum file size is 1MB.

Table 6-4 lists the minimum required memory consumption under the following assumptions:

- System files in use consume 64 blocks (256KB).
- Vendor files are not taken into account.
- MCU code is taken as the maximal possible size for the CC3235 with fail-safe enabled to account for future updates, such as through OTA.
- Gang image:
  - Storage for the gang image is rounded up to 32 blocks (meaning 128-KB resolution).
  - Gang image size depends on the actual content size of all components. Additionally, the image should be 128KB aligned so unaligned memory is considered lost. Service pack, system files, and the 128KB aligned memory are assumed to occupy 256KB.
- All calculations consider that the restore-to-default is enabled.

**Table 6-4. Recommended Flash Size**

ITEM	CC3235S (KB)	CC3235SF (KB)
File system allocation table	20	20
System and configuration files <sup>(1)</sup>	256	256
Service pack <sup>(1)</sup>	264	264
MCU Code <sup>(1)</sup>	512	2048

(1) Including fail-safe.

**Table 6-4. Recommended Flash Size (continued)**

ITEM	CC3235S (KB)	CC3235SF (KB)
Gang image size	256 + MCU	256 + MCU
Total	1308 + MCU	2844 + MCU
Minimal flash size <sup>(2)</sup>	16 MBit	32 MBit
Recommended flash size <sup>(2)</sup>	16 MBit	32 MBit

(2) For maximum MCU size.

**NOTE**

The maximum supported serial flash size is 32MB (256Mb) (see [Using Serial Flash on CC3135/CC3235 SimpleLink™ Wi-Fi® and Internet-of-Things Devices](#)).

**6.8.2 Internal Memory**

The CC3235x device includes on-chip SRAM to which application programs are downloaded and executed. The application developer must share the SRAM for code and data. The micro direct memory access ( $\mu$ DMA) controller can transfer data to and from SRAM and various peripherals. The CC3235x ROM holds the rich set of peripheral drivers, which saves SRAM space. For more information on drivers, see the CC3235x API list.

**6.8.2.1 SRAM**

The CC3235x family provides 256KB of on-chip SRAM. Internal RAM is capable of selective retention during LPDS mode. This internal SRAM is at offset 0x2000 0000 of the device memory map.

Use the  $\mu$ DMA controller to transfer data to and from the SRAM.

When the device enters low-power mode, the application developer can choose to retain a section of memory based on need. Retaining the memory during low-power mode provides a faster wakeup. The application developer can choose the amount of memory to retain in multiples of 64KB. For more information, see the API guide.

**6.8.2.2 ROM**

The internal zero-wait-state ROM of the CC3235x device is at address 0x0000 0000 of the device memory and is programmed with the following components:

- Bootloader
- Peripheral driver library (DriverLib) release for product-specific peripherals and interfaces

The bootloader is used as an initial program loader (when the serial flash memory is empty). The CC3235x DriverLib software library controls on-chip peripherals with a bootloader capability. The library performs peripheral initialization and control functions, with a choice of polled or interrupt-driven peripheral support. The DriverLib APIs in ROM can be called by applications to reduce flash memory requirements and free the flash memory for other purposes.

**6.8.2.3 Flash Memory**

The CC3235SF device comes with an on-chip flash memory of 1MB that allows application code to execute in place while freeing SRAM exclusively for read-write data. The flash memory is used for code and constant data sections and is directly attached to the icode/dcode bus of the Arm Cortex-M4 core. A 128-bit-wide instruction prefetch buffer allows maintenance of maximum performance for linear code or loops that fit inside the buffer.

The flash memory is organized as 2KB sectors that can be independently erased. Reads and writes can be performed at word (32-bit) level.

### 6.8.2.4 Memory Map

Table 6-5 describes the various MCU peripherals and how they are mapped to the processor memory. For more information on peripherals, see the API document.

**Table 6-5. Memory Map**

START ADDRESS	END ADDRESS	DESCRIPTION	COMMENT
0x0000 0000	0x0007 FFFF	On-chip ROM (bootloader + DriverLib)	
0x0100 0000	0x010F FFFF	On-chip flash (for user application code)	CC3235SF device only
0x2000 0000	0x2003 FFFF	Bit-banded on-chip SRAM	
0x2200 0000	0x23FF FFFF	Bit-band alias of 0x2000 0000 to 0x200F FFFF	
0x4000 0000	0x4000 0FFF	Watchdog timer A0	
0x4000 4000	0x4000 4FFF	GPIO port A0	
0x4000 5000	0x4000 5FFF	GPIO port A1	
0x4000 6000	0x4000 6FFF	GPIO port A2	
0x4000 7000	0x4000 7FFF	GPIO port A3	
0x4000 C000	0x4000 CFFF	UART A0	
0x4000 D000	0x4000 DFFF	UART A1	
0x4002 0000	0x4000 07FF	I <sup>2</sup> C A0 (master)	
0x4002 0800	0x4002 0FFF	I <sup>2</sup> C A0 (slave)	
0x4002 4000	0x4002 4FFF	GPIO group 4	
0x4003 0000	0x4003 0FFF	General-purpose timer A0	
0x4003 1000	0x4003 1FFF	General-purpose timer A1	
0x4003 2000	0x4003 2FFF	General-purpose timer A2	
0x4003 3000	0x4003 3FFF	General-purpose timer A3	
0x400F 7000	0x400F 7FFF	Configuration registers	
0x400F E000	0x400F EFFF	System control	
0x400F F000	0x400F FFFF	μDMA	
0x4200 0000	0x43FF FFFF	Bit band alias of 0x4000 0000 to 0x400F FFFF	
0x4401 0000	0x4401 0FFF	SDIO master	
0x4401 8000	0x4401 8FFF	Camera Interface	
0x4401 C000	0x4401 DFFF	McASP	
0x4402 0000	0x4402 1FFF	SSPI	Used for external serial flash
0x4402 1000	0x4402 2FFF	GSPI	Used by application processor
0x4402 5000	0x4402 5FFF	MCU reset clock manager	
0x4402 6000	0x4402 6FFF	MCU configuration space	
0x4402 D000	0x4402 DFFF	Global power, reset, and clock manager (GPRCM)	
0x4402 E000	0x4402 EFFF	MCU shared configuration	
0x4402 F000	0x4402 FFFF	Hibernate configuration	
0x4403 0000	0x4403 FFFF	Crypto range (includes apertures for all crypto-related blocks as follows)	
0x4403 0000	0x4403 0FFF	DTHE registers and TCP checksum	
0x4403 5000	0x4403 5FFF	MD5/SHA	
0x4403 7000	0x4403 7FFF	AES	
0x4403 9000	0x4403 9FFF	DES	
0xE000 0000	0xE000 0FFF	Instrumentation trace Macrocell™	
0xE000 1000	0xE000 1FFF	Data watchpoint and trace (DWT)	
0xE000 2000	0xE000 2FFF	Flash patch and breakpoint (FPB)	
0xE000 E000	0xE000 EFFF	NVIC	
0xE004 0000	0xE004 0FFF	Trace port interface unit (TPIU)	

**Table 6-5. Memory Map (continued)**

START ADDRESS	END ADDRESS	DESCRIPTION	COMMENT
0xE004 1000	0xE004 1FFF	Reserved for embedded trace macrocell (ETM)	
0xE004 2000	0xE00F FFFF	Reserved	

## 6.9 Restoring Factory Default Configuration

The device has an internal recovery mechanism that allows rolling back the file system to its predefined factory image or restoring the factory default parameters of the device. The factory image is kept in a separate sector on the serial flash in a secure manner and cannot be accessed from the host processor. The following restore modes are supported:

- None – no factory restore settings
- Enable restore of factory default parameters
- Enable restore of factory image and factory default parameters

The restore process is performed by calling software APIs, or by pulling or forcing SOP[2:0] = 110 pins and toggling the nRESET pin from low to high.

The process is fail-safe and resumes operation if a power failure occurs before the restore is finished. The restore process typically takes about 8 seconds, depending on the attributes of the serial flash vendor.



## 6.10 Boot Modes

### 6.10.1 Boot Mode List

The CC3235x device implements a sense-on-power (SOP) scheme to determine the device operation mode.

SOP values are sensed from the device pin during power up. This encoding determines the boot flow. Before the device is taken out of reset, the SOP values are copied to a register and used to determine the device operation mode while powering up. These values determine the boot flow as well as the default mapping (to JTAG, SWD, UART0) for some of the pins. [Table 6-6](#) lists the pull configurations.

**Table 6-6. CC3235x Functional Configurations**

BOOT MODE NAME	SOP[2]	SOP[1]	SOP[0]	SOP MODE	COMMENT
UARTLOAD	Pullup	Pulldown	Pulldown	LDfrUART	Factory, lab flash, and SRAM loads through the UART. The device waits indefinitely for the UART to load code. The SOP bits then must be toggled to configure the device in functional mode. Also puts JTAG in 4-wire mode.
FUNCTIONAL_2WJ	Pulldown	Pulldown	Pullup	Fn2WJ	Functional development mode. In this mode, 2-pin SWD is available to the developer. TMS and TCK are available for debugger connection.
FUNCTIONAL_4WJ	Pulldown	Pulldown	Pulldown	Fn4WJ	Functional development mode. In this mode, 4-pin JTAG is available to the developer. TDI, TMS, TCK, and TDO are available for debugger connection.

**Table 6-6. CC3235x Functional Configurations (continued)**

BOOT MODE NAME	SOP[2]	SOP[1]	SOP[0]	SOP MODE	COMMENT
UARTLOAD_FUNCTIONAL_4WJ	Pulldown	Pullup	Pulldown	LDfrUART_Fn4WJ	Supports flash and SRAM load through UART and functional mode. The MCU bootloader tries to detect a UART break on UART receive line. If the break signal is present, the device enters the UARTLOAD mode, otherwise, the device enters the functional mode. TDI, TMS, TCK, and TDO are available for debugger connection.
RET_FACTORY_IMAGE	Pulldown	Pullup	Pullup	RetFactDef	When device reset is toggled, the MCU bootloader kick-starts the procedure to restore factory default images.

The recommended values of pull down resistors are 69.8-k $\Omega$  for SOP0 and SOP1 and 100-k $\Omega$  for SOP2. The application can use SOP2 for other functions after the device has powered up. To avoid spurious SOP values from being sensed at power up, TI strongly recommends using the SOP2 pin only for output signals. The SOP0 and SOP1 pins are used as 5 GHz control switch and are not available for other functions. Ensure the SOP pins are configured as shown in [Figure 7-7](#), this is the recommended configuration to ensure the RF Switch, SOP boot modes, and Factory restore process operates optimally without conflict.

## 6.11 Hostless Mode

The SimpleLink Wi-Fi CC3235 device incorporates a scripting ability that enables offloading of simple tasks from the host processor. Using simple and conditional scripts, repetitive tasks can be handled internally, which allows the host processor to remain in a low-power state. In some cases where the scripter is being used to send packets, it reduces code footprint and memory consumption. The *if-this-then-that* style conditioning can include anything from GPIO toggling to transmitting packets.

The conditional scripting abilities can be divided into conditions and actions. The conditions define when to trigger actions. Only one action can be defined per condition, but multiple instances of the same condition may be used, so in effect multiple actions can be defined for a single condition. In total, 16 condition and action pairs can be defined. The conditions can be simple, or complex using sub-conditions (using a combinatorial AND condition between them). The actions are divided into two types, those that can occur during runtime and those that can occur only during the initialization phase.

The following actions can only be performed when triggered by the pre-initialization condition:

- Set roles AP, station, P2P, and Tag modes
- Delete all stored profiles
- Set connection policy
- Hardware GPIO indication allows an I/O to be driven directly from the WLAN core hardware to indicate internal signaling

The following actions may be activated during runtime:

- Send transceiver packet
- Send UDP packet
- Send TCP packet
- Increment counter increments one of the user counters by 1
- Set counter allows setting a specific value to a counter
- Timer control
- Set GPIO allows GPIO output from the device using the internal networking core
- Enter Hibernate state

---

### NOTE

Consider the following limitations:

- Timing cannot be ensured when using the network scripter because some variable latency will apply depending on the utilization of the networking core.
  - The scripter is limited to 16 pairs of conditions and reactions.
  - Both timers and counters are limited to 8 instances each. Timers are limited to a resolution of 1 second. Counters are 32 bits wide.
  - Packet length is limited to the size of one packet and the number of possible packet tokens is limited to 8.
-

## 7 Applications, Implementation, and Layout

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### NOTE

Information in the following Applications section is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

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### 7.1 Application Information

#### 7.1.1 BLE/2.4 GHz Radio Coexistence

The CC3235x device is designed to support BLE/2.4 GHz radio coexistence. Because WLAN is inherently more tolerant to time-domain disturbances, the coexistence mechanism gives priority to the Bluetooth® low energy entity over the WLAN. Bluetooth® low energy operates in the 2.4 GHz band, therefore the coexistence mechanism does not affect the 5 GHz band. The CC3235x device can operate normally on the 5 GHz band, while the Bluetooth® low energy works on the 2.4 GHz band without mutual interference.

The following coexistence modes can be configured by the user:

- Off mode or intrinsic mode
  - No BLE/2.4 GHz radio coexistence, or no synchronization between WLAN and Bluetooth® low energy—in case Bluetooth® low energy exists in this mode, collisions can randomly occur.
- Time Division Multiplexing (TDM, Single Antenna)
  - 2.4 GHz Wi-Fi band (see [Figure 7-1](#))  
In this mode, the two entities share the antenna through an RF switch using two GPIOs (one input and one output from the WLAN perspective).
  - 5 GHz Wi-Fi band (see [Figure 7-2](#))  
In this mode, the WLAN operates on the 5 GHz band and Bluetooth® low energy operates on the 2.4 GHz band. A 2.4- or 5 GHz diplexer is required for sharing the single antenna.
- Time Division Multiplexing (TDM, Dual Antenna)
  - 2.4 GHz Wi-Fi Band (see [Figure 7-3](#))  
In this mode, the two entities have separate antennas. No RF switch is required and only a single GPIO (one input from the WLAN perspective).
  - 5 GHz Wi-Fi band (see [Figure 7-4](#))  
In this mode, the WLAN operates on the 5 GHz band and Bluetooth® low energy operates on the 2.4 GHz band. No diplexer is required for the dual-antenna solution.

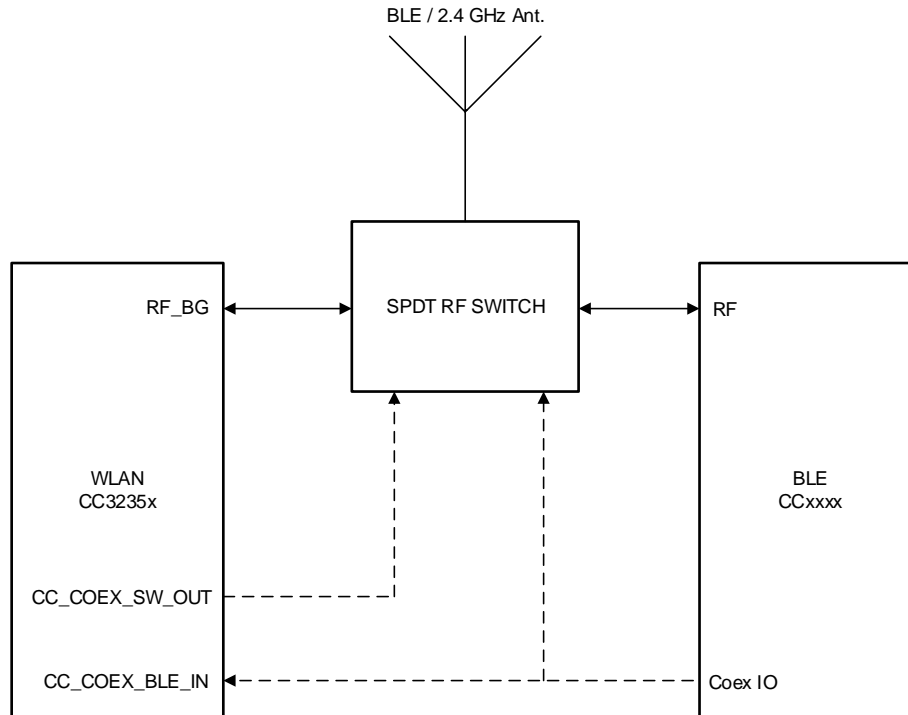


Figure 7-1. 2.4 GHz, Single-Antenna Coexistence Mode Block Diagram

Figure 7-2 shows the single antenna implementation of a complete Bluetooth® low energy and WLAN coexistence network with the WLAN operating on either a 2.4- or a 5 GHz band. The SOP lines control the 5 GHz switch. The Coex switch is controlled by a GPIO signal from the BLE device and a GPIO signal from the CC3235x device.

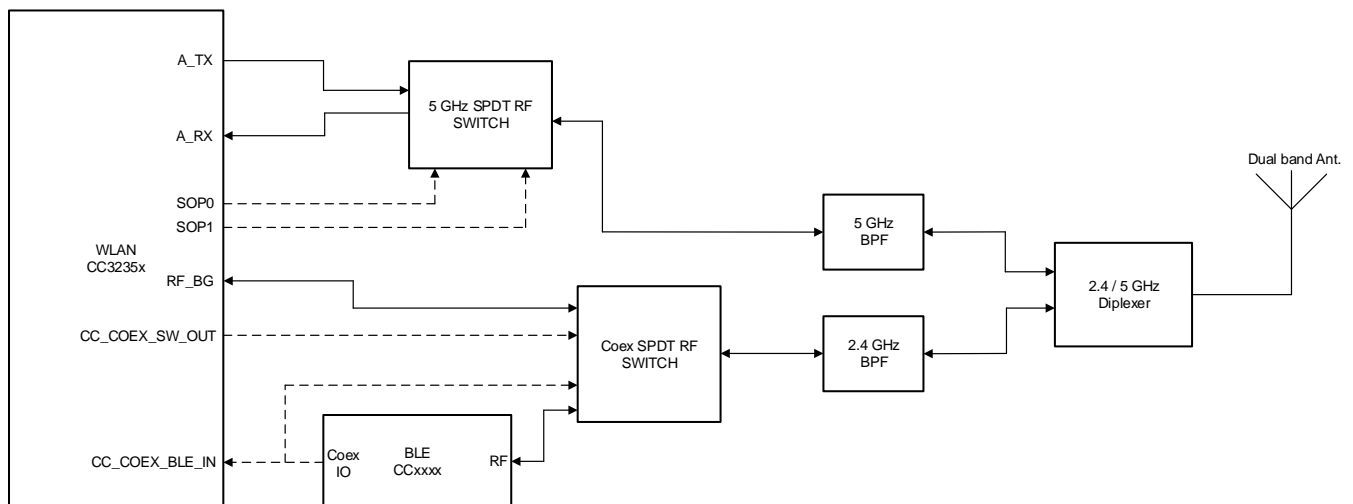


Figure 7-2. Single Antenna Coexistence Solution with 5 GHz Wi-Fi

Figure 7-3 shows the dual antenna implementation of a complete Bluetooth® low energy and WLAN coexistence network with the WLAN operating on either a 2.4- or a 5 GHz band. Note in this implementation no Coex switch is required and only a single GPIO from the BLE device to the CC3235x device is required.

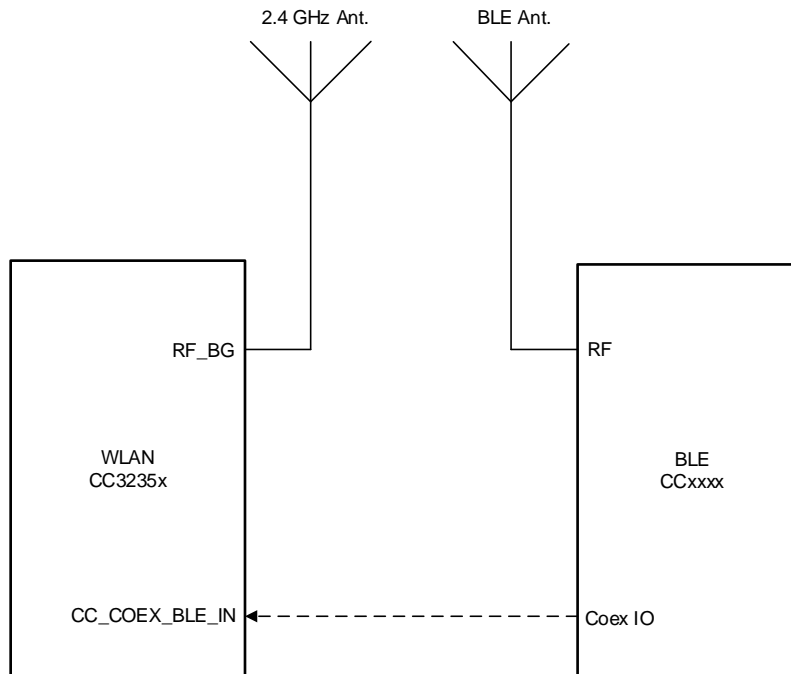


Figure 7-3. Dual-Antenna Coexistence Mode Block Diagram

Figure 7-4 shows the dual antenna implementation of a complete Bluetooth® low energy and WLAN coexistence network with the WLAN operating on either a 2.4- or a 5 GHz band. In this case the 2.4 GHz and 5 GHz Wi-Fi share an antenna and the BLE has it's own dedicated antenna. The SOP lines control the 5 GHz switch. Note in this implementation no Coex switch is required and only a single GPIO from the BLE device to the CC3235x device is required.

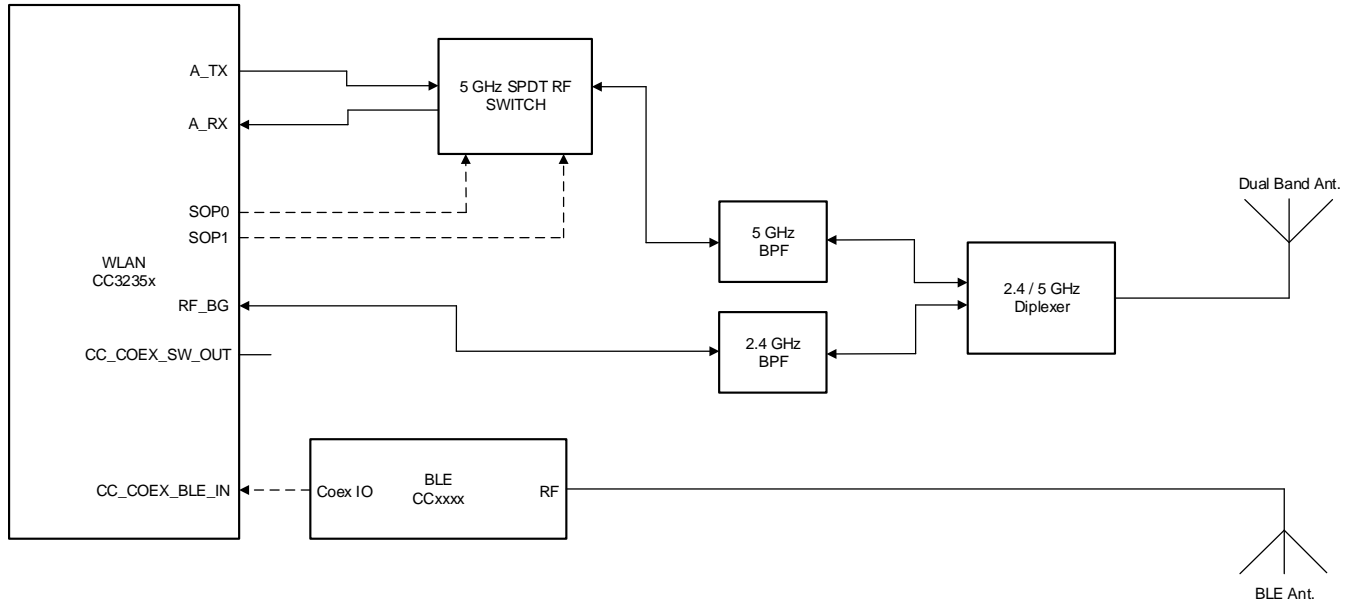


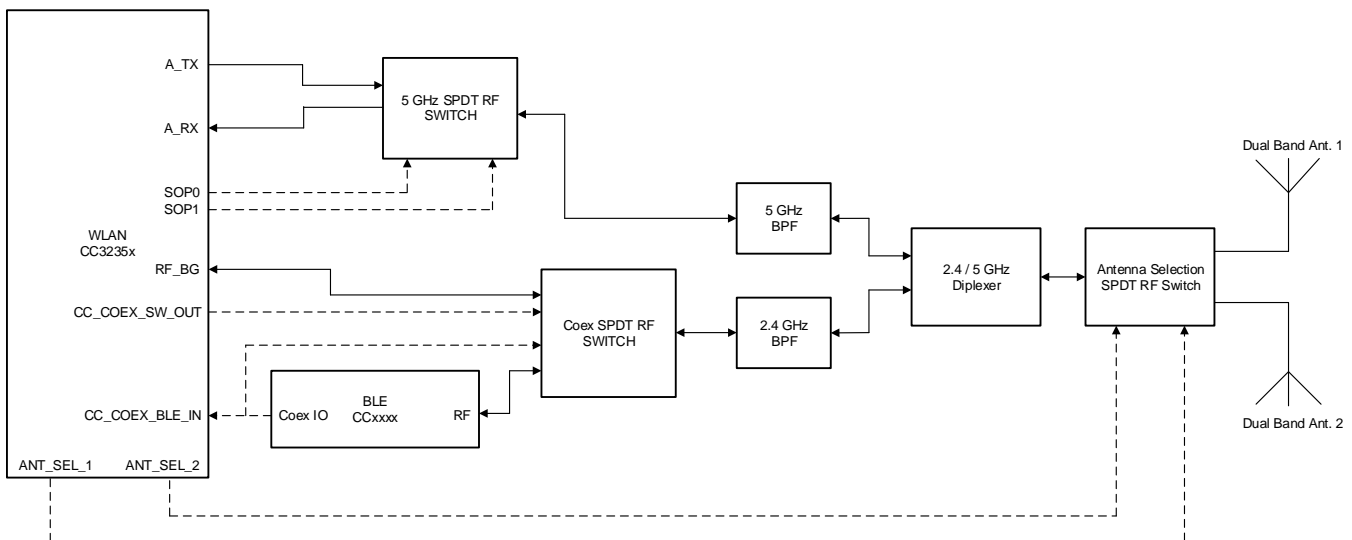
Figure 7-4. Dual Antenna Coexistence Solution with 5 GHz Wi-Fi

## 7.1.2 Antenna Selection

The CC3235x device is designed to also support antenna selection and is controlled from Image Creator. When enabled, there are 3 options possible options:

- ANT 1: When selected, the GPIOs that are defined for antenna selection will set the RF path for antenna 1.
- ANT 2: When selected, the GPIOs that are defined for antenna selection will set the RF path for antenna 2.
- Autoselect: When selected, during a scan and prior to connecting to an AP, CC3235x device will determine the best RF path and select the appropriate antenna <sup>(1)</sup> <sup>(2)</sup>. The result is the saved as part of the profile.

Figure 7-5 shows the implementation of a complete Bluetooth® low energy and WLAN coexistence network with the WLAN operating on either a 2.4- or a 5 GHz band with antenna selection. The SOP lines control the 5 GHz switch. The Coex switch is controlled by a GPIO signal from the BLE device and a GPIO signal from the CC3235x device. The Antenna switch is controlled by 2 GPIO lines from the CC3235x device.



**Figure 7-5. Antenna Selection Solution with Coexistence Solution and 5 GHz Wi-Fi**

- (1) When selecting Autoselect via the API, a reset is required in order for the CC3235x device to determine the best antenna for use.
- (2) Refer to the [Uniflash with Image Creator User Guide](#) for more information.



Figure 7-6 shows the antenna selection implementation for Wi-Fi, with BLE operating on it's own antenna. The SOP lines control the 5 GHz switch. Note in this implementation no Coex switch is required and only a single GPIO from the BLE device to the CC3235x device is required. The Antenna switch is controlled by 2 GPIO lines from the CC3235x device.

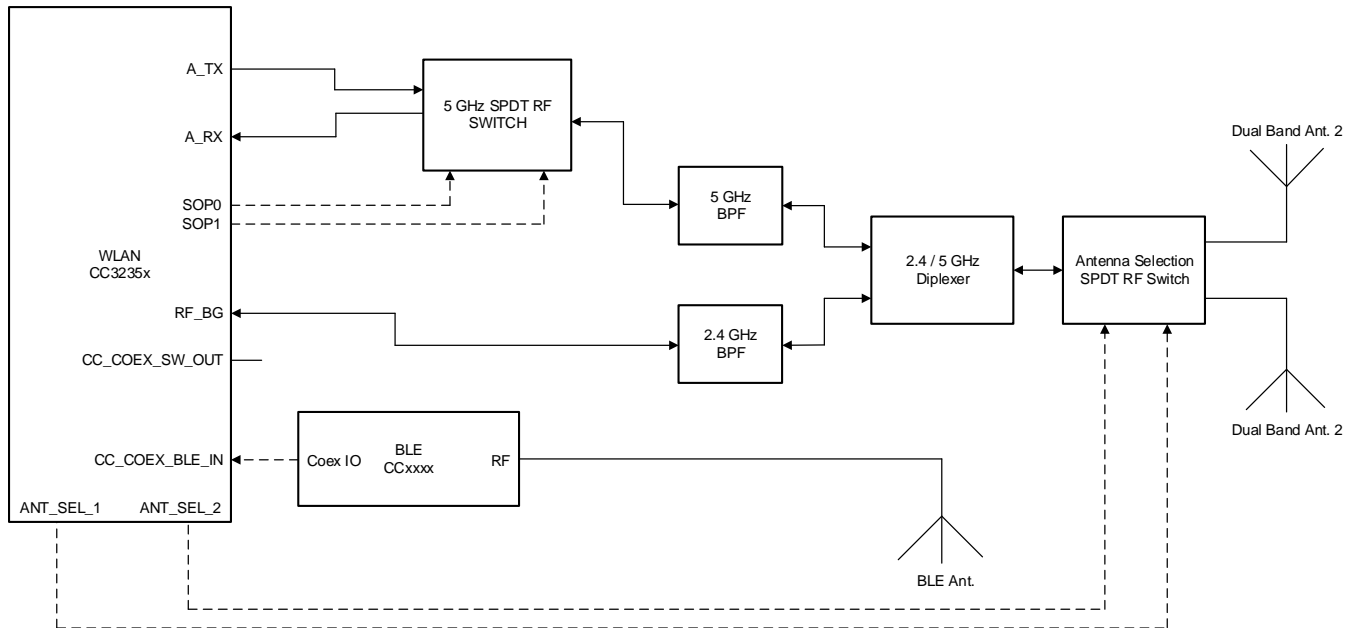


Figure 7-6. Coexistence Solution with Wi-Fi Antenna Selection and dedicated BLE antenna

### 7.1.3 Typical Application

Figure 7-7 shows the schematic of the engine area for the CC3235x device in the wide-voltage mode of operation, with the corresponding bill of materials show in Table 7-1. Figure 7-8 provides the schematic for the RF implementation with and without BLE/2.4 GHz coexistence, with the corresponding bill of materials shown in Table 7-2. For a full operation reference design, see the [CC3235x SimpleLink™ and Internet of Things Hardware Design Files](#).

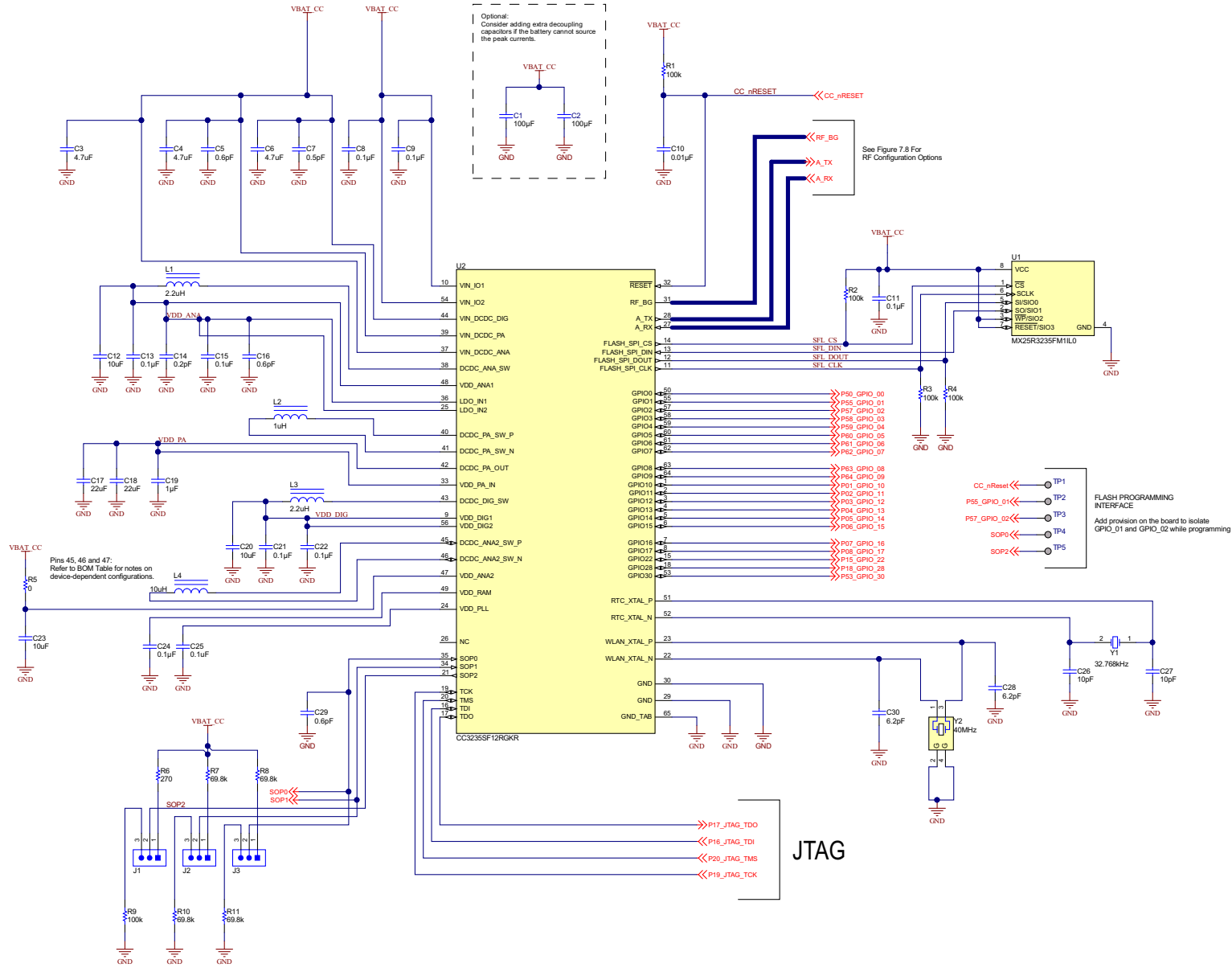


Figure 7-7. CC3235x Engine Area

**Table 7-1. Bill of Materials for CC3235x Engine Area**

Quantity	Designator	Value	Manufacturer	Part Number	Description
2	C1, C2	100 µF	Taiyo Yuden	LMK325ABJ107MMHT	CAP, CERM, 100 µF, 10 V, +/- 20%, X5R, AEC-Q200 Grade 3, 1210
3	C3, C4, C6	4.7 µF	Taiyo Yuden	JMK105BC6475MV-F	CAP, CERM, 4.7 µF, 6.3 V, +/- 20%, X6S, 0402
3	C5, C16, C29	0.6 pF	MuRata	GJM0335C1ER60BB01D	CAP, CERM, 0.6pF, 25 V, +/- 16%, COG/NPO, 0201
1	C7	0.5 pF	Murata	GJM0335C1ER50BB01D	CAP, CERM, 0.5 pF, 25 V, +/- 20%, COG/NPO, 0201
7	C8, C9, C11, C13, C21, C22, C24	0.1 µF	Walsin	CL05B104KO5NNNC	CAP, CERM, 0.1 µF, 16 V, +/- 10%, X7R, 0402
1	C10	0.01 µF	Walsin	0402B103K500CT	CAP, CERM, 0.01 µF, 50 V, +/- 10%, X7R, 0402
3	C12, C20, C23	10 µF	Taiyo Yuden	LMK107BC6106MA-T	CAP, CERM, 10 µF, 10 V, +/- 20%, X6S, 0603
1	C14	0.2 pF	MuRata	GJM0335C1ER20BB01D	CAP, CERM, 0.2pF, 25 V, +/- 50%, COG/NPO, 0201
2	C15, C25	0.1 µF	Samsung Electro-Mechanics	CL03A104KP3NNNC	CAP, CERM, 0.1 µF, 10 V, +/- 10%, X5R, 0201
2	C17, C18	22 µF	MuRata	GRM188C80G226ME15J	CAP, CERM, 22 µF, 4 V, +/- 20%, X6S, 0603
1	C19	1 µF	Walsin	CL05A105MP5NNNC	CAP, CERM, 1 µF, 10 V, +/- 20%, X5R, 0402
2	C26, C27	10 pF	Walsin	0402N100J500CT	CAP, CERM, 10 pF, 50 V, +/- 5%, COG/NPO, 0402
2	C28, C30	6.2 pF	Walsin	0402N6R2C500CT	CAP, CERM, 6.2 pF, 50 V, +/- 4%, COG/NPO, 0402
3	J1, J2, J3		Würth Elektronik	61300311121	Header, 2.54 mm, 3x1, Gold, TH
2	L1, L3	2.2 µH	MuRata	LQM2MPN2R2NG0	Inductor, Multilayer, Ferrite, 2.2 µH, 1.2 A, 0.11 ohm, SMD
1	L2	1 µH	MuRata	LQM2HPN1R0MG0L	Inductor, Multilayer, Ferrite, 1 µH, 1.6 A, 0.055 ohm, SMD
1	L4 <sup>(1)</sup>	10 µH	TDK	MLP2520S100MT0S1	Inductor, Multilayer, Ferrite, 10 µH, 0.7 A, 0.364 ohm, SMD
5	R1, R2, R3, R4, R9	100k	Vishay-Dale	CRCW0402100KJNED	RES, 100 k, 5%, 0.063 W, AEC-Q200 Grade 0, 0402
1	R5 <sup>(2)</sup>	0	Panasonic	ERJ-2GE0R00X	RES, 0, 5%, 0.063 W, 0402
1	R6	270	Vishay-Dale	CRCW0402270RJNED	RES, 270, 5%, 0.063 W, AEC-Q200 Grade 0, 0402
4	R7, R8, R10, R11	69.8k	Vishay-Dale	CRCW040269K8FKED	RES, 69.8 k, 1%, 0.063 W, AEC-Q200 Grade 0, 0402
1	U1		Macronix International Co., LTD	MX25R3235FM1IL0	Ultra low power, 32M-bit [x 1/x 2/x 4] CMOS MXSMIO(serial multi I/O) Flash memory, SOP-8
1	U2		Texas Instruments	CC3235SF12RGKR	SimpleLink Wi-Fi and Internet-of-Things Solution, a Single-Chip Wireless MCU, RGK0064B (VQFN-64)
1	Y1		Abracon Corporation	ABS07-32.768KHZ-9-T	Crystal, 32.768KHz, 9PF, SMD
1	Y2		TXC Corporation	8Y40072002	Crystal, 40 MHz, 8 pF, SMD

(1) For the CC3235SF device, L4 is populated. For the CC3235S device, L4 is not populated.

(2) For the CC3220SF device, R5 is not populated. For the CC3235S device if R5 is populated, Pin 45 can be used as GPIO\_31.

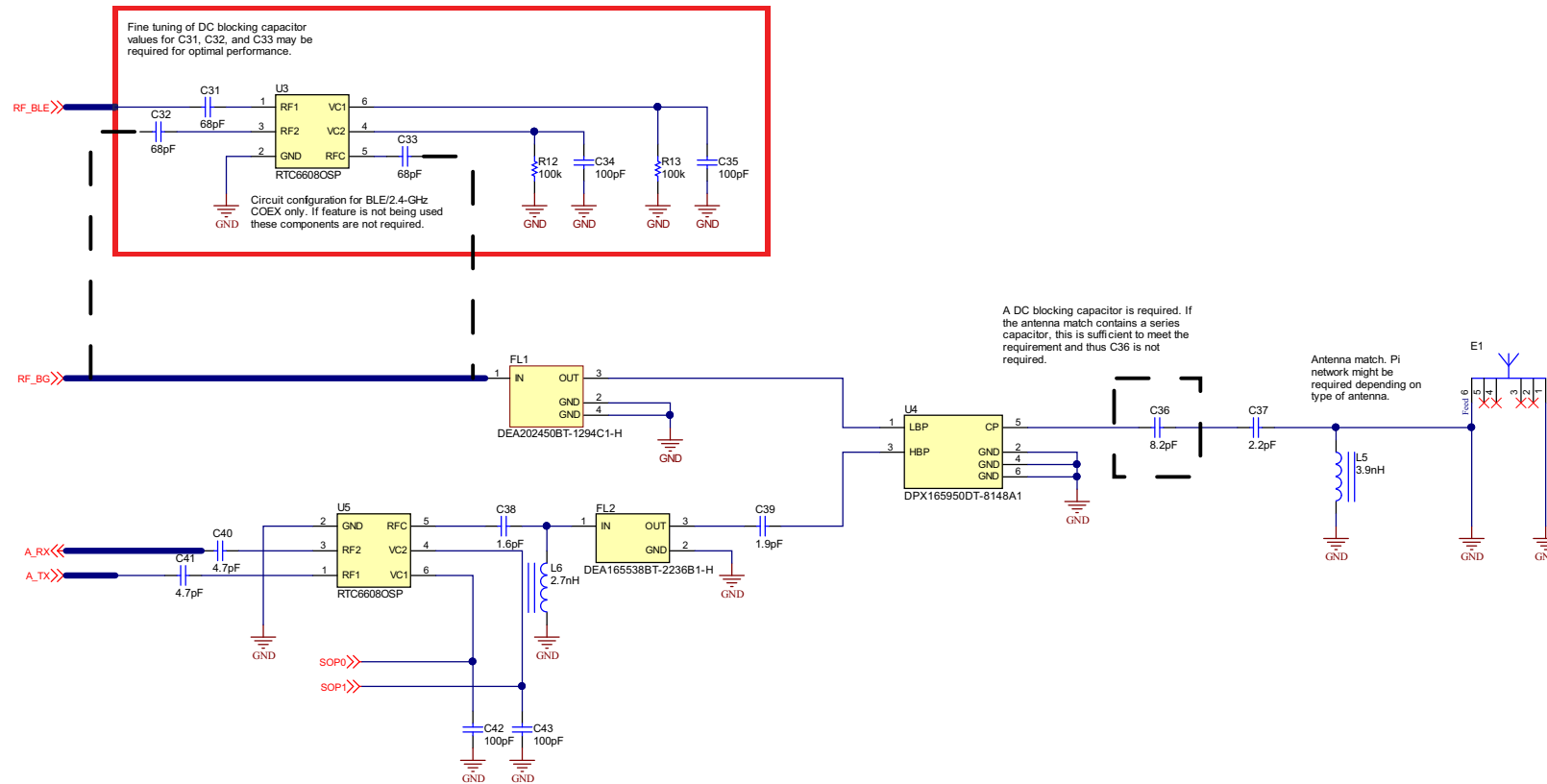


Figure 7-8. CC3235x RF Schematic Implementation with and without Coexistence

**NOTE**

The Following guidelines are recommended for implementation of the RF design:

- Ensure an RF path is designed with an impedance of 50 Ω
- Tuning of the antenna impedance  $\pi$  matching network is recommended after manufacturing of the PCB to account for PCB parasitics
- $\pi$  or L matching and tuning may be required between cascaded passive components on the RF path

**Table 7-2. Bill of Materials For CC3235x RF Section**

Quantity	Designator	Value	Manufacturer	Part Number	Description
3	C31 <sup>(1)</sup> , C32 <sup>(1)</sup> , C33 <sup>(1)</sup>	68 pF	Murata	GRM0335C1H680JA1D	CAP, CERM, 68 pF, 50 V, +/- 5%, C0G/NP0, 0201
4	C34 <sup>(1)</sup> , C35 <sup>(1)</sup> , C42, C43	100 pF	Yageo	CC0201JRNPO8BN101	CAP, CERM, 100 pF, 25 V, +/- 5%, C0G/NP0, 0201
1	C36	8.2 pF	Walsin	0402N8R2C500CT	CAP, CERM, 8.2 pF, 50 V, +/- 3%, C0G/NP0, 0402
1	C37	2.2 pF	MuRata	GJM1555C1H2R2BB01D	CAP, CERM, 2.2 pF, 50 V, +/- 4.5%, C0G/NP0, 0402
1	C38	1.6 pF	MuRata	GRM0335C1H1R6BA01D	CAP, CERM, 1.6 pF, 50 V, +/- 7%, C0G/NP0, 0201
1	C39	1.9 pF	MuRata	GJM1555C1H1R9WB01D	CAP, CERM, 1.9 pF, 50 V, +/- 2.6%, C0G/NP0, 0402
2	C40, C41	4.7 pF	MuRata	GRM0335C1H4R7BA01D	CAP, CERM, 4.7 pF, 50 V, +/- 3%, C0G/NP0, 0201
1	E1		Ethertronics	M830520	WLAN Antenna 802.11, SMD
1	FL1		TDK	DEA202450BT-1294C1-H	Multilayer Chip Band Pass Filter For 2.4 GHz W-LAN/Bluetooth, SMD
1	FL2		TDK	DEA165538BT-2236B1-H	Multilayer Band Pass Filter For 5 GHz W-LAN/LTE-U
1	L5	3.9 nH	MuRata	LQG15HS3N9S02D	Inductor, Multilayer, Air Core, 3.9 nH, 0.75 A, 0.14 ohm, SMD
1	L6	2.7 nH	MuRata	LQG15WH2N7C02D	Inductor, Multilayer, Air Core, 2.7 nH, 0.9 A, 0.07 ohm, AEC-Q200 Grade 1, SMD
2	R12 <sup>(1)</sup> , R13 <sup>(1)</sup>	100k	Vishay-Dale	CRCW0402100KJNED	RES, 100 k, 5%, 0.063 W, AEC-Q200 Grade 0, 0402
2	U3 <sup>(1)</sup> , U5		Richwave	RTC6608OSP	0.03 GHz-6 GHz SPDT Switch
1	U4		TDK	DPX165950DT-8148A1	Multilayer Diplexer for 2.4 GHz W-LAN & Bluetooth / 5 GHz W-LAN

(1) If the BLE/2.4 GHz Coexistence features is not used, these components are not required.

## 7.2 PCB Layout Guidelines

This section details the PCB guidelines to speed up the PCB design using the CC3235x VQFN device. Follow these guidelines ensures that the design will minimize the risk with regulatory certifications including FCC, ETSI, and CE. For more information, see [CC3135 and CC3235 SimpleLink™ Wi-Fi® and IoT Solution Layout Guidelines](#).

### 7.2.1 General PCB Guidelines

Use the following PCB guidelines:

- Verify the recommended PCB stackup in the PCB design guidelines, as well as the recommended layers for signals and ground.
- Ensure that the VQFN PCB footprint follows the information in [§ 9](#).
- Ensure that the VQFN PCB GND and solder paste follow the recommendations provided in [CC3135 and CC3235 SimpleLink™ Wi-Fi® and IoT Solution Layout Guidelines](#).
- Decoupling capacitors must be as close as possible to the VQFN device.

### 7.2.2 Power Layout and Routing

Three critical DC/DC converters must be considered for the CC3235x device.

- Analog DC/DC converter
- PA DC/DC converter
- Digital DC/DC converter

Each converter requires an external inductor and capacitor that must be laid out with care. DC current loops are formed when laying out the power components.

#### 7.2.2.1 Design Considerations

The following design guidelines must be followed when laying out the CC3235x device:

- Ground returns of the input decoupling capacitors (C13, C15, and C22) should be routed on Layer 2 using thick traces to isolate the RF ground from the noisy supply ground. This step is also required to meet the IEEE spectral mask specifications.
- Maintain the thickness of power traces to be greater than 12 mils. Take special consideration for power amplifier supply lines (pin 33, 40, 41, and 42), and all input supply pins (pin 37, 39, and 44).
- Ensure the shortest grounding loop for the PLL supply decoupling capacitor (pin 24).
- Place all decoupling capacitors as close to the respective pins as possible.
- Power budget—the CC3235x device can consume up to 450 mA for 3.3 V, 670 mA for 2.1 V, for 24 ms during the calibration cycle.
- Ensure the power supply is designed to source this current without any issues. The complete calibration (TX and RX) can take up to 17 mJ of energy from the battery over a time of 24 ms.
- The CC3235x device contains many high-current input pins. Ensure the trace feeding these pins can handle the following currents:
  - VIN\_DCDC\_PA input (pin 39) maximum 1 A
  - VIN\_DCDC\_ANA input (pin 37) maximum 600 mA
  - VIN\_DCDC\_DIG input (pin 44) maximum 500 mA
  - DCDC\_PA\_SW\_P (pin 40) and DCDC\_PA\_SW\_N (pin 41) switching nodes maximum 1 A
  - DCDC\_PA\_OUT output node (pin 42) maximum 1 A
  - DCDC\_ANA\_SW switching node (pin 38) maximum 600 mA
  - DCDC\_DIG\_SW switching node (pin 43) maximum 500 mA
  - VDD\_PA\_IN supply (pin 33) maximum 500 mA

Figure 7-9 shows the ground routing for the input decoupling capacitors.

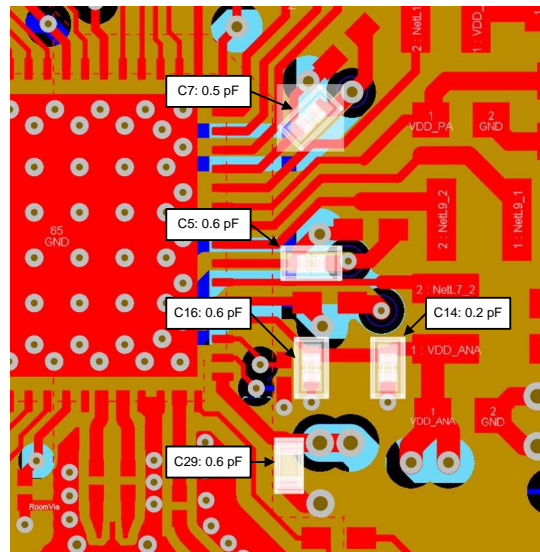


Figure 7-9. Ground Routing for Input Decoupling Capacitors

**NOTE**

The ground returns for the input capacitors are routed on layer two to reduce the EMI and improve the spectral mask. This routing must be strictly followed because it is critical for the overall performance of the device.

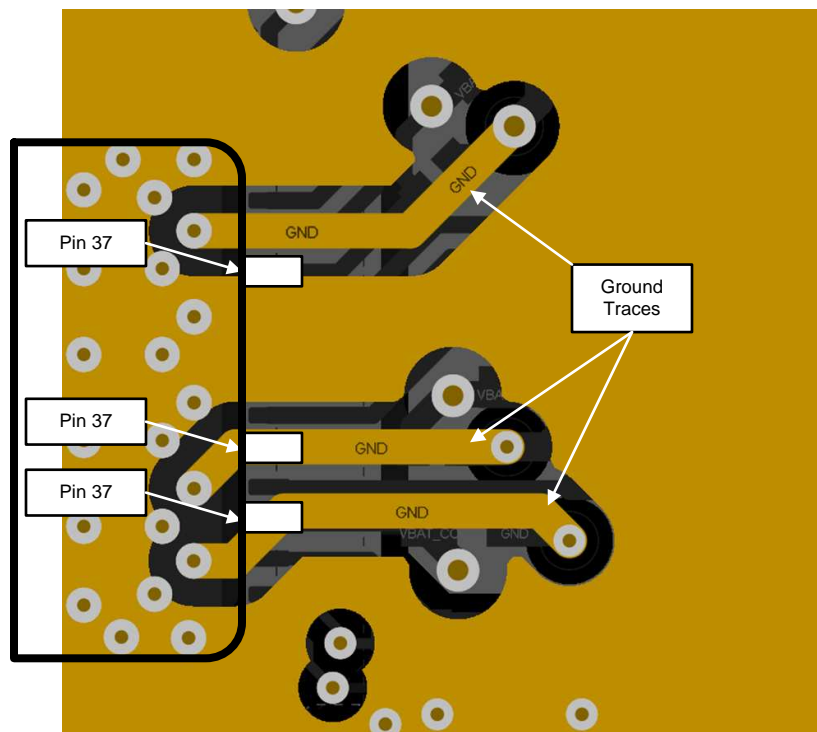


Figure 7-10. Ground Returns for Input Capacitors

### 7.2.3 Clock Interface Guidelines

The following guidelines are for the slow clock:

- The 32.768-kHz crystal must be placed close to the VQFN package.
- Ensure that the load capacitance is tuned according to the board parasitics to the frequency tolerance within  $\pm 150$  ppm.
- The ground plane on layer two is solid below the trace lanes, and there is ground around these traces on the top layer.

The following guidelines are for the fast clock:

- The 40-MHz crystal must be placed close to the VQFN package.
- Ensure that the load capacitance is tuned according to the board parasitics to the frequency tolerance within  $\pm 10$  ppm at room temperature. The total frequency across parts, temperature, and with aging must be  $\pm 20$  ppm to meet the WLAN specification.
- To avoid noise degradation, ensure that no high-frequency lines are routed close to the routing of the crystal pins.
- Ensure that crystal tuning capacitors are close to the crystal pads.
- Both traces (XTAL\_N and XTAL\_P) should be as close as possible to parallel and approximately the same length.
- The ground plane on layer two is solid below the trace lines, and there should be ground around these traces on the top layer.
- For frequency tuning, see [CC31xx & CC32xx Frequency Tuning](#).

### 7.2.4 Digital Input and Output Guidelines

The following guidelines are for the digital I/Os:

- Route SPI and UART lines away from any RF traces.
- Keep the length of the high-speed lines as short as possible to avoid transmission line effects.
- Keep the line lower than 1/10 of the rise time of the signal to ignore transmission line effects (required if the traces cannot be kept short). Place the resistor at the source end closer to the device that is driving the signal.
- Add a series-terminating resistor for each high-speed line (for example, SPI\_CLK or SPI\_DATA) to match the driver impedance to the line. Typical terminating-resistor values range from 27 to 36  $\Omega$  for a 50- $\Omega$  line impedance.
- Route high-speed lines with a continuous ground reference plane below it to offer good impedance throughout. This routing also helps shield the trace against EMI.
- Avoid stubs on high-speed lines to minimize the reflections. If the line must be routed to multiple locations, use a separate line driver for each line.
- If the lines are longer compared to the rise time, add series-terminating resistors near the driver for each high-speed line to match the driver impedance to the line. Typical terminating-resistor values range from 27 to 36  $\Omega$  for a 50- $\Omega$  line impedance.



### 7.2.5 RF Interface Guidelines

The following guidelines are for the RF interface. Follow guidelines specified in the vendor-specific antenna design guides (including placement of the antenna). Also see [CC3135 and CC3235 SimpleLink™ Wi-Fi® and IoT Solution Layout Guidelines](#) for general antenna guidelines.

- Ensure that the antenna is matched for 50-Ω. A  $\pi$ -matching network is recommended. Ensure that the  $\pi$  pad is available for tuning the matching network after PCB manufacture.
- A DC blocking capacitor is required before the antenna. If the antenna matching network contains a series capacitor, this is sufficient to meet the requirement.
- Ensure that the area underneath the BPFs pads have a solid plane on layer 2 and that the minimum filter requirements are met.
- Ensure that the area underneath the RF switch pads have a solid plane on layer 2 and that the minimum switch isolation requirements are met.
- Ensure that the area underneath the diplexer pads have a solid plane on layer 2 and that the minimum diplexer requirements are met.
- Verify that the Wi-Fi RF trace is a 50-Ω, impedance-controlled trace with a reference to solid ground.
- The RF trace bends must be made with gradual curves. Avoid 90-degree bends.
- The RF traces must not have sharp corners.
- There must be no traces or ground under the antenna section.
- The RF traces must have via stitching on the ground plane beside the RF trace on both sides.
- For optimal antenna performance, ensure adequate ground plane around the antenna on all layers.
- Ensure RF connectors for conducted testing are isolated from the top layer ground using vias.
- Maintain a controlled pad to trace shapes using filleted edges if necessary to avoid mismatch.
- Diplexers, switches, BPF, and other elements on the RF route should be isolated from the top layer ground using vias.

## 8 器件和文档支持

### 8.1 第三方产品免责声明

TI 发布的与第三方产品或服务有关的信息，不能构成与此类产品或服务或保修的适用性有关的认可，不能构成此类产品或服务单独或与任何 TI 产品或服务一起的表示或认可。

### 8.2 工具与软件

TI 提供大量的开发工具。此部分列出了用于评估器件性能、生成代码和开发解决方案的工具和软件。

有关开发工具和软件的最新列表，请访问 [CC3235 工具和软件](#) 产品页面。用户也可以点击 [CC3235 工具和软件](#) 页面右上角的“通知我”按钮，随时获取有关 CC3235x 器件的最新消息。

#### 开发工具

**引脚复用工具** 支持的器件包括：CC3200、CC3220x 和 CC3235x。

**Pin Mux** 工具是一款软件工具，可提供图形用户界面 (GUI) 用于配置引脚多路复用设置、解决冲突以及指定 TI MPU 的 I/O 单元特性。结果采用 C 头文件/代码文件的形式输出，可导入到软件开发套件 (SDK) 中或用于配置客户的定制软件。**Pin Mux** 工具版本 3 增加了一个功能，可自动选择多路复用器配置以满足输入要求。

**SimpleLink™ Wi-Fi® Starter Pro** 支持的器件包括：CC3100、CC3200、CC3120R、CC3220x、CC3135 和 CC3235x。

**SimpleLink™ Wi-Fi® Starter Pro** 移动应用是一款用于配置 **SimpleLink™** 的新型移动应用。该应用与嵌入式配置库和在设备端运行的示例搭配使用（请参阅 [SimpleLink™ Wi-Fi® SDK 插件](#) 和 [TI SimpleLink™ CC32XX 软件开发套件 \(SDK\)](#)）。如需使用 **SimpleLink™ Wi-Fi®** 产品进行 Wi-Fi® 配置，TI 建议使用新配置版本。此配置版本可以借助反馈和备用选项实现高级 AP 模式和 **SmartConfig™** 技术配置，确保成功完成处理。客户可以使用嵌入式库和移动库来集成其最终产品。

**SimpleLink™ CC32XX 软件开发套件 (SDK)** 支持 CC3235x 器件。

**SimpleLink™ CC32XX SDK** 包含用于 CC3235 可编程 MCU 的驱动程序、30 个以上的示例应用，以及使用该解决方案所需的文档。它还包含闪存编程器、系统文件和用户文件（证书、网页等等）。闪存编程器是一种命令行工具，用于闪存软件以及配置网络和软件参数（SSID、接入点通道、网络配置文件、BS NIEW）。此 SDK 可与 TI 的 **SimpleLink™ Wi-Fi® CC3235 LaunchPad™** 开发套件配合使用。

**适用于 TI 微控制器 (MCU)、Sitara 处理器和 SimpleLink 器件的 Uniflash 独立闪存工具** 支持的器件包括：CC3120R、CC3220x、CC3135 和 CC3235x。

**CCS Uniflash** 是一个独立工具，用于编程 TI MCU 的片上闪存和 **Sitara™** 处理器的板载闪存。Uniflash 具有 GUI、命令行和脚本界面。CCS UniFlash 免费提供。

**SimpleLink™ Wi-Fi® 无线电测试工具** 支持的器件包括：CC3100、CC3200、CC3120R、CC3220、CC3135 和 CC3235x。

SimpleLink™ Wi-Fi® 无线电测试工具是一款基于 Windows 系统的软件工具，用于在开发和认证过程中对 SimpleLink™ Wi-Fi® CC3x20 和 CC3x35 设计进行射频评估和测试。通过手动将无线电设置为传输或接收模式，该工具可提供低级无线电测试功能。使用此工具需要熟悉并了解无线电电路理论和无线电测试方法的知识。

专为物联网 (IoT) 打造的 SimpleLink™ Wi-Fi® CC31xx 和 CC32xx 系列器件包含片上 Wi-Fi®、互联网和稳固的安全协议，您无需事先具备 Wi-Fi® 经验就可以以更快的速度进行开发。有关这些器件的更多信息，请访问 SimpleLink™ Wi-Fi® 系列 Internet-on-a-chip™ 解决方案。

适用于 TI 微控制器 (MCU)、Sitara™ 处理器和 SimpleLink™ 器件的 UniFlash 独立闪存工具 CCS UniFlash 是一个独立工具，用于编程 TI MCU 的片上闪存和 Sitara 处理器的板载™ 闪存。UniFlash 具有 GUI、命令行和脚本接口。CCS UniFlash 免费提供。

### TI 设计和参考设计

TI Designs 参考设计库是一个涵盖模拟、嵌入式处理器和连接的强大参考设计资源库。所有 TI 设计由 TI 专家构建，旨在帮助您快速开始系统设计，其中包括原理图或方框图、BOM 和设计文件，助您加速产品上市时间。

### 8.3 固件更新

即使未发布 相关计划，TI 也会不时更新此模块相应服务包中的功能。由于更改不断发生，TI 建议用户在其用于生产的模块中使用最新服务包。

若要随时获取更新消息，请点击产品页面右上角的 SDK“通知我”按钮，或者访问 SimpleLink™ CC32XX SDK。

### 8.4 器件命名规则

为了标示产品开发周期所处的阶段，TI 为 3235CCx 器件和支持工具的部件号分配了前缀（请参阅图 8-1）。

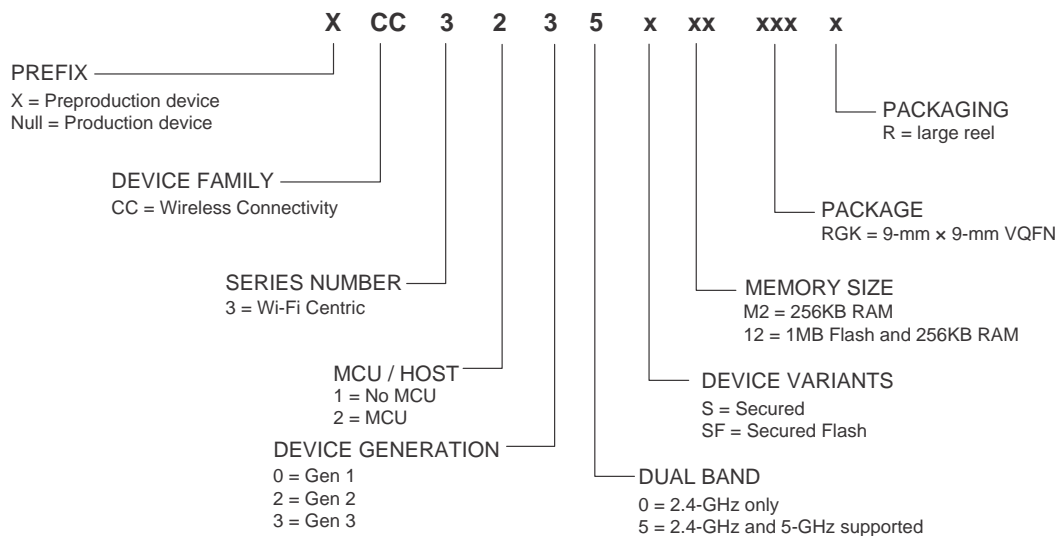


图 8-1. CC3235x 器件命名规则

## 8.5 文档支持

要接收文档更新通知（包括器件勘误表），请转至 [ti.com](http://ti.com) (CC3235)。请单击右上角的“通知我”按钮。点击注册后，即可收到产品信息更改每周摘要（如有）。有关更改的详细信息，请查阅已修订文档的修订历史记录。下面列出了介绍处理器、相关外设以及其他配套技术资料的最新文档。

以下文档为 CC3235 器件提供支持。

### 应用报告

**CC3135 和 CC3235 SimpleLink™ Wi-Fi® 嵌入式编程用户手册** CC3135 和 CC3235 SimpleLink Wi-Fi 嵌入式编程用户手册

**SimpleLink™ CC3135、CC3235 Wi-Fi® Internet-on-a-chip™ 网络子系统电源管理** 该应用报告介绍了进行电源管理和延长电池寿命的最佳实践，适用于嵌入式低功耗 Wi-Fi 器件，例如德州仪器 (TI) 提供的 SimpleLink Wi-Fi 片上互联网解决方案。

**SimpleLink™ CC31xx、CC32xx Wi-Fi® Internet-on-a-chip™ 解决方案内置安全特性** 德州仪器 (TI) 提供的 SimpleLink Wi-Fi CC31xx 和 CC32xx 片上互联网系列器件具有多种内置的安全特性，以便帮助开发人员解决各种安全需求，且不会对主微控制器 (MCU) 造成任何处理负担。本文档介绍这些安全相关的特性，并提供有关在实际系统实现环境中利用每个特性的建议。

**SimpleLink™ CC3135、CC3235 Wi-Fi® 和物联网无线更新** 本文档介绍德州仪器 (TI) 所提供的 SimpleLink Wi-Fi CC3x35 系列器件的 OTA 库，并说明了如何准备新的云就绪更新供 OTA 库下载。

**SimpleLink™ CC3135、CC3235 Wi-Fi® Internet-on-a-chip™ 解决方案器件配置** 本指南介绍了为 SimpleLink Wi-Fi 器件提供连接无线网络所需信息（网络名称、密码等）的配置过程。

**将 TI 的 Wi-Fi® Alliance 认证转移到基于 SimpleLink™ 的产品** 本文档介绍如何使用 Wi-Fi® Alliance (WFA) 衍生认证策略将德州仪器 (TI) 已获得的 WFA 认证转移到您开发的系统之中。

**在 SimpleLink™ CC3135 和 CC3235 Wi-Fi® 以及物联网器件上使用串行闪存** 本应用手册分为两个部分。第一部分提供重要指南以及在选择和嵌入与 CC3135 和 CC3235 (CC3x35) 器件配对的串行闪存时应考虑的最佳实践设计技巧。第二部分介绍文件系统，同时为使用 CC3x35 器件的系统设计人员提供相关指南及注意事项。

## 用户指南

**SimpleLink™ Wi-Fi® 以及物联网 CC31xx 和 CC32xx 网络处理器** 本文档为软件 (SW) 程序员提供了使用 SimpleLink Wi-Fi 器件网络子系统所需的全部知识。本指南提供了编写强大、优化型网络主机应用的基本指南，并介绍了网络子系统的功能。本指南包含一些示例代码屏幕截图，以便使用户明白如何使用主机驱动程序。可在正式的软件开发套件 (SDK) 中找到更全面的代码示例。本指南未提供有关主机驱动程序 API 的详细说明。

**SimpleLink™ Wi-Fi® CC3135 和 CC3235 以及物联网解决方案布局指南** 本文档提供了用于德州仪器 (TI) CC3135 和 CC3235 SimpleLink Wi-Fi 系列器件的 4 层 PCB 设计指南。CC3135 和 CC3235 器件易于布置，采用四方扁平无引线 (QFN) 封装。当设计电路板时，请遵循本文档中的建议，以优化电路板的性能。

**SimpleLink™ CC3235 Wi-Fi® LaunchPad™ 开发套件硬件** CC3235 SimpleLink LaunchPad 开发套件 (LAUNCHXL-CC3235) 是一个低成本评估平台，适用于基于 Arm Cortex-M4 的 MCU。LaunchPad 设计着重强调 CC3235 片上互联网解决方案和 Wi-Fi 功能。CC3235 LaunchPad 还具有温度和加速计传感器、可编程用户按钮、用于自定义应用的三个 LED 和用于调试的板载仿真。CC3235 LaunchPad XL 接口采用可堆叠接头，在与多种现有的 BoosterPack™ 插件模块附加板上的其他外设连接时，可轻松扩展 LaunchPad 的功能，例如图形显示、音频编解码器、天线选择、环境感应等。

**SimpleLink™ Wi-Fi® 和 Internet-on-a-chip™ CC3135 和 CC3235 解决方案无线电工具** 这款无线电工具作为直接接入无线电的控制面板，可用于射频 (RF) 评估及获取认证。本指南介绍如何使该工具在德州仪器 (TI) 评估平台（例如 CC3235 器件的 BoosterPack™ 和 FTDI 仿真板或 CC3235 器件的 LaunchPad™）上无缝运行。

面向移动应用的 **SimpleLink™ Wi-Fi® CC3135 和 CC3235 配置** 本指南介绍了适用于移动应用的 TI SimpleLink Wi-Fi 配置解决方案，特别是介绍如何使用 Android™ 和 iOS® 符合 UI 要求的构建块、网络和构建移动应用所需的配置 API。

## 更多文献

**CC3235 SimpleLink™ Wi-Fi® 和物联网技术参考手册** 本技术参考手册详细介绍了 CC3235 SimpleLink™ Wi-Fi® MCU 的模块和外设。每个说明均从一般意义上介绍模块或外设。可能并未对所有器件上的所有模块或外设的所有特性和功能都作出介绍。引脚功能、内部信号连接和操作参数都因器件不同而各异。有关这些细节，用户应查阅具体器件的产品说明书。

**CC3x35 SimpleLink™ Wi-Fi® 硬件设计检查清单**

**CC3235S/CC3235SF SimpleLink™ Wi-Fi® LaunchPad™ 设计文件**

## 8.6 相关链接

下表列出了快速访问链接。类别包括技术文档、支持和社区资源、工具和软件，以及立即订购快速访问。

表 8-1. 相关链接

器件	产品文件夹	立即订购	技术文档	工具与软件	支持和社区
CC3235S	<a href="#">单击此处</a>	<a href="#">单击此处</a>	<a href="#">单击此处</a>	<a href="#">单击此处</a>	<a href="#">单击此处</a>
CC3235SF	<a href="#">单击此处</a>	<a href="#">单击此处</a>	<a href="#">单击此处</a>	<a href="#">单击此处</a>	<a href="#">单击此处</a>

## 8.7 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

**TI E2E™ Online Community** The TI engineer-to-engineer (E2E) community was created to foster collaboration among engineers. At [e2e.ti.com](http://e2e.ti.com), you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**TI Embedded Processors Wiki** Established to help developers get started with Embedded Processors from Texas Instruments and to foster innovation and growth of general knowledge about the hardware and software surrounding these devices.

## 8.8 商标

SimpleLink, Internet-on-a chip, LaunchPad, BoosterPack, SmartConfig, 板载, E2E are trademarks of Texas Instruments.

Arm, Cortex, Thumb are registered trademarks of Arm Limited.

低功耗蓝牙 is a registered trademark of Bluetooth SIG Inc.

IOS is a registered trademark of Cisco.

Android is a trademark of Google LLC.

Macrocell is a trademark of Kappa Global Inc.

获得 Wi-Fi Alliance, WPA, WPA2 are trademarks of Wi-Fi Alliance.

Wi-Fi CERTIFIED 认证, Wi-Fi, Wi-Fi Direct are registered trademarks of Wi-Fi Alliance.

All other trademarks are the property of their respective owners.

## 8.9 静电放电警告



ESD 可能会损坏该集成电路。德州仪器 (TI) 建议通过适当的预防措施处理所有集成电路。如果不遵守正确的处理措施和安装程序，可能会损坏集成电路。

ESD 的损坏小至导致微小的性能降级，大至整个器件故障。精密的集成电路可能更容易受到损坏，这是因为非常细微的参数更改都可能会导致器件与其发布的规格不相符。

## 8.10 Export Control Notice

Recipient agrees to not knowingly export or re-export, directly or indirectly, any product or technical data (as defined by the U.S., EU, and other Export Administration Regulations) including software, or any controlled product restricted by other applicable national regulations, received from disclosing party under nondisclosure obligations (if any), or any direct product of such technology, to any destination to which such export or re-export is restricted or prohibited by U.S. or other applicable laws, without obtaining prior authorization from U.S. Department of Commerce and other competent Government authorities to the extent required by those laws.

## 8.11 Glossary

**TI Glossary** This glossary lists and explains terms, acronyms, and definitions.

## 9 机械、封装和可订购信息

### 9.1 封装信息

以下页面包含机械、封装和可订购信息。这些信息是指定器件的最新可用数据。数据如有变更，恕不另行通知，且不会对此文档进行修订。如需获取此数据表的浏览器版本，请查阅左侧的导航栏。



## 9.1.1 Package Option Addendum

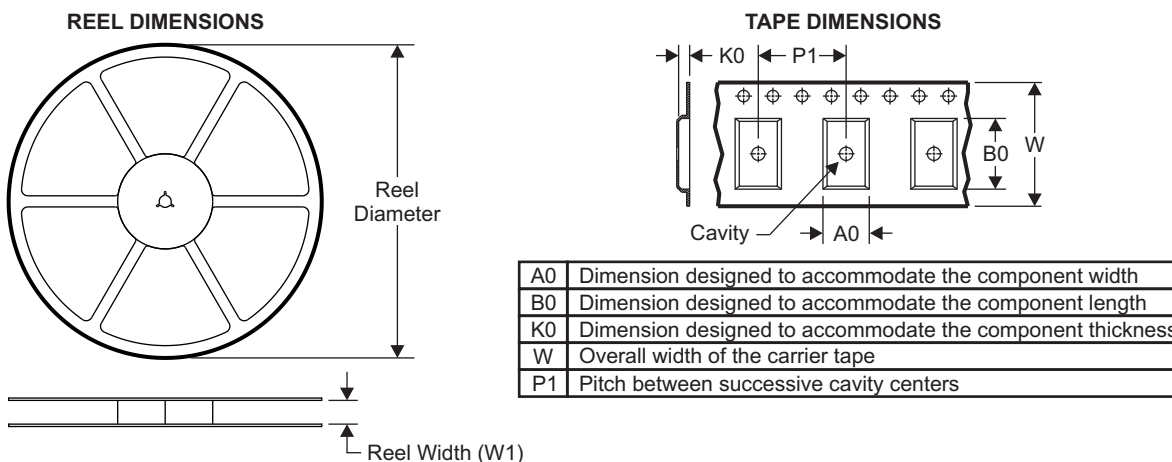
### 9.1.1.1 Packaging Information

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/Ball Finish <sup>(3)</sup>	MSL Peak Temp <sup>(4)</sup>	Op Temp (°C)	Device Marking <sup>(5) (6)</sup>
CC3235SM2RGKR	ACTIVE	VQFN	RGK	64	2500	Green (RoHS & no Sb/Br)	CU NIPDAU   CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC3235SM2
CC3235SF12RGKR	ACTIVE	VQFN	RGK	64	2500	Green (RoHS & no Sb/Br)	CU NIPDAU   CU NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC3235SF12

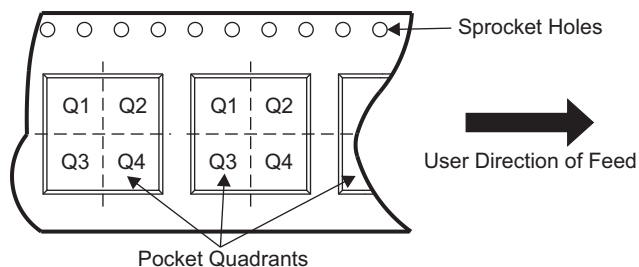
- (1) The marketing status values are defined as follows:  
**ACTIVE:** Product device recommended for new designs.  
**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.  
**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.  
**PRE\_PROD:** Unannounced device, not in production, not available for mass market, nor on the web, samples not available.  
**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.  
**OBSOLETE:** TI has discontinued the production of the device.
- (2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.  
**TBD:** The Pb-Free/Green conversion plan has not been defined.  
**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.  
**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.  
**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material).
- (3) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.
- (4) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (5) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (6) Multiple Device markings will be inside parentheses. Only on Device Marking contained in parentheses and separated by a "-" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

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9.1.1.2 Tape and Reel Information

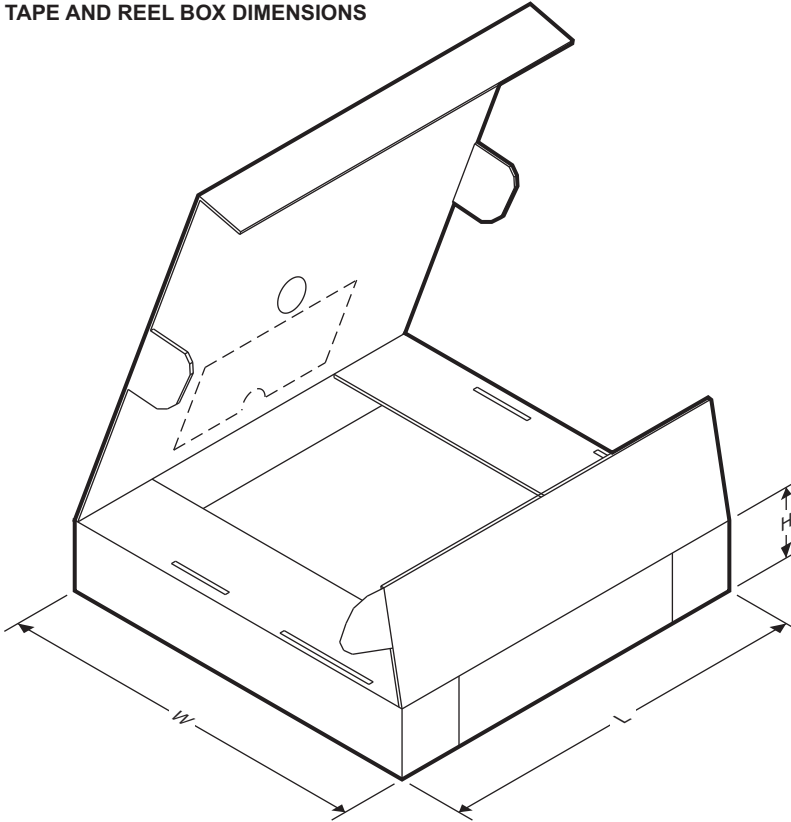


QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CC3235SM2RGKR	VQFN	RGK	64	2500	330.0	16.4	9.3	9.3	1.1	12.0	16.0	Q2
CC3235SF12RGKR	VQFN	RGK	64	2500	330.0	16.4	9.3	9.3	1.1	12.0	16.0	Q2

**TAPE AND REEL BOX DIMENSIONS**



Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
CC3235SM2RGKR	VQFN	RGK	64	2500	367.0	367.0	38.0
CC3235SF12RGKR	VQFN	RGK	64	2500	367.0	367.0	38.0

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**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
CC3235SF12RGKR	ACTIVE	VQFN	RGK	64	2500	Green (RoHS & no Sb/Br)	NIPDAU   NIPDAUAG	Level-3-260C-168 HR	-40 to 85	CC3235SF 12	<a href="#">Samples</a>
CC3235SM2RGKR	ACTIVE	VQFN	RGK	64	2500	Green (RoHS & no Sb/Br)	NIPDAU	Level-3-260C-168 HR	-40 to 85	CC3235S M2	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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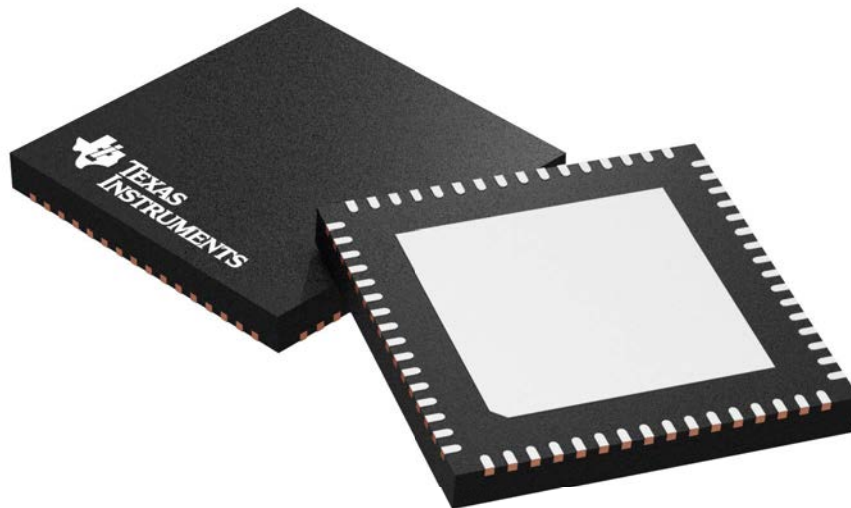


**RGK 64**

**GENERIC PACKAGE VIEW**

**VQFN - 1 mm max height**

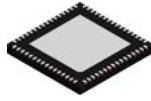
PLASTIC QUAD FLATPACK - NO LEAD



Images above are just a representation of the package family, actual package may vary.  
Refer to the product data sheet for package details.

4211520/D

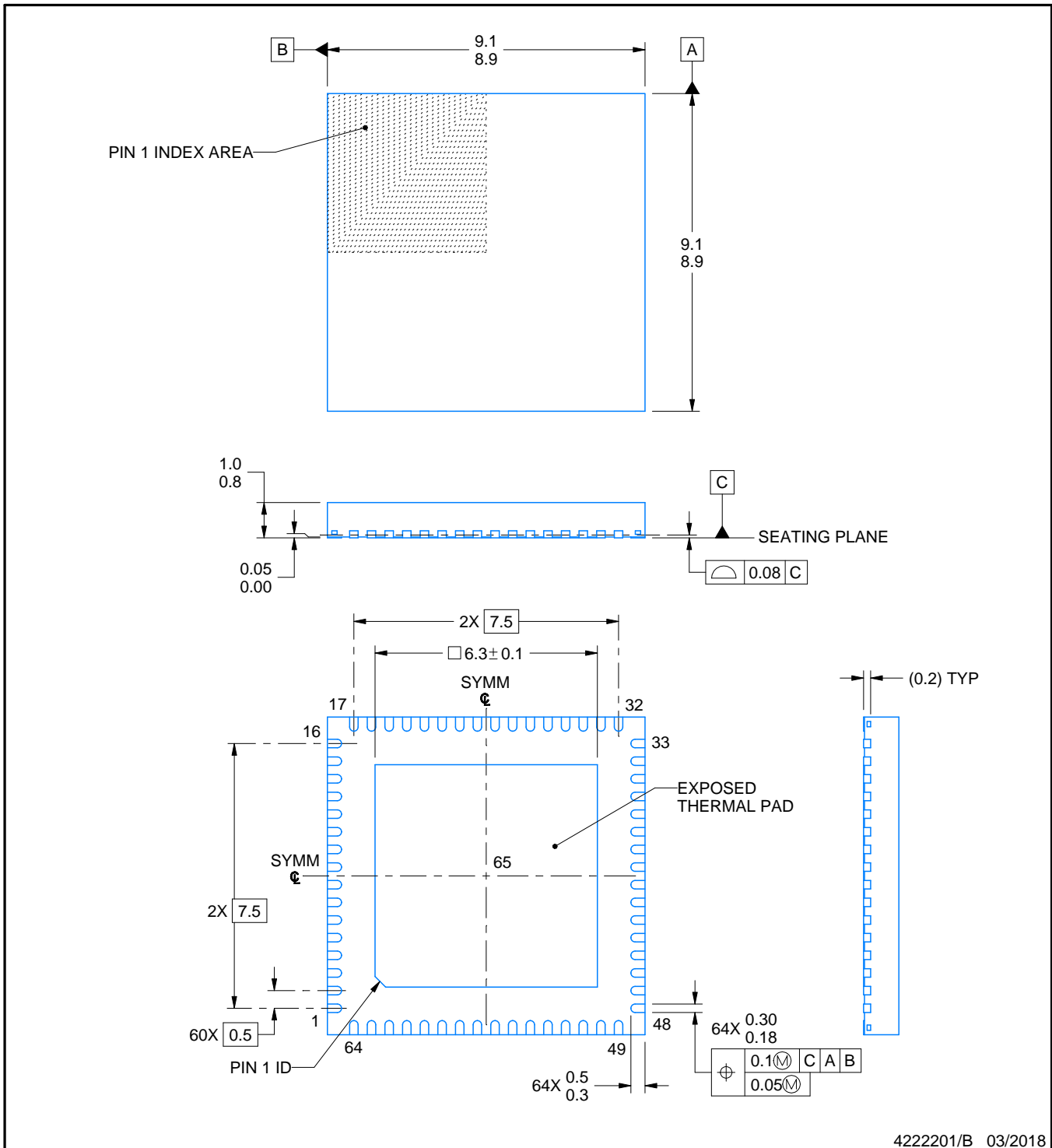
# RGK0064B



## PACKAGE OUTLINE

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



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### NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

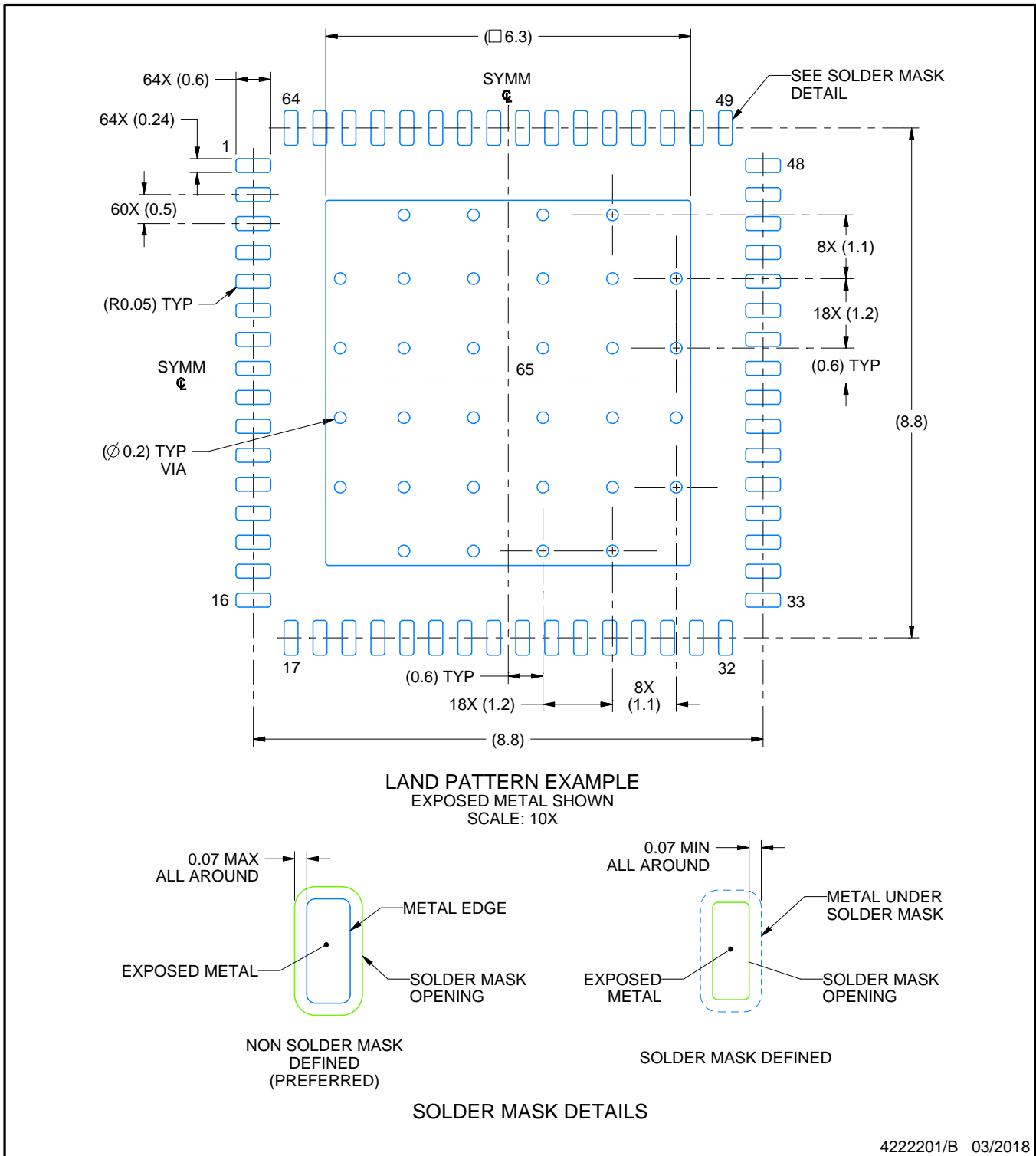


# EXAMPLE BOARD LAYOUT

**RGK0064B**

**VQFN - 1 mm max height**

PLASTIC QUAD FLATPACK - NO LEAD



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NOTES: (continued)

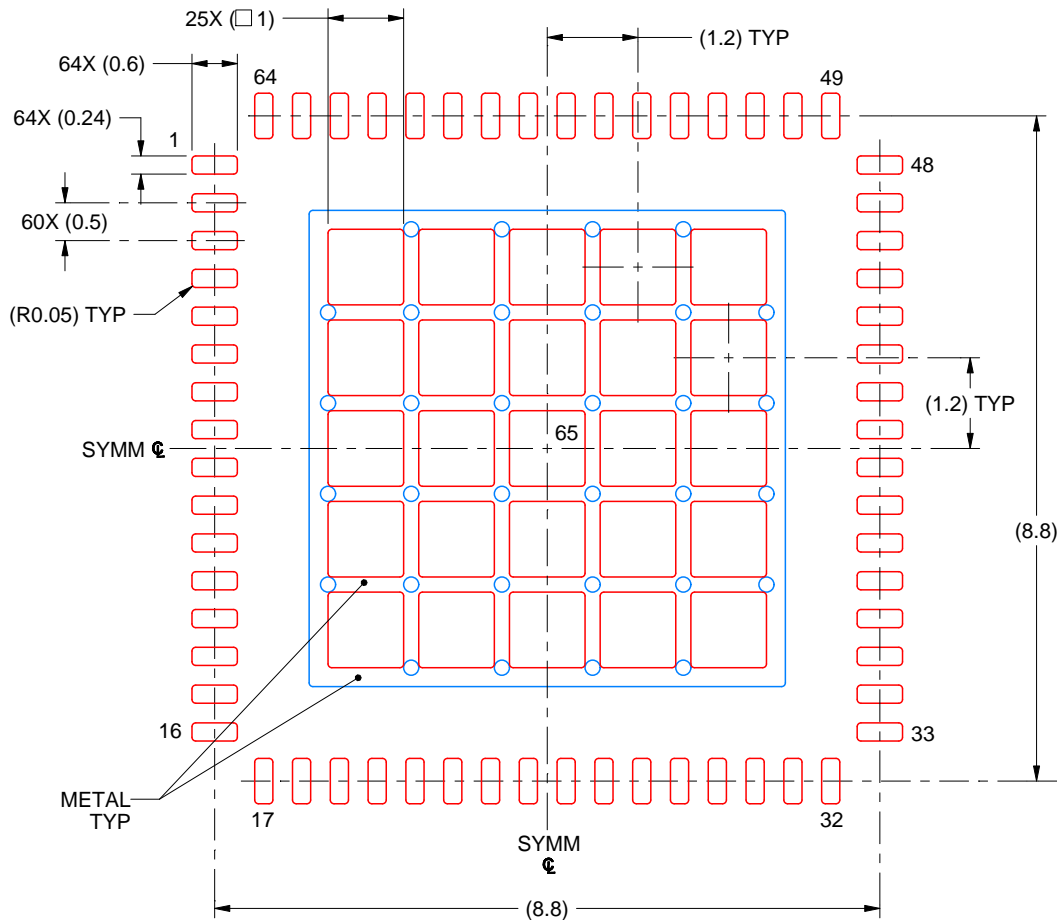
4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 ([www.ti.com/lit/sluea271](http://www.ti.com/lit/sluea271)).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

# EXAMPLE STENCIL DESIGN

RGK0064B

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



SOLDER PASTE EXAMPLE  
 BASED ON 0.1 MM THICK STENCIL  
 SCALE: 10X

EXPOSED PAD 65  
 63% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE

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NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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